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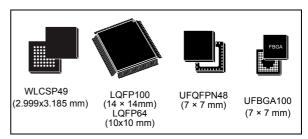
STM32F411xC STM32F411xE

Arm[®] Cortex[®]-M4 32b MCU+FPU, 125 DMIPS, 512KB Flash, 128KB RAM, USB OTG FS, 11 TIMs, 1 ADC, 13 comm. interfaces

Datasheet - production data

Features

- Dynamic Efficiency Line with BAM (Batch Acquisition Mode)
 - 1.7 V to 3.6 V power supply
 - 40°C to 85/105/125 °C temperature range
- Core: Arm[®] 32-bit Cortex[®]-M4 CPU with FPU, Adaptive real-time accelerator (ART Accelerator[™]) allowing 0-wait state execution from Flash memory, frequency up to 100 MHz, memory protection unit, 125 DMIPS/1.25 DMIPS/MHz (Dhrystone 2.1), and DSP instructions
- Memories
 - Up to 512 Kbytes of Flash memory
 - 128 Kbytes of SRAM
- · Clock, reset and supply management
 - 1.7 V to 3.6 V application supply and I/Os
 - POR, PDR, PVD and BOR
 - 4-to-26 MHz crystal oscillator
 - Internal 16 MHz factory-trimmed RC
 - 32 kHz oscillator for RTC with calibration
 - Internal 32 kHz RC with calibration
- Power consumption
 - Run: 100 µA/MHz (peripheral off)
 - Stop (Flash in Stop mode, fast wakeup time): 42 μA Typ @ 25C; 65 μA max @25 °C
 - Stop (Flash in Deep power down mode, slow wakeup time): down to 9 μA @ 25 °C; 28 μA max @25 °C
 - Standby: 1.8 μ A @25 °C / 1.7 V without RTC; 11 μ A @85 °C @1.7 V
 - V_{BAT} supply for RTC: 1 μA @25 °C
- 1×12-bit, 2.4 MSPS A/D converter: up to 16 channels
- General-purpose DMA: 16-stream DMA controllers with FIFOs and burst support
- Up to 11 timers: up to six 16-bit, two 32-bit timers up to 100 MHz, each with up to four IC/OC/PWM or pulse counter and quadrature (incremental) encoder input, two watchdog timers (independent and window) and a SysTick timer



- Debug mode
 - Serial wire debug (SWD) & JTAG interfaces
 - Cortex[®]-M4 Embedded Trace Macrocell™
- Up to 81 I/O ports with interrupt capability
 - Up to 78 fast I/Os up to 100 MHz
 - Up to 77 5 V-tolerant I/Os
- Up to 13 communication interfaces
 - Up to 3 x I²C interfaces (SMBus/PMBus)
 - Up to 3 USARTs (2 x 12.5 Mbit/s, 1 x 6.25 Mbit/s), ISO 7816 interface, LIN, IrDA, modem control)
 - Up to 5 SPI/I2Ss (up to 50 Mbit/s, SPI or I2S audio protocol), SPI2 and SPI3 with muxed full-duplex I²S to achieve audio class accuracy via internal audio PLL or external clock
 - SDIO interface (SD/MMC/eMMC)
 - Advanced connectivity: USB 2.0 full-speed device/host/OTG controller with on-chip PHY
- CRC calculation unit
- 96-bit unique ID
- RTC: subsecond accuracy, hardware calendar
- All packages (WLCSP49, LQFP64/100, UFQFPN48, UFBGA100) are ECOPACK[®]2

Table 1. Device summary

Reference	Part number		
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STM32F411xE	STM32F411CE, STM32F411RE, STM32F411VE		

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1 Introduction

This datasheet provides the description of the STM32F411xC/xE microcontrollers.

The STM32F411xC/xE datasheet should be read in conjunction with RM0383 reference manual which is available from the STMicroelectronics website *www.st.com*. It includes all information concerning Flash memory programming.

For information on the Cortex[®]-M4 core, please refer to the Cortex[®]-M4 programming manual (PM0214) available from *www.st.com*.

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2 Description

The STM32F411xC/xE devices are based on the high-performance Arm[®] Cortex[®] -M4 32-bit RISC core operating at a frequency of up to 100 MHz. The Cortex[®]-M4 core features a Floating point unit (FPU) single precision which supports all Arm single-precision data-processing instructions and data types. It also implements a full set of DSP instructions and a memory protection unit (MPU) which enhances application security.

The STM32F411xC/xE belongs to the STM32 Dynamic EfficiencyTM product line (with products combining power efficiency, performance and integration) while adding a new innovative feature called Batch Acquisition Mode (BAM) allowing to save even more power consumption during data batching.

The STM32F411xC/xE incorporate high-speed embedded memories (up to 512 Kbytes of Flash memory, 128 Kbytes of SRAM), and an extensive range of enhanced I/Os and peripherals connected to two APB buses, two AHB bus and a 32-bit multi-AHB bus matrix.

All devices offer one 12-bit ADC, a low-power RTC, six general-purpose 16-bit timers including one PWM timer for motor control, two general-purpose 32-bit timers. They also feature standard and advanced communication interfaces.

- Up to three I²Cs
- Five SPIs
- Five I²Ss out of which two are full duplex. To achieve audio class accuracy, the I²S
 peripherals can be clocked via a dedicated internal audio PLL or via an external clock
 to allow synchronization.
- Three USARTs
- SDIO interface
- USB 2.0 OTG full speed interface

The STM32F411xC/xE operate in the - 40 to + 125 °C temperature range from a 1.7 (PDR OFF) to 3.6 V power supply. A comprehensive set of power-saving mode allows the design of low-power applications.

These features make the STM32F411xC/xE microcontrollers suitable for a wide range of applications:

- Motor drive and application control
- Medical equipment
- Industrial applications: PLC, inverters, circuit breakers
- Printers, and scanners
- Alarm systems, video intercom, and HVAC
- Home audio appliances
- Mobile phone sensor hub

Table 2. STM32F411xC/xE features and peripheral counts

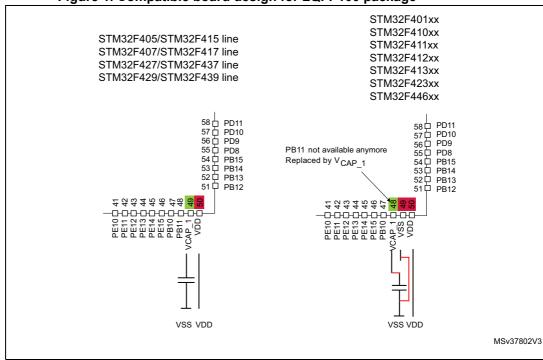
Peripherals		STM32F411xC			STM32F411xE				
Flash memory in k	256 512								
SRAM in Kbytes	System			,	128				
Timers	General- purpose		7						
	Advanced- control		1						
	SPI/ I ² S			5/5 (2 ft	ull duplex)				
	I ² C		3						
Communication interfaces	USART	3							
	SDIO	1							
	USB OTG FS	1							
GPIOs		36	50	81	36	50	81		
12-bit ADC		1							
Number of channe	els	10 16 10			16				
Maximum CPU fre	quency	100 MHz							
Operating voltage		1.7 to 3.6 V							
0	.1	Ambient temperatures: - 40 to +85 °C / - 40 to + 105 °C/ - 40 to + 125 °C							
Operating tempera	atures	Junction temperature: – 40 to + 130 °C							
Package		WLCSP49 UFQFPN48	LQFP64	UFBGA100 LQFP100	WLCSP49 UFQFPN48	LQFP64	UFBGA100 LQFP100		

2.1 Compatibility with STM32F4 Series

The STM32F411xC/xE are fully software and feature compatible with the STM32F4 series (STM32F42x, STM32F401, STM32F43x, STM32F41x, STM32F405 and STM32F407)

The STM32F411xC/xE can be used as drop-in replacement of the other STM32F4 products but some slight changes have to be done on the PCB board.

Figure 1. Compatible board design for LQFP100 package





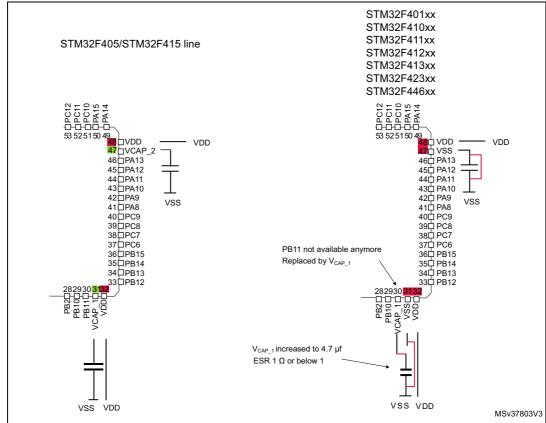


Figure 2. Compatible board design for LQFP64 package



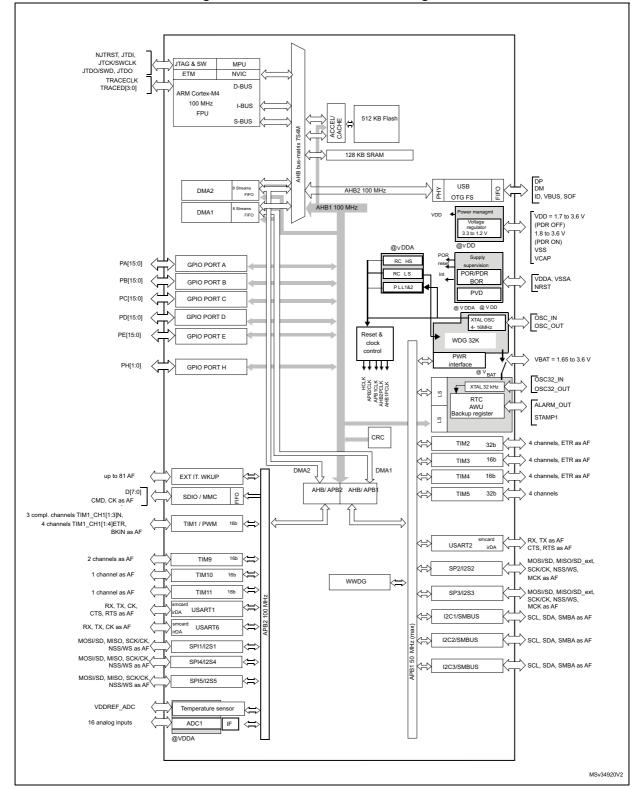


Figure 3. STM32F411xC/xE block diagram

The timers connected to APB2 are clocked from TIMxCLK up to 100 MHz, while the timers connected to APB1 are clocked from TIMxCLK up to 100 MHz.



3 Functional overview

3.1 Arm[®] Cortex[®]-M4 with FPU core with embedded Flash and SRAM

The Arm[®] Cortex[®]-M4 with FPU processor is the latest generation of Arm processors for embedded systems. It was developed to provide a low-cost platform that meets the needs of MCU implementation, with a reduced pin count and low-power consumption, while delivering outstanding computational performance and an advanced response to interrupts.

The Arm[®] Cortex[®]-M4 with FPU 32-bit RISC processor features exceptional code-efficiency, delivering the high-performance expected from an Arm core in the memory size usually associated with 8- and 16-bit devices. The processor supports a set of DSP instructions which allow efficient signal processing and complex algorithm execution. Its single precision FPU (floating point unit) speeds up software development by using metalanguage development tools, while avoiding saturation.

The STM32F411xC/xE devices are compatible with all Arm tools and software.

Figure 3 shows the general block diagram of the STM32F411xC/xE.

Note: Cortex[®]-M4 with FPU is binary compatible with Cortex[®]-M3.

3.2 Adaptive real-time memory accelerator (ART Accelerator™)

The ART Accelerator™ is a memory accelerator which is optimized for STM32 industry-standard Arm® Cortex®-M4 with FPU processors. It balances the inherent performance advantage of the Arm® Cortex®-M4 with FPU over Flash memory technologies, which normally requires the processor to wait for the Flash memory at higher frequencies.

To release the processor full 105 DMIPS performance at this frequency, the accelerator implements an instruction prefetch queue and branch cache, which increases program execution speed from the -bit Flash memory. Based on CoreMark benchmark, the performance achieved thanks to the ART accelerator is equivalent to 0 wait state program execution from Flash memory at a CPU frequency up to 100 MHz.

3.3 Batch Acquisition mode (BAM)

The Batch acquisition mode allows enhanced power efficiency during data batching. It enables data acquisition through any communication peripherals directly to memory using the DMA in reduced power consumption as well as data processing while the rest of the system is in low-power mode (including the flash and ART). For example in an audio system, a smart combination of PDM audio sample acquisition and processing from the I2S directly to RAM (flash and ARTTM stopped) with the DMA using BAM followed by some very short processing from flash allows to drastically reduce the power consumption of the application. A dedicated application note (AN4515) describes how to implement the BAM to allow the best power efficiency.

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3.4 Memory protection unit

The memory protection unit (MPU) is used to manage the CPU accesses to memory to prevent one task to accidentally corrupt the memory or resources used by any other active task. This memory area is organized into up to 8 protected areas that can in turn be divided up into 8 subareas. The protection area sizes are between 32 bytes and the whole 4 gigabytes of addressable memory.

The MPU is especially helpful for applications where some critical or certified code has to be protected against the misbehavior of other tasks. It is usually managed by an RTOS (real-time operating system). If a program accesses a memory location that is prohibited by the MPU, the RTOS can detect it and take action. In an RTOS environment, the kernel can dynamically update the MPU area setting, based on the process to be executed.

The MPU is optional and can be bypassed for applications that do not need it.

3.5 Embedded Flash memory

The devices embed up to 512 Kbytes of Flash memory available for storing programs and data.

To optimize the power consumption the Flash memory can also be switched off in Run or in Sleep mode (see *Section 3.18: Low-power modes*). Two modes are available: Flash in Stop mode or in DeepSleep mode (trade off between power saving and startup time, see *Table 34: Low-power mode wakeup timings*⁽¹⁾). Before disabling the Flash memory, the code must be executed from the internal RAM.

One-time programmable bytes

A one-time programmable area is available with 16 OTP blocks of 32 bytes. Each block can be individually locked.

(Additional information can be found in the product reference manual.)

3.6 CRC (cyclic redundancy check) calculation unit

The CRC (cyclic redundancy check) calculation unit is used to get a CRC code from a 32-bit data word and a fixed generator polynomial.

Among other applications, CRC-based techniques are used to verify data transmission or storage integrity. In the scope of the EN/IEC 60335-1 standard, they offer a means of verifying the Flash memory integrity. The CRC calculation unit helps compute a software signature during runtime, to be compared with a reference signature generated at link-time and stored at a given memory location.

3.7 Embedded SRAM

All devices embed:

 128 Kbytes of system SRAM which can be accessed (read/write) at CPU clock speed with 0 wait states

3.8 Multi-AHB bus matrix

The 32-bit multi-AHB bus matrix interconnects all the masters (CPU, DMAs) and the slaves (Flash memory, RAM, AHB and APB peripherals) and ensures a seamless and efficient operation even when several high-speed peripherals work simultaneously.

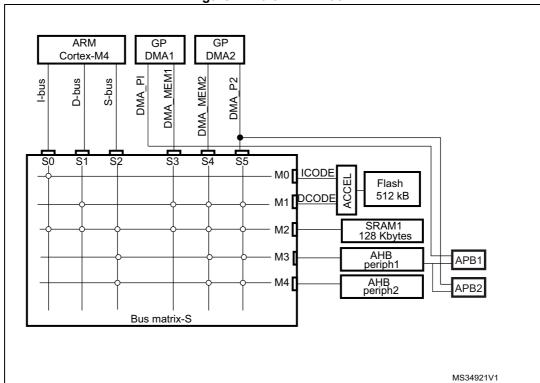


Figure 4. Multi-AHB matrix

3.9 DMA controller (DMA)

The devices feature two general-purpose dual-port DMAs (DMA1 and DMA2) with 8 streams each. They are able to manage memory-to-memory, peripheral-to-memory and memory-to-peripheral transfers. They feature dedicated FIFOs for APB/AHB peripherals, support burst transfer and are designed to provide the maximum peripheral bandwidth (AHB/APB).

The two DMA controllers support circular buffer management, so that no specific code is needed when the controller reaches the end of the buffer. The two DMA controllers also have a double buffering feature, which automates the use and switching of two memory buffers without requiring any special code.

Each stream is connected to dedicated hardware DMA requests, with support for software trigger on each stream. Configuration is made by software and transfer sizes between source and destination are independent.

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The DMA can be used with the main peripherals:

- SPI and I²S
- I²C
- USART
- General-purpose, basic and advanced-control timers TIMx
- SD/SDIO/MMC/eMMC host interface
- ADC

3.10 Nested vectored interrupt controller (NVIC)

The devices embed a nested vectored interrupt controller able to manage 16 priority levels, and handle up to 62 maskable interrupt channels plus the 16 interrupt lines of the Cortex[®]-M4 with FPU.

- Closely coupled NVIC gives low-latency interrupt processing
- Interrupt entry vector table address passed directly to the core
- Allows early processing of interrupts
- Processing of late arriving, higher-priority interrupts
- Support tail chaining
- Processor state automatically saved
- Interrupt entry restored on interrupt exit with no instruction overhead

This hardware block provides flexible interrupt management features with minimum interrupt latency.

3.11 External interrupt/event controller (EXTI)

The external interrupt/event controller consists of 21 edge-detector lines used to generate interrupt/event requests. Each line can be independently configured to select the trigger event (rising edge, falling edge, both) and can be masked independently. A pending register maintains the status of the interrupt requests. The EXTI can detect an external line with a pulse width shorter than the Internal APB2 clock period. Up to 81 GPIOs can be connected to the 16 external interrupt lines.

3.12 Clocks and startup

On reset the 16 MHz internal RC oscillator is selected as the default CPU clock. The 16 MHz internal RC oscillator is factory-trimmed to offer 1% accuracy at 25 °C. The application can then select as system clock either the RC oscillator or an external 4-26 MHz clock source. This clock can be monitored for failure. If a failure is detected, the system automatically switches back to the internal RC oscillator and a software interrupt is generated (if enabled). This clock source is input to a PLL thus allowing to increase the frequency up to 100 MHz. Similarly, full interrupt management of the PLL clock entry is available when necessary (for example if an indirectly used external oscillator fails).

Several prescalers allow the configuration of the two AHB buses, the high-speed APB (APB2) and the low-speed APB (APB1) domains. The maximum frequency of the two AHB



buses is 100 MHz while the maximum frequency of the high-speed APB domains is 100 MHz. The maximum allowed frequency of the low-speed APB domain is 50 MHz.

The devices embed a dedicated PLL (PLLI2S) which allows to achieve audio class performance. In this case, the I²S master clock can generate all standard sampling frequencies from 8 kHz to 192 kHz.

3.13 Boot modes

At startup, boot pins are used to select one out of three boot options:

- Boot from user Flash
- Boot from system memory
- Boot from embedded SRAM

The bootloader is located in system memory. It is used to reprogram the Flash memory by using USART1(PA9/10), USART2(PD5/6), USB OTG FS in device mode (PA11/12) through DFU (device firmware upgrade), I2C1(PB6/7), I2C2(PB10/3), I2C3(PA8/PB4), SPI1(PA4/5/6/7), SPI2(PB12/13/14/15) or SPI3(PA15, PC10/11/12).

For more detailed information on the bootloader, refer to Application Note: AN2606, STM32 microcontroller system memory boot mode.

3.14 Power supply schemes

- VDD = 1.7 to 3.6 V: external power supply for I/Os with the internal supervisor (POR/PDR) disabled, provided externally through VDD pins. Requires the use of an external power supply supervisor connected to the VDD and NRST pins.
- V_{SSA}, V_{DDA} = 1.7 to 3.6 V: external analog power supplies for ADC, Reset blocks, RCs and PLL. V_{DDA} and V_{SSA} must be connected to V_{DD} and V_{SS}, respectively, with decoupling technique.
- V_{BAT} = 1.65 to 3.6 V: power supply for RTC, external clock 32 kHz oscillator and backup registers (through power switch) when V_{DD} is not present.

Refer to Figure 17: Power supply scheme for more details.



3.15 Power supply supervisor

3.15.1 Internal reset ON

This feature is available for V_{DD} operating voltage range 1.8 V to 3.6 V.

The internal power supply supervisor is enabled by holding PDR_ON high.

The devices have an integrated power-on reset (POR) / power-down reset (PDR) circuitry coupled with a Brownout reset (BOR) circuitry. At power-on, POR is always active, and ensures proper operation starting from 1.8 V. After the 1.8 V POR threshold level is reached, the option byte loading process starts, either to confirm or modify default thresholds, or to disable BOR permanently. Three BOR thresholds are available through option bytes.

The devices remain in reset mode when V_{DD} is below a specified threshold, $V_{POR/PDR}$ or V_{BOR} , without the need for an external reset circuit.

The devices also feature an embedded programmable voltage detector (PVD) that monitors the V_{DD}/V_{DDA} power supply and compares it to the V_{PVD} threshold. An interrupt can be generated when V_{DD}/V_{DDA} drops below the V_{PVD} threshold and/or when V_{DD}/V_{DDA} is higher than the V_{PVD} threshold. The interrupt service routine can then generate a warning message and/or put the MCU into a safe state. The PVD is enabled by software.

3.15.2 Internal reset OFF

This feature is available only on packages featuring the PDR_ON pin. The internal power-on reset (POR) / power-down reset (PDR) circuitry is disabled by setting the PDR_ON pin to low.

An external power supply supervisor should monitor V_{DD} and should set the device in reset mode when V_{DD} is below 1.7 V. NRST should be connected to this external power supply supervisor. Refer to *Figure 5: Power supply supervisor interconnection with internal reset OFF*.

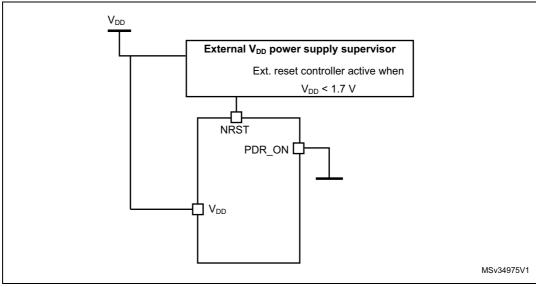


Figure 5. Power supply supervisor interconnection with internal reset OFF⁽¹⁾

1. The PRD_ON pin is only available on the WLCSP49 and UFBGA100 packages.

A comprehensive set of power-saving mode allows to design low-power applications.

When the internal reset is OFF, the following integrated features are no longer supported:

- The integrated power-on reset (POR) / power-down reset (PDR) circuitry is disabled.
- The brownout reset (BOR) circuitry must be disabled.
- The embedded programmable voltage detector (PVD) is disabled.
- V_{BAT} functionality is no more available and VBAT pin should be connected to V_{DD}.

3.16 Voltage regulator

The regulator has four operating modes:

- Regulator ON
 - Main regulator mode (MR)
 - Low power regulator (LPR)
 - Power-down
- Regulator OFF

3.16.1 Regulator ON

On packages embedding the BYPASS_REG pin, the regulator is enabled by holding BYPASS_REG low. On all other packages, the regulator is always enabled.

There are three power modes configured by software when the regulator is ON:

- MR is used in the nominal regulation mode (With different voltage scaling in Run)
 In Main regulator mode (MR mode), different voltage scaling are provided to reach the best compromise between maximum frequency and dynamic power consumption.
- LPR is used in the Stop modes
 - The LP regulator mode is configured by software when entering Stop mode.
- Power-down is used in Standby mode.
 - The Power-down mode is activated only when entering in Standby mode. The regulator output is in high impedance and the kernel circuitry is powered down, inducing zero consumption. The contents of the registers and SRAM are lost.

Depending on the package, one or two external ceramic capacitors should be connected on the V_{CAP_1} and V_{CAP_2} pins. The V_{CAP_2} pin is only available for the LQFP100 and UFBGA100 packages.

All packages have the regulator ON feature.

3.16.2 Regulator OFF

The Regulator OFF is available only on the UFBGA100, which features the BYPASS_REG pin. The regulator is disabled by holding BYPASS_REG high. The regulator OFF mode allows to supply externally a V12 voltage source through V_{CAP-1} and V_{CAP-2} pins.

Since the internal voltage scaling is not managed internally, the external voltage value must be aligned with the targeted maximum frequency. Refer to *Table 14: General operating conditions*.

The two 2.2 μ F V_{CAP} ceramic capacitors should be replaced by two 100 nF decoupling capacitors. Refer to *Figure 17: Power supply scheme*.

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When the regulator is OFF, there is no more internal monitoring on V12. An external power supply supervisor should be used to monitor the V12 of the logic power domain. PA0 pin should be used for this purpose, and act as power-on reset on V12 power domain.

In regulator OFF mode, the following features are no more supported:

- PA0 cannot be used as a GPIO pin since it allows to reset a part of the V12 logic power domain which is not reset by the NRST pin.
- As long as PA0 is kept low, the debug mode cannot be used under power-on reset. As a consequence, PA0 and NRST pins must be managed separately if the debug connection under reset or pre-reset is required.

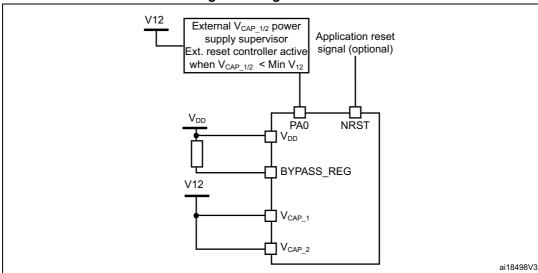


Figure 6. Regulator OFF

The following conditions must be respected:

- V_{DD} should always be higher than V_{CAP_1} and V_{CAP_2} to avoid current injection between power domains.
- If the time for V_{CAP_1} and V_{CAP_2} to reach V₁₂ minimum value is faster than the time for V_{DD} to reach 1.7 V, then PA0 should be kept low to cover both conditions: until V_{CAP_1} and V_{CAP_2} reach V₁₂ minimum value and until V_{DD} reaches 1.7 V (see *Figure* 7).
- Otherwise, if the time for V_{CAP_1} and V_{CAP_2} to reach V_{12} minimum value is slower than the time for V_{DD} to reach $\overline{1}.7$ V, then $\overline{P}A0$ could be asserted low externally (see *Figure 8*).
- If V_{CAP_1} and V_{CAP_2} go below V₁₂ minimum value and V_{DD} is higher than 1.7 V, then a
 reset must be asserted on PA0 pin.

Note: The minimum value of V_{12} depends on the maximum frequency targeted in the application



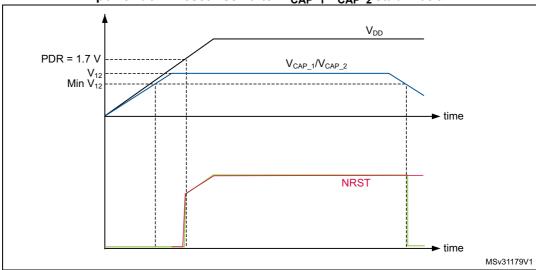


Figure 7. Startup in regulator OFF: slow V_{DD} slope - power-down reset risen after V_{CAP-1}/V_{CAP-2} stabilization

1. This figure is valid whatever the internal reset mode (ON or OFF).

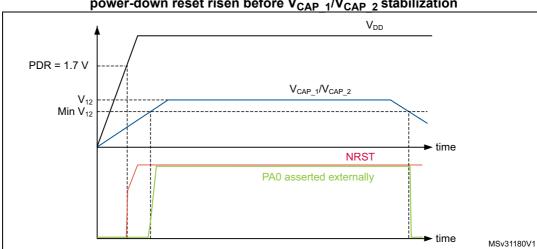


Figure 8. Startup in regulator OFF mode: fast V_{DD} slope -power-down reset risen before V_{CAP-1}/V_{CAP-2} stabilization

1. This figure is valid whatever the internal reset mode (ON or OFF).

Yes

PDR ON external

control (1)

Yes

PDR ON set to VDD

3.16.3 Regulator ON/OFF and internal power supply supervisor availability

Package	Regulator ON	Regulator OFF	Power supply supervisor ON	Power supply supervisor OFF	
UFQFPN48	Yes	No	Yes	No	
WLCSP49	Yes	No	Yes PDR_ON set to VDD	Yes PDR_ON external control ⁽¹⁾	
LQFP64	Yes	No	Yes	No	
LQFP100	Yes	No	Yes	No	

Yes

BYPASS REG set to

VDD

Table 3. Regulator ON/OFF and internal power supply supervisor availability

UFBGA100

3.17 Real-time clock (RTC) and backup registers

The backup domain includes:

Yes

BYPASS REG set to

VSS

- The real-time clock (RTC)
- 20 backup registers

The real-time clock (RTC) is an independent BCD timer/counter. Dedicated registers contain the second, minute, hour (in 12/24 hour), week day, date, month, year, in BCD (binary-coded decimal) format. Correction for 28, 29 (leap year), 30, and 31 day of the month are performed automatically. The RTC features a reference clock detection, a more precise second source clock (50 or 60 Hz) can be used to enhance the calendar precision. The RTC provides a programmable alarm and programmable periodic interrupts with wakeup from Stop and Standby modes. The sub-seconds value is also available in binary format.

It is clocked by a 32.768 kHz external crystal, resonator or oscillator, the internal low-power RC oscillator or the high-speed external clock divided by 128. The internal low-speed RC has a typical frequency of 32 kHz. The RTC can be calibrated using an external 512 Hz output to compensate for any natural quartz deviation.

Two alarm registers are used to generate an alarm at a specific time and calendar fields can be independently masked for alarm comparison. To generate a periodic interrupt, a 16-bit programmable binary auto-reload downcounter with programmable resolution is available and allows automatic wakeup and periodic alarms from every 120 µs to every 36 hours.

A 20-bit prescaler is used for the time base clock. It is by default configured to generate a time base of 1 second from a clock at 32.768 kHz.

The backup registers are 32-bit registers used to store 80 bytes of user application data when V_{DD} power is not present. Backup registers are not reset by a system, a power reset, or when the device wakes up from the Standby mode (see Section 3.18: Low-power modes).

Additional 32-bit registers contain the programmable alarm subseconds, seconds, minutes, hours, day, and date.



^{1.} Refer to Section 3.15: Power supply supervisor

The RTC and backup registers are supplied through a switch that is powered either from the V_{DD} supply when present or from the V_{BAT} pin.

3.18 Low-power modes

The devices support three low-power modes to achieve the best compromise between low power consumption, short startup time and available wakeup sources:

Sleep mode

In Sleep mode, only the CPU is stopped. All peripherals continue to operate and can wake up the CPU when an interrupt/event occurs.

To further reduce the power consumption, the Flash memory can be switched off before entering in Sleep mode. Note that this requires a code execution from the RAM.

Stop mode

The Stop mode achieves the lowest power consumption while retaining the contents of SRAM and registers. All clocks in the 1.2 V domain are stopped, the PLL, the HSI RC and the HSE crystal oscillators are disabled. The voltage regulator can also be put either in normal or in low-power mode.

The devices can be woken up from the Stop mode by any of the EXTI line (the EXTI line source can be one of the 16 external lines, the PVD output, the RTC alarm/ wakeup/ tamper/ time stamp events).

Standby mode

The Standby mode is used to achieve the lowest power consumption. The internal voltage regulator is switched off so that the entire 1.2 V domain is powered off. The PLL, the HSI RC and the HSE crystal oscillators are also switched off. After entering Standby mode, the SRAM and register contents are lost except for registers in the backup domain when selected.

The devices exit the Standby mode when an external reset (NRST pin), an IWDG reset, a rising edge on the WKUP pin, or an RTC alarm/ wakeup/ tamper/time stamp event occurs.

Standby mode is not supported when the embedded voltage regulator is bypassed and the 1.2 V domain is controlled by an external power.

3.19 V_{BAT} operation

The VBAT pin allows to power the device V_{BAT} domain from an external battery, an external super-capacitor, or from V_{DD} when no external battery and an external super-capacitor are present.

V_{BAT} operation is activated when V_{DD} is not present.

The VBAT pin supplies the RTC and the backup registers.

Note:

When the microcontroller is supplied from VBAT, external interrupts and RTC alarm/events do not exit it from V_{BAT} operation. When PDR_ON pin is not connected to V_{DD} (internal Reset OFF), the V_{BAT} functionality is no more available and VBAT pin should be connected to V_{DD} .



3.20 Timers and watchdogs

The devices embed one advanced-control timer, seven general-purpose timers and two watchdog timers.

All timer counters can be frozen in debug mode.

Table 4 compares the features of the advanced-control and general-purpose timers.

Table 4. Timer feature comparison

Table 4. Timer reacure comparison									
Timer type	Timer	Counter resolution	Counter type	Prescaler factor	DMA request generation	Capture/ compare channels	Complemen- tary output	Max. interface clock (MHz)	Max. timer clock (MHz)
Advanced -control	TIM1	16-bit	Up, Down, Up/down	Any integer between 1 and 65536	Yes	4	Yes	100	100
General purpose	TIM2, TIM5	32-bit	Up, Down, Up/down	Any integer between 1 and 65536	Yes	4	No	50	100
	TIM3, TIM4	16-bit	Up, Down, Up/down	Any integer between 1 and 65536	Yes	4	No	50	100
	TIM9	16-bit	Up	Any integer between 1 and 65536	No	2	No	100	100
	TIM10, TIM11	16-bit	Up	Any integer between 1 and 65536	No	1	No	100	100

3.20.1 Advanced-control timers (TIM1)

The advanced-control timer (TIM1) can be seen as three-phase PWM generators multiplexed on 4 independent channels. It has complementary PWM outputs with programmable inserted dead times. It can also be considered as a complete general-purpose timer. Its 4 independent channels can be used for:

- Input capture
- Output compare
- PWM generation (edge- or center-aligned modes)
- One-pulse mode output



If configured as standard 16-bit timers, it has the same features as the general-purpose TIMx timers. If configured as a 16-bit PWM generator, it has full modulation capability (0-100%).

The advanced-control timer can work together with the TIMx timers via the Timer Link feature for synchronization or event chaining.

TIM1 supports independent DMA request generation.

3.20.2 General-purpose timers (TIMx)

There are seven synchronizable general-purpose timers embedded in the STM32F411xC/xE (see *Table 4* for differences).

TIM2, TIM3, TIM4, TIM5

The STM32F411xC/xE devices are 4 full-featured general-purpose timers: TIM2, TIM5, TIM3, and TIM4. The TIM2 and TIM5 timers are based on a 32-bit auto-reload up/downcounter and a 16-bit prescaler. The TIM3 and TIM4 timers are based on a 16-bit auto-reload up/downcounter and a 16-bit prescaler. They all feature four independent channels for input capture/output compare, PWM or one-pulse mode output. This gives up to 15 input capture/output compare/PWMs.

The TIM2, TIM3, TIM4, TIM5 general-purpose timers can work together, or with the other general-purpose timers and the advanced-control timer TIM1 via the Timer Link feature for synchronization or event chaining.

Any of these general-purpose timers can be used to generate PWM outputs. TIM2, TIM3, TIM4, TIM5 all have independent DMA request generation. They are capable of handling quadrature (incremental) encoder signals and the digital outputs from 1 to 4 hall-effect sensors.

TIM9, TIM10 and TIM11

These timers are based on a 16-bit auto-reload upcounter and a 16-bit prescaler. TIM10 and TIM11 feature one independent channel, whereas TIM9 has two independent channels for input capture/output compare, PWM or one-pulse mode output. They can be synchronized with the TIM2, TIM3, TIM4, TIM5 full-featured general-purpose timers. They can also be used as simple time bases.

3.20.3 Independent watchdog

The independent watchdog is based on a 12-bit downcounter and 8-bit prescaler. It is clocked from an independent 32 kHz internal RC and as it operates independently from the main clock, it can operate in Stop and Standby modes. It can be used either as a watchdog to reset the device when a problem occurs, or as a free-running timer for application timeout management. It is hardware- or software-configurable through the option bytes.

3.20.4 Window watchdog

The window watchdog is based on a 7-bit downcounter that can be set as free-running. It can be used as a watchdog to reset the device when a problem occurs. It is clocked from the main clock. It has an early warning interrupt capability and the counter can be frozen in debug mode.

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3.20.5 SysTick timer

This timer is dedicated to real-time operating systems, but could also be used as a standard downcounter. It features:

- A 24-bit downcounter
- Autoreload capability
- Maskable system interrupt generation when the counter reaches 0
- Programmable clock source.

3.21 Inter-integrated circuit interface (I²C)

Up to three I²C bus interfaces can operate in multimaster and slave modes. They can support the standard (up to 100 kHz) and fast (up to 400 kHz) modes. The I2C bus frequency can be increased up to 1 MHz. For more details about the complete solution, please contact your local ST sales representative. They also support the 7/10-bit addressing mode and the 7-bit dual addressing mode (as slave). A hardware CRC generation/verification is embedded.

They can be served by DMA and they support SMBus 2.0/PMBus.

The devices also include programmable analog and digital noise filters (see Table 5).

Table 5. Comparison of I2C analog and digital filters

	Analog filter	Digital filter
Pulse width of suppressed spikes	≥ 50 ns	Programmable length from 1 to 15 I2C peripheral clocks

3.22 Universal synchronous/asynchronous receiver transmitters (USART)

The devices embed three universal synchronous/asynchronous receiver transmitters (USART1, USART2 and USART6).

These three interfaces provide asynchronous communication, IrDA SIR ENDEC support, multiprocessor communication mode, single-wire half-duplex communication mode and have LIN Master/Slave capability. The USART1 and USART6 interfaces are able to communicate at speeds of up to 12.5 Mbit/s. The USART2 interface communicates at up to 6.25 bit/s.

USART1 and USART2 also provide hardware management of the CTS and RTS signals, Smart Card mode (ISO 7816 compliant) and SPI-like communication capability. All interfaces can be served by the DMA controller.

USART name	Standard features	Modem (RTS/CTS)	LIN	SPI master	irDA	Smartcard (ISO 7816)	Max. baud rate in Mbit/s (oversampling by 16)	Max. baud rate in Mbit/s (oversampling by 8)	APB mapping
USART1	х	Х	Х	Х	Х	Х	6.25	12.5	APB2 (max. 100 MHz)
USART2	Х	Х	Х	Х	X	Х	3.12	6.25	APB1 (max. 50 MHz)
USART6	Х	N.A	Х	Х	X	Х	6.25	12.5	APB2 (max. 100 MHz)

Table 6. USART feature comparison

3.23 Serial peripheral interface (SPI)

The devices feature five SPIs in slave and master modes in full-duplex and simplex communication modes. SPI1, SPI4 and SPI5 can communicate at up to 50 Mbit/s, SPI2 and SPI3 can communicate at up to 25 Mbit/s. The 3-bit prescaler gives 8 master mode frequencies and the frame is configurable to 8 bits or 16 bits. The hardware CRC generation/verification supports basic SD Card/MMC modes. All SPIs can be served by the DMA controller.

The SPI interface can be configured to operate in TI mode for communications in master mode and slave mode.

3.24 Inter-integrated sound (I²S)

Five standard I²S interfaces (multiplexed with SPI1 to SPI5) are available. They can be operated in master or slave mode, in simplex communication modes and full duplex for I2S2 and I2S3 and can be configured to operate with a 16-/32-bit resolution as an input or output channel. All the I2Sx audio sampling frequencies from 8 kHz up to 192 kHz are supported. When either or both of the I²S interfaces is/are configured in master mode, the master clock can be output to the external DAC/CODEC at 256 times the sampling frequency.

All I²Sx can be served by the DMA controller.

3.25 Audio PLL (PLLI2S)

The devices feature an additional dedicated PLL for audio I²S application. It allows to achieve error-free I²S sampling clock accuracy without compromising on the CPU performance.

The PLLI2S configuration can be modified to manage an I^2S sample rate change without disabling the main PLL (PLL) used for the CPU.

The audio PLL can be programmed with very low error to obtain sampling rates ranging from 8 kHz to 192 kHz.

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In addition to the audio PLL, a master clock input pin can be used to synchronize the I2S flow with an external PLL (or Codec output).

3.26 Secure digital input/output interface (SDIO)

An SD/SDIO/MMC/eMMC host interface is available, that supports MultiMediaCard System Specification Version 4.2 in three different databus modes: 1-bit (default), 4-bit and 8-bit.

The interface allows data transfer at up to 50 MHz, and is compliant with the SD Memory Card Specification Version 2.0.

The SDIO Card Specification Version 2.0 is also supported with two different databus modes: 1-bit (default) and 4-bit.

The current version supports only one SD/SDIO/MMC4.2 card at any one time and a stack of MMC4.1 or previous.

In addition to SD/SDIO/MMC/eMMC, this interface is fully compliant with the CE-ATA digital protocol Rev1.1.

3.27 Universal serial bus on-the-go full-speed (OTG FS)

The devices embed an USB OTG full-speed device/host/OTG peripheral with integrated transceivers. The USB OTG FS peripheral is compliant with the USB 2.0 specification and with the OTG 1.0 specification. It has software-configurable endpoint setting and supports suspend/resume. The USB OTG full-speed controller requires a dedicated 48 MHz clock that is generated by a PLL connected to the HSE oscillator. The major features are:

- Combined Rx and Tx FIFO size of 320 × 35 bits with dynamic FIFO sizing
- Supports the session request protocol (SRP) and host negotiation protocol (HNP)
- 4 bidirectional endpoints
- 8 host channels with periodic OUT support
- HNP/SNP/IP inside (no need for any external resistor)
- For OTG/Host modes, a power switch is needed in case bus-powered devices are connected

3.28 General-purpose input/outputs (GPIOs)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain, with or without pull-up or pull-down), as input (floating, with or without pull-up or pull-down) or as peripheral alternate function. Most of the GPIO pins are shared with digital or analog alternate functions. All GPIOs are high-current-capable and have speed selection to better manage internal noise, power consumption and electromagnetic emission.

The I/O configuration can be locked if needed by following a specific sequence in order to avoid spurious writing to the I/Os registers.

Fast I/O handling allowing maximum I/O toggling up to 100 MHz.

3.29 Analog-to-digital converter (ADC)

One 12-bit analog-to-digital converter is embedded and shares up to 16 external channels, performing conversions in the single-shot or scan mode. In scan mode, automatic conversion is performed on a selected group of analog inputs.

The ADC can be served by the DMA controller. An analog watchdog feature allows very precise monitoring of the converted voltage of one, some or all selected channels. An interrupt is generated when the converted voltage is outside the programmed thresholds.

To synchronize A/D conversion and timers, the ADCs could be triggered by any of TIM1, TIM2. TIM3. TIM4 or TIM5 timer.

3.30 Temperature sensor

The temperature sensor has to generate a voltage that varies linearly with temperature. The conversion range is between 1.7 V and 3.6 V. The temperature sensor is internally connected to the ADC_IN18 input channel which is used to convert the sensor output voltage into a digital value. Refer to the reference manual for additional information.

As the offset of the temperature sensor varies from chip to chip due to process variation, the internal temperature sensor is mainly suitable for applications that detect temperature changes instead of absolute temperatures. If an accurate temperature reading is needed, then an external temperature sensor part should be used.

3.31 Serial wire JTAG debug port (SWJ-DP)

The Arm SWJ-DP interface is embedded, and is a combined JTAG and serial wire debug port that enables either a serial wire debug or a JTAG probe to be connected to the target.

Debug is performed using 2 pins only instead of 5 required by the JTAG (JTAG pins could be re-use as GPIO with alternate function): the JTAG TMS and TCK pins are shared with SWDIO and SWCLK, respectively, and a specific sequence on the TMS pin is used to switch between JTAG-DP and SW-DP.

3.32 Embedded Trace Macrocell™

The Arm Embedded Trace Macrocell provides a greater visibility of the instruction and data flow inside the CPU core by streaming compressed data at a very high rate from the STM32F411xC/xE through a small number of ETM pins to an external hardware trace port analyzer (TPA) device. The TPA is connected to a host computer using any high-speed channel available. Real-time instruction and data flow activity can be recorded and then formatted for display on the host computer that runs the debugger software. TPA hardware is commercially available from common development tool vendors.

The Embedded Trace Macrocell operates with third party debugger software tools.

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4 Pinouts and pin description

Figure 9. STM32F411xC/xE WLCSP49 pinout

A (VDD) (VSS) B (BAT) (PDR) C (C14) (PC15)	PB8 PB9	(PB4) (PB5) (PB6)	(PB3) (PA13) (PA12)	(VDD) (PA10)	(PA14) (VSS) (PA11)
C osc32 N osc32 ou	O D D D D D D D D D D D D D D D D D D D				
05C32/IN 05C32 ØU	T (PB9)	PB6	PA12	(PA10)	PA11
- 600 600					_
D (PHO) (PH1) OSC_IN OSC_OU	r PC13	(PB7)	(VSS)	PA9	PA8
E (IRS) (SSA)	PA2	PA3	(PB10)	(PB12)	PB15
F (DDA) (PA0)	PA5	PA6	(PA7)	VDD	PB14
G PA1 PA4	(PB0)	(PB1)	(PB2)	VCAP1	PB13

1. The above figure shows the package bump side.

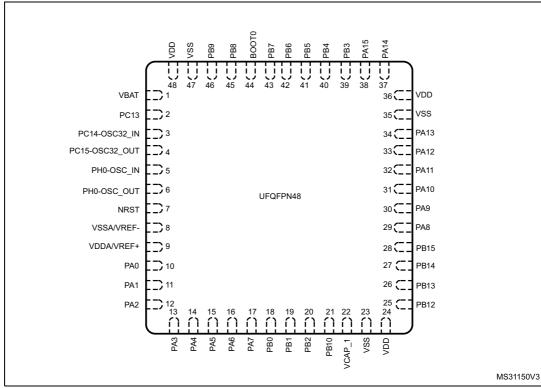


Figure 10. STM32F411xC/xE UFQFPN48 pinout

1. The above figure shows the package top view.



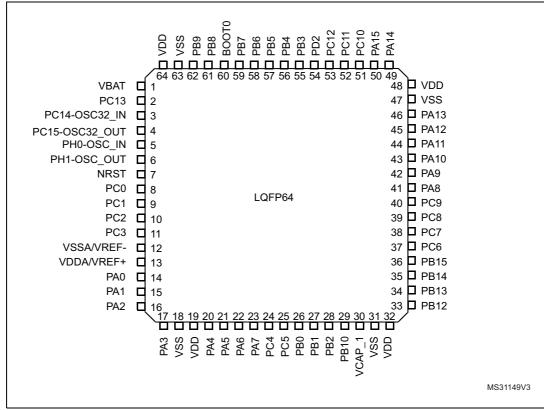


Figure 11. STM32F411xC/xE LQFP64 pinout

1. The above figure shows the package top view.

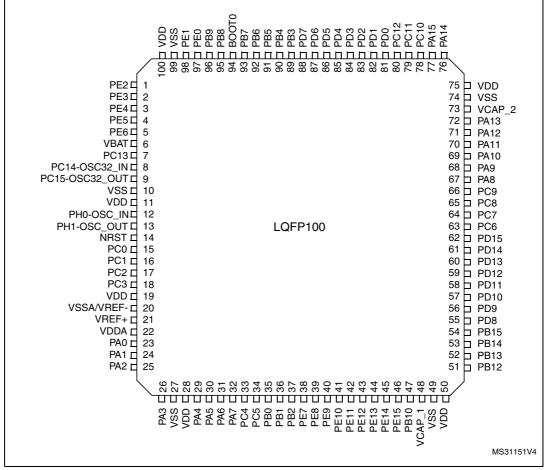


Figure 12. STM32F411xC/xE LQFP100 pinout

1. The above figure shows the package top view.



Figure 13. STM32F411xC/xE UFBGA100 pinout

	1	2	3	4	5	6	7	8	9	10	11	12
A	PE3	(PE1)	PB8	(BOO)10	(PD7)	PD5	 PB4	(PB3)	PA15)	PA14	PA13	PA12
В	PE4	(PE2)	(PB9)	(PB7)	PB6	PD6	PD4	PD3	(PD1)	C12	PC10	PA11
С	C13 ANTI_TA	AMP PE5	(PE0)	VDD	(PB5)		 	PD2	(PD0)	C11	VCAP 2	PA10
D	FC14 OSC32_II	N (PE6)	VSS				 			PA9	PA8	PC9
E	PC15 OSC32_	OUT	BYPAS	SS_REG			 			PC8	PC7	PC6
F	PHO N	VSS									VSS	VSS
G _	PH1 OSC O	UT (VDD)					† — – 				(VDD)	(VDD)
Н	PCO	NRS	PDR	DN			 			(D15)	(D14)	PD13
J	VSSA	PC1	PC2							(D12)	(D11)	PD10
К	VREIT-	PC3	(PA2)	PA5	PC4		 	PD9	(PB11)	PB15	(B14)	(PB13)
L	VREP+	PA0 WKUP	(PA3)	PA6	PC5	(PB2)	PE8	PE10	PE12	PB10	(CAP)	PB12
М	(VDDA)	PA1	PA4	PA7	(PB0)	(PB1)	PE7	PE9	(PE11)	PE13	PE14	PE15

1. This figure shows the package top view

Table 7. Legend/abbreviations used in the pinout table

Name	Abbreviation	Definition						
Pin name		specified in brackets below the pin name, the pin function during and after as the actual pin name						
	S	Supply pin						
Pin type	I	Input only pin						
	I/O	Input/ output pin						
	FT	5 V tolerant I/O						
I/O structure	TC Standard 3.3 V I/O							
i/O structure	B Dedicated BOOT0 pin							
	NRST Bidirectional reset pin with embedded weak pull-up resist							
Notes	Unless otherwise	specified by a note, all I/Os are set as floating inputs during and after reset						
Alternate functions	Functions selected through GPIOx_AFR registers							
Additional functions	Functions directly	selected/enabled through peripheral registers						

Table 8. STM32F411xC/xE pin definitions

	Pir	numl	oer							
UFQFPN48	LQFP64	WLCSP49	LQFP100	UFBGA100	Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
-	-	-	1	B2	PE2	I/O	FT	-	TRACECLK, SPI4_SCK/I2S4_CK, SPI5_SCK/I2S5_CK, EVENTOUT	-
-	-	-	2	A1	PE3	I/O	FT	-	TRACEDO, EVENTOUT	-
-	-	-	3	B1	PE4	I/O	FT	-	TRACED1, SPI4_NSS/I2S4_WS, SPI5_NSS/I2S5_WS, EVENTOUT	-
-	-	-	4	C2	PE5	I/O	FT	-	TRACED2, TIM9_CH1, SPI4_MISO, SPI5_MISO, EVENTOUT	-

Table 8. STM32F411xC/xE pin definitions (continued)

	Pir	numk							intions (continued)	
UFQFPN48	LQFP64	WLCSP49	LQFP100	UFBGA100	Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
-	-	-	5	D2	PE6	I/O	FT	-	TRACED3, TIM9_CH2, SPI4_MOSI/I2S4_SD, SPI5_MOSI/I2S5_SD, EVENTOUT	-
-	ı	İ	-	D3	VSS	S	-	ı	-	-
-	-	ı	-	C4	VDD	S	-	ı	-	-
1	1	B7	6	E2	VBAT	S	1	-	-	-
2	2	D5	7	C1	PC13- ANTI_TAMP	I/O	FT	(2)(3)	-	RTC_AMP1, RTC_OUT, RTC_TS
3	3	C7	8	D1	PC14- OSC32_IN	I/O	FT	(2)(3) (4)	-	OSC32_IN
4	4	C6	9	E1	PC15- OSC32_OUT	I/O	FT	-	-	OSC32_OUT
-	-	-	10	F2	VSS	S	-	-	-	-
-	-	-	11	G2	VDD	S	-	-	-	-
5	5	D7	12	F1	PH0 - OSC_IN	I/O	FT	-	-	OSC_IN
6	6	D6	13	G1	PH1 - OSC_OUT	I/O	FT	-	-	OSC_OUT
7	7	E7	14	H2	NRST	I/O	FT	-	EVENTOUT	-
-	8	-	15	H1	PC0	I/O	FT	-	EVENTOUT	ADC1_10
-	9	-	16	J2	PC1	I/O	FT	-	EVENTOUT	ADC1_11
-	10	·	17	J3	PC2	I/O	FT	ſ	SPI2_MISO, I2S2ext_SD, EVENTOUT	ADC1_12
-	11	-	18	K2	PC3	I/O	FT	-	SPI2_MOSI/I2S2_SD, EVENTOUT	ADC1_13
-	-	-	19	-	VDD	S	-	-	-	-
8	12	E6	20	J1	VSSA	S	-	-	-	-
-	-	-	-	K1	VREF-	S	-	-	-	-
9	13	F7	21	L1	VREF+	S	-	1	-	-
-	-	-	22	M1	VDDA	S	-	-	-	-



Table 8. STM32F411xC/xE pin definitions (continued)

	Pir	n numb							inions (continued)	
UFQFPN48	LQFP64	WLCSP49	LQFP100	UFBGA100	Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
10	14	F6	23	L2	PA0-WKUP	I/O	тс	(5)	TIM2_CH1/TIM2_ET, TIM5_CH1, USART2_CTS, EVENTOUT	ADC1_0, WKUP1
11	15	G7	24	M2	PA1	I/O FT		-	TIM2_CH2, TIM5_CH2, SPI4_MOSI/I2S4_SD, USART2_RTS, EVENTOUT	ADC1_1
12	16	E5	25	К3	PA2	I/O	FT	-	TIM2_CH3, TIM5_CH3, TIM9_CH1, I2S2_CKIN, USART2_TX, EVENTOUT	ADC1_2
13	17	E4	26	L3	PA3	I/O	FT	-	TIM2_CH4, TIM5_CH4, TIM9_CH2, I2S2_MCK, USART2_RX, EVENTOUT	ADC1_3
-	18	-	27	-	VSS	S	-	-	-	-
-	-	-	-	E3	BYPASS_REG	S	-	-	-	-
-	19	-	28	-	VDD	I	FT	-	EVENTOUT	-
14	20	G6	29	МЗ	PA4	I/O	FT	-	SPI1_NSS/I2S1_WS, SPI3_NSS/I2S3_WS, USART2_CK, EVENTOUT	ADC1_4
15	21	F5	30	K4	PA5	I/O	FT	-	TIM2_CH1/TIM2_ET, SPI1_SCK/I2S1_CK, EVENTOUT	ADC1_5
16	22	F4	31	L4	PA6	I/O	FT	-	TIM1_BKIN, TIM3_CH1, SPI1_MISO, I2S2_MCK, SDIO_CMD, EVENTOUT	ADC1_6
17	23	F3	32	M4	I PA7 I/O F1		FT	-	TIM1_CH1N, TIM3_CH2, SPI1_MOSI/I2S1_SD, EVENTOUT	ADC1_7

Table 8. STM32F411xC/xE pin definitions (continued)

	Pir	n numb	oer							
UFQFPN48	LQFP64	WLCSP49	LQFP100	UFBGA100	Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
-	24	-	33	K5	PC4	I/O	FT	ı	EVENTOUT	ADC1_14
-	25	-	34	L5	PC5	I/O	FT	ı	EVENTOUT	ADC1_15
18	26	G5	35	M5	PB0	I/O	FT	-	TIM1_CH2N, TIM3_CH3, SPI5_SCK/I2S5_CK, EVENTOUT	ADC1_8
19	27	G4	36	M6	PB1	I/O	FT	ı	TIM1_CH3N, TIM3_CH4, SPI5_NSS/I2S5_WS, EVENTOUT	ADC1_9
20	28	G3	37	L6	PB2	I/O	FT	ı	EVENTOUT	BOOT1
-	-	-	38	M7	PE7	I/O	FT	-	TIM1_ETR, EVENTOUT	-
-	-	-	39	L7	PE8	I/O	FT	-	TIM1_CH1N, EVENTOUT	-
-	-	-	40	M8	PE9	I/O	FT	-	TIM1_CH1, EVENTOUT	-
-	-	-	41	L8	PE10	I/O	FT	-	TIM1_CH2N, EVENTOUT	-
-	-	-	42	M9	PE11	I/O	FT	-	TIM1_CH2, SPI4_NSS/I2S4_WS, SPI5_NSS/I2S5_WS, EVENTOUT	-
-	-	-	43	L9	PE12	I/O	FT	-	TIM1_CH3N, SPI4_SCK/I2S4_CK, SPI5_SCK/I2S5_CK, EVENTOUT	-
-	-	-	44	M10	PE13	I/O	FT	-	TIM1_CH3, SPI4_MISO, SPI5_MISO, EVENTOUT	-
-	-	-	45	M11	PE14	I/O	FT	-	TIM1_CH4, SPI4_MOSI/I2S4_SD, SPI5_MOSI/I2S5_SD, EVENTOUT	-
-	-	-	46	M12	PE15	I/O	FT	-	TIM1_BKIN, EVENTOUT	-



Table 8. STM32F411xC/xE pin definitions (continued)

	Pir	n numb							inions (continued)	
UFQFPN48	LQFP64	WLCSP49	LQFP100	UFBGA100	Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
21	29	E3	47	L10	PB10	I/O	FT	-	TIM2_CH3, I2C2_SCL, SPI2_SCK/I2S2_CK, I2S3_MCK, SDIO_D7, EVENTOUT	-
-	-	1	-	K9	PB11		-			
22	30	G2	48	L11	VCAP_1	EVENTOUT		-	-	
23	31	D3	49	F12	VSS	S	-	-	-	-
24	32	F2	50	G12	VDD	S	-	-	-	-
25	33	E2	51	L12	PB12	I/O	FT	-	TIM1_BKIN, I2C2_SMBA, SPI2_NSS/I2S2_WS, SPI4_NSS/I2S4_WS, SPI3_SCK/I2S3_CK, EVENTOUT	-
26	34	G1	52	K12	PB13	I/O	FT	-	TIM1_CH1N, SPI2_SCK/I2S2_CK, SPI4_SCK/I2S4_CK, EVENTOUT	-
27	35	F1	53	K11	PB14	I/O	FT	-	TIM1_CH2N, SPI2_MISO, I2S2ext_SD, SDIO_D6, EVENTOUT	-
28	36	E1	54	K10	PB15	RTC_50Hz, TIM1_CH3N,		TIM1_CH3N, SPI2_MOSI/I2S2_SD, SDIO_CK,	RTC_REFIN	
-	-	-	55	-	PD8	I/O	FT	-	-	-
-	-	-	56	K8	PD9	I/O	FT	-	-	-
-	-	-	57	J12	PD10	I/O	FT	-	-	-
-	-	-	58	J11	PD11	I/O	FT	-	-	-
-	-	-	59	J10	PD12	I/O ET TIN		TIM4_CH1, EVENTOUT	-	

Table 8. STM32F411xC/xE pin definitions (continued)

	Pir	n numl	oer							
UFQFPN48	LQFP64	WLCSP49	LQFP100	UFBGA100	Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
-	-	-	60	H12	PD13	I/O	FT	-	TIM4_CH2, EVENTOUT	-
-	-	-	61	H11	PD14	I/O	FT	-	TIM4_CH3, EVENTOUT	-
-	-	-	62	H10	PD15	1/0 F1 - E\		-	TIM4_CH4, EVENTOUT	-
-	37	-	63	E12	PC6			-	TIM3_CH1, I2S2_MCK, USART6_TX, SDIO_D6, EVENTOUT	-
-	38	-	64	E11	PC7	I/O	FT	-	TIM3_CH2, SPI2_SCK/I2S2_CK, I2S3_MCK, USART6_RX, SDIO_D7, EVENTOUT	-
-	39	-	65	E10	PC8	I/O	FT	-	TIM3_CH3, USART6_CK, SDIO_D0, EVENTOUT	-
-	40	-	66	D12	PC9	I/O	FT	-	MCO_2, TIM3_CH4, I2C3_SDA, I2S2_CKIN, SDIO_D1, EVENTOUT	-
29	41	D1	67	D11	PA8	I/O	FT	-	MCO_1, TIM1_CH1, I2C3_SCL, USART1_CK, USB_FS_SOF, SDIO_D1, EVENTOUT	-
30	42	D2	68	D10	PA9	I/O	FT	-	TIM1_CH2, I2C3_SMBA, USART1_TX, USB_FS_VBUS, SDIO_D2, EVENTOUT	OTG_FS_VBUS



Table 8. STM32F411xC/xE pin definitions (continued)

	Pir	n numb							inions (continued)	
UFQFPN48	LQFP64	WLCSP49	LQFP100	UFBGA100	Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
31	43	C2	69	C12	PA10	I/O	FT	-	TIM1_CH3, SPI5_MOSI/I2S5_SD, USART1_RX, USB_FS_ID, EVENTOUT	-
32	44	C1	70	B12	PA11 I/O		FT	1	TIM1_CH4, SPI4_MISO, USART1_CTS, USART6_TX, USB_FS_DM, EVENTOUT	
33	45	C3	71	A12	PA12	I/O	FT	-	TIM1_ETR, SPI5_MISO, USART1_RTS, USART6_RX, USB_FS_DP, EVENTOUT	-
34	46	В3	72	A11	PA13	I/O	FT	-	JTMS-SWDIO, EVENTOUT	-
-	-	-	73	C11	VCAP_2	S	-	-	-	-
35	47	B1	74	F11	VSS	S	-	-	-	-
36	48	B2	75	G11	VDD	S	1	-	-	-
37	49	A1	76	A10	PA14	I/O	FT	-	JTCK-SWCLK, EVENTOUT	-
38	50	A2	77	A9	PA15	I/O	FT	-	JTDI, TIM2_CH1/TIM2_ETR, , SPI1_NSS/I2S1_WS, SPI3_NSS/I2S3_WS, USART1_TX, EVENTOUT	-
-	51	-	78	B11	PC10	I/O FT -		-	SPI3_SCK/I2S3_CK, SDIO_D2, EVENTOUT	-
-	52	-	79	C10	PC11	I/O	FT	-	I2S3ext_SD, SPI3_MISO, SDIO_D3, EVENTOUT	-
-	53	-	80	B10	PC12	PC12 I/O		-	SPI3_MOSI/I2S3_SD, SDIO_CK, EVENTOUT	-

Table 8. STM32F411xC/xE pin definitions (continued)

	Pir	numl	oer						maiorio (contanaca)	
UFQFPN48	LQFP64	WLCSP49	LQFP100	UFBGA100	Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
-	-	i	81	C9	PD0	I/O	FT - EVENTOUT TIM3_ETR, SDIO_CMD,		EVENTOUT	-
-	-	ı	82	B9	PD1	I/O	FT	ı	EVENTOUT	-
-	54	1	83	C8	PD2	I/O	FT	FT - SDIO_CMD, EVENTOUT		-
-	-	1	84	B8	PD3				-	
-	-	i	85	B7	PD4	I/O	FT	ı	USART2_RTS, EVENTOUT	-
-	-	-	86	A6	PD5	I/O	FT	-	USART2_TX, EVENTOUT	-
-	-	-	87	В6	PD6	I/O	FT	-	SPI3_MOSI/I2S3_SD, USART2_RX, EVENTOUT	-
-	-	-	88	A5	PD7	I/O	FT	-	USART2_CK, EVENTOUT	-
39	55	А3	89	A8	РВ3	I/O	FT	-	JTDO-SWO, TIM2_CH2, SPI1_SCK/I2S1_CK, SPI3_SCK/I2S3_CK, USART1_RX, I2C2_SDA, EVENTOUT	-
40	56	A4	90	A7	PB4	I/O	FT	-	JTRST, TIM3_CH1, SPI1_MISO, SPI3_MISO, I2S3ext_SD, I2C3_SDA, SDIO_D0, EVENTOUT	-
41	57	B4	91	C5	PB5	I/O	тс	-	TIM3_CH2, I2C1_SMBA, SPI1_MOSI/I2S1_SD, SPI3_MOSI/I2S3_SD, SDIO_D3, EVENTOUT	-
42	58	C4	92	B5	PB6	I/O	FT	-	TIM4_CH1, I2C1_SCL, USART1_TX, EVENTOUT	-



				Table	0. STW32F4117	KC/XE	: pii	uem	nitions (continued)	
	Pir	n numl	oer				4			
UFQFPN48	LQFP64	WLCSP49	LQFP100	UFBGA100	Pin name (function after reset) ⁽¹⁾	Pin type	I/O structure	Notes	Alternate functions	Additional functions
43	59	D4	93	B4	PB7	I/O	FT	-	TIM4_CH2, I2C1_SDA, USART1_RX, SDIO_D0, EVENTOUT	-
44	60	A5	94	A4	воото	-	В	-	-	VPP
45	61	B5	95	A3	PB8	I/O	FT	-	TIM4_CH3, TIM10_CH1, I2C1_SCL, SPI5_MOSI/I2S5_SD, I2C3_SDA, SDIO_D4, EVENTOUT	-
46	62	C5	96	В3	PB9	I/O	FT	-	TIM4_CH4, TIM11_CH1, I2C1_SDA, SPI2_NSS/I2S2_WS, I2C2_SDA, SDIO_D5, EVENTOUT	-
-	-	-	97	C3	PE0	I/O	FT	-	TIM4_ETR, EVENTOUT	-
-	-	-	98	A2	PE1	I/O	FT	1	EVENTOUT	-
47	63	A6	99	-	VSS	S	-	-	-	-
-	-	В6	-	НЗ	PDR_ON	I	FT	-	-	-

Table 8. STM32F411xC/xE pin definitions (continued)

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VDD

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^{1.} Function availability depends on the chosen device.

PC13, PC14 and PC15 are supplied through the power switch. Since the switch only sinks a limited amount of current (3 mA), the use of GPIOs PC13 to PC15 in output mode is limited:

- The speed should not exceed 2 MHz with a maximum load of 30 pF.

- These I/Os must not be used as a current source (e.g. to drive an LED).

^{3.} Main function after the first backup domain power-up. Later on, it depends on the contents of the RTC registers even after reset (because these registers are not reset by the main reset). For details on how to manage these I/Os, refer to the RTC register description sections in the STM32F411xx reference manual.

^{4.} FT = 5 V tolerant except when in analog mode or oscillator mode (for PC14, PC15, PH0 and PH1).

If the device is delivered in an UFBGA100 and the BYPASS_REG pin is set to VDD (Regulator off/internal reset ON mode), then PA0 is used as an internal Reset (active low)



Table 9. Alternate function mapping

		AF00	AF01	AF02	AF03	AF04	AF05	AF06	AF07	AF08	AF09	AF10	AF11	AF12	AF13	AF14	AF15
		AF00	AFUI	Al UZ	AF03	AFU4		SPI2/I2S2/	AFUI	Alto	Al US	AFIU	AFII	AFIZ	AFIS	AF 14	AF15
	Port	SYS_AF	TIM1/TIM2	TIM3/ TIM4/ TIM5	TIM9/ TIM10/ TIM11	I2C1/I2C2/ I2C3	SPI1/I2S1S PI2/ I2S2/SPI3/ I2S3	SPI3/ I2S3/SPI4/ I2S4/SPI5/ I2S5	SPI3/I2S3/ USART1/ USART2	USART6	12C2/ 12C3	OTG1_FS		SDIO			
	PA0	-	TIM2_CH1/ TIM2_ETR	TIM5_CH1	-	-	-	-	USART2_ CTS	-	-	-	-	-	-	-	EVENT OUT
	PA1	-	TIM2_CH2	TIM5_CH2	-	-	SPI4_MOSI /I2S4_SD	-	USART2_ RTS	-	-	-	-	-	-	-	EVENT OUT
	PA2	-	TIM2_CH3	TIM5_CH3	TIM9_CH1	-	I2S2_CKIN	-	USART2_ TX	-	-	-	ı	1	ı	ı	EVENT OUT
	PA3	-	TIM2_CH4	TIM5_CH4	TIM9_CH2	-	I2S2_MCK	-	USART2_ RX	-	-	-	-	1	-	-	EVENT OUT
	PA4	1	-	-	-	-	SPI1_NSS/I 2S1_WS	SPI3_NSS/I2 S3_WS	USART2_ CK	-	-	-	1	ı	i	1	EVENT OUT
	PA5	-	TIM2_CH1/ TIM2_ETR	-	-	-	SPI1_SCK/I 2S1_CK	-	-	-	-	-	-	-	-	-	EVENT OUT
	PA6	-	TIM1_BKIN	TIM3_CH1	-	-	SPI1_MISO	12S2_MCK	-	-	-	-	-	SDIO_ CMD	-	-	EVENT OUT
τA	PA7	1	TIM1_CH1N	TIM3_CH2	-	-	SPI1_MOSI /I2S1_SD	-	-	-	-	-	1	ı	i	1	EVENT OUT
Port,	PA8	MCO_1	TIM1_CH1	-	-	I2C3_ SCL	-	-	USART1_ CK	-	-	USB_FS_ SOF	1	SDIO_ D1	i	1	EVENT OUT
	PA9	-	TIM1_CH2	-	-	I2C3_ SMBA	-	-	USART1_ TX	-	-	USB_FS_ VBUS	-	SDIO_ D2	-	-	EVENT OUT
	PA10	-	TIM1_CH3	-	-	-	-	SPI5_MOSI/I 2S5_SD	USART1_ RX	-	-	USB_FS_ ID	-	-	-	-	EVENT OUT
	PA11	-	TIM1_CH4	-	-	-	-	SPI4_MISO	USART1_ CTS	USART6_ TX	-	USB_FS_ DM	-	-	-	-	EVENT OUT
	PA12	-	TIM1_ETR	-	-	-	-	SPI5_MISO	USART1_ RTS	USART6_ RX	-	USB_FS_ DP	-	-	-	-	EVENT OUT
	PA13	JTMS- SWDIO	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT
	PA14	JTCK- SWCLK	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT
	PA15	JTDI	TIM2_CH1/ TIM2_ETR	-	-	-	SPI1_NSS/I 2S1_WS	SPI3_NSS/I2 S3_WS	USART1_ TX	-	-	-	-	-	-	-	EVENT OUT

					Т	able 9. A	Iternate f	unction ma	apping (d	ontinue	d)						
		AF00	AF01	AF02	AF03	AF04	AF05	AF06	AF07	AF08	AF09	AF10	AF11	AF12	AF13	AF14	AF15
	Port	SYS_AF	TIM1/TIM2	TIM3/ TIM4/ TIM5	TIM9/ TIM10/ TIM11	I2C1/I2C2/ I2C3	SPI1/I2S1S PI2/ I2S2/SPI3/ I2S3	SPI2/I2S2/ SPI3/ I2S3/SPI4/ I2S4/SPI5/ I2S5	SPI3/I2S3/ USART1/ USART2	USART6	12C2/ 12C3	OTG1_FS		SDIO			
	PB0	-	TIM1_CH2N	TIM3_CH3	-	-	-	SPI5_SCK /I2S5_CK		-	-	-	-	-	-	-	EVENT OUT
	PB1	-	TIM1_CH3N	TIM3_CH4	-	-	-	SPI5_NSS /I2S5_WS		-	-	-	-	-	-	-	EVENT OUT
	PB2	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT
	PB3	JTDO- SWO	TIM2_CH2	-	-	-	SPI1_SCK/I 2S1_CK	SPI3_SCK /I2S3_CK	USART1_ RX	-	I2C2_SDA	-	-	-	-	-	EVENT OUT
	PB4	JTRST		TIM3_CH1	-	-	SPI1_MISO	SPI3_MISO	I2S3ext_S D	-	I2C3_SDA			SDIO_ D0	-	-	EVENT OUT
	PB5	-	-	TIM3_CH2	-	I2C1_SMB A	SPI1_MOSI /I2S1_SD	SPI3_MOSI/ I2S3_SD		-	-	-	-	SDIO_ D3	-	-	EVENT OUT
	PB6	-	-	TIM4_CH1	-	I2C1_SCL	-	-	USART1_ TX	-	-	-	-		-	-	EVENT OUT
Œ.	PB7	-	-	TIM4_CH2	-	I2C1_SDA	-	-	USART1_ RX	-	-	-	-	SDIO_ D0	-	-	EVENT OUT
Port B	PB8	-	-	TIM4_CH3	TIM10_CH1	I2C1_SCL	-	SPI5_MOSI/ I2S5_SD	-	-	I2C3_SDA	-	-	SDIO_ D4	-	-	EVENT OUT
	PB9	-	-	TIM4_CH4	TIM11_CH1	I2C1_SDA	SPI2_NSS/I 2S2_WS	-	-	-	I2C2_SDA	-	-	SDIO_ D5	-	-	EVENT OUT
	PB10	-	TIM2_CH3	-	-	I2C2_SCL	SPI2_SCK/I 2S2_CK	I2S3_MCK	-	-	-	-	-	SDIO_ D7	-	-	EVENT OUT
	PB11	-	TIM2_CH4	-	-	I2C2_SDA	I2S2_CKIN	-	-	-	-	-	-	-	-	-	EVENT OUT
	PB12	-	TIM1_BKIN	-	-	I2C2_SMB A	SPI2_NSS/I 2S2_WS	SPI4_NSS /I2S4_WS	SPI3_SCK /I2S3_CK	-	-	-	-	-	-	-	EVENT OUT
	PB13	-	TIM1_CH1N	-	-	-	SPI2_SCK/I 2S2_CK	SPI4_SCK/ I2S4_CK	-	-	-	-	-	-	-	-	EVENT OUT
	PB14	-	TIM1_CH2N	-	-	-	SPI2_MISO	I2S2ext_SD	-	-	-	-	-	SDIO_ D6	-	-	EVENT OUT
	PB15	RTC_50H z	TIM1_CH3N	-	-	-	SPI2_MOSI /I2S2_SD	-	-	-	-	-	-	SDIO_ CK	-	ı	EVENT OUT





Table 9. Alternate function mapping (continued)

		AF00	AF01	AF02	AF03	AF04	AF05	AF06	AF07	AF08	AF09	AF10	AF11	AF12	AF13	AF14	AF15
	Port	SYS_AF	TIM1/TIM2	TIM3/ TIM4/ TIM5	TIM9/ TIM10/ TIM11	I2C1/I2C2/ I2C3	SPI1/I2S1S PI2/ I2S2/SPI3/ I2S3	SPI2/I2S2/ SPI3/ I2S3/SPI4/ I2S4/SPI5/ I2S5	SPI3/I2S3/ USART1/ USART2	USART6	I2C2/ I2C3	OTG1_FS		SDIO			
	PC0	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT
	PC1	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT
	PC2	-	-	-	-	-	SPI2_MISO	I2S2ext_SD	-	-	-	-	-	-	-	-	EVENT OUT
	PC3	-	-	-	-	-	SPI2_MOSI /I2S2_SD	-	-	-	-	-	-	-	-	1	EVENT OUT
	PC4	-	-	-	-	-		-	-	-	-	-	-	-	-	-	EVENT OUT
	PC5	-	-	-	-	-		-	-	-	-	-	-	-	-	1	EVENT OUT
	PC6	-	-	TIM3_CH1	-	-	I2S2_MCK	-	-	USART6_ TX	-	-	-	SDIO_ D6	-	-	EVENT OUT
Port	PC7	-	-	TIM3_CH2	-	-	SPI2_SCK/I 2S2_CK	12S3_MCK	-	USART6_ RX	-	-	-	SDIO_ D7	-	-	EVENT OUT
	PC8	-	-	TIM3_CH3	-	-	-	-	-	USART6_ CK	-	-	-	SDIO_ D0	-	1	EVENT OUT
	PC9	MCO_2	-	TIM3_CH4	-	I2C3_SDA	I2S2_CKIN	-	-		-	-	-	SDIO_ D1	-	-	EVENT OUT
	PC10	-	-	-	-	-	-	SPI3_SCK/I2 S3_CK	-	-	-	-	-	SDIO_ D2	-	-	EVENT OUT
	PC11	-	-	-	-	-	I2S3ext_SD	SPI3_MISO	-	-	-	-	-	SDIO_ D3	-	1	EVENT OUT
	PC12	-	-	-	-	-	-	SPI3_MOSI/I 2S3_SD	-	-	-	-	-	SDIO_ CK	-	-	EVENT OUT
	PC13	-	-	-	=	-	-	-	-	-	-	-	-	-	-	1	-
	PC14	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-
	PC15	-	-	-	-	-	-	-	-	-	-	-	-	-	-	1	-

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		AF00	AF01	AF02	AF03	AF04	AF05	AF06	AF07	AF08	AF09	AF10	AF11	AF12	AF13	AF14	AF15
	Port	SYS_AF	TIM1/TIM2	TIM3/ TIM4/ TIM5	TIM9/ TIM10/ TIM11	I2C1/I2C2/ I2C3	SPI1/I2S1S PI2/ I2S2/SPI3/ I2S3	SPI2/I2S2/ SPI3/ I2S3/SPI4/ I2S4/SPI5/ I2S5	SPI3/I2S3/ USART1/ USART2	USART6	I2C2/ I2C3	OTG1_FS		SDIO			
	PD0	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT
	PD1	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT
	PD2	-	-	TIM3_ETR	-	-	-	-	-	-	-	-	-	SDIO_ CMD			EVENT OUT
	PD3	-	-	-	-	-	SPI2_SCK/I 2S2_CK		USART2_ CTS	-	-	-	-	_	-	-	EVENT OUT
	PD4	-	-	-	-	-	-	-	USART2_ RTS	-	-	-	-	-	-	-	EVENT OUT
	PD5	-	-	-	-	-	-	-	USART2_ TX	-	-	-	-	-	-	-	EVENT OUT
	PD6	-	-	-	-	-	SPI3_MOSI /I2S3_SD	-	USART2_ RX	-	-	-	-	-	-	-	EVENT OUT
Q.	PD7	-	-	-	-	-	-	-	USART2_ CK	-	-	-	-	-	-	-	EVENT OUT
Port D	PD8	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT
	PD9	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT
	PD10	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT
	PD11	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT
	PD12	-	-	TIM4_CH1	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT
	PD13	-	-	TIM4_CH2	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT
	PD14	-	-	TIM4_CH3	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT
	PD15	-	-	TIM4_CH4	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT





Table 9. Alternate function mapping (continued)

		AF00	AF01	AF02	AF03	AF04	AF05	AF06	AF07	AF08	AF09	AF10	AF11	AF12	AF13	AF14	AF15
	Port	SYS_AF	TIM1/TIM2	TIM3/ TIM4/ TIM5	TIM9/ TIM10/ TIM11	I2C1/I2C2/ I2C3	SPI1/I2S1S PI2/ I2S2/SPI3/ I2S3	SPI2/I2S2/ SPI3/ I2S3/SPI4/ I2S4/SPI5/ I2S5	SPI3/I2S3/ USART1/ USART2	USART6	12C2/ 12C3	OTG1_FS		SDIO			
	PE0	-	-	TIM4_ETR	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT
	PE1	-	-		-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT
	PE2	TRACECL K	-	-	-	-	SPI4_SCK/I 2S4_CK	SPI5_SCK/I2 S5_CK	-	-	-	-	-	-	-	-	EVENT OUT
	PE3	TRACED0	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT
	PE4	TRACED1	-	-	-	-	SPI4_NSS/I 2S4_WS	SPI5_NSS/I2 S5_WS	-	-	-	-	-	-	-	-	EVENT OUT
	PE5	TRACED2	-	-	TIM9_CH1	-	SPI4_MISO	SPI5_MISO	-	-	-	-	-	-	-	-	EVENT OUT
	PE6	TRACED3	-	-	TIM9_CH2	-	SPI4_MOSI /I2S4_SD	SPI5_MOSI/I 2S5_SD	-	-	-	-	-	-	-	-	EVENT OUT
Ē	PE7	-	TIM1_ETR	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT
Port E	PE8	-	TIM1_CH1N	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT
	PE9	-	TIM1_CH1	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT
	PE10	-	TIM1_CH2N	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT
	PE11	-	TIM1_CH2	-	-	-	SPI4_NSS/I 2S4_WS	SPI5_NSS/I2 S5_WS	-	-	-	-	-	-	-	-	EVENT OUT
	PE12	-	TIM1_CH3N	-	-	-	SPI4_SCK/I 2S4_CK	SPI5_SCK/I2 S5_CK	-	-	-	-	-	-	-	-	EVENT OUT
	PE13	-	TIM1_CH3	-	-	-	SPI4_MISO	SPI5_MISO	-	-	-	-	-	-	-	-	EVENT OUT
	PE14	-	TIM1_CH4	-	-	-	SPI4_MOSI /I2S4_SD	SPI5_MOSI/I 2S5_SD	-	-	-	-	-	-	-	-	EVENT OUT
	PE15	-	TIM1_BKIN	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENT OUT

T	able 9. A	Iternate f	unction n	napping (continue	d)

		AF00	AF01	AF02	AF03	AF04	AF05	AF06	AF07	AF08	AF09	AF10	AF11	AF12	AF13	AF14	AF15
	Port	SYS_AF	TIM1/TIM2	TIM3/ TIM4/ TIM5	TIM9/ TIM10/ TIM11	I2C1/I2C2/ I2C3	SPI1/I2S1S PI2/ I2S2/SPI3/ I2S3	SPI2/I2S2/ SPI3/ I2S3/SPI4/ I2S4/SPI5/ I2S5	SPI3/I2S3/ USART1/ USART2	USART6	I2C2/ I2C3	OTG1_FS		SDIO			
T H	PH0	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-
Por	PH1	1	-	-	1	-	-	-	-	-	ı	-	ı	-	-	-	1

5 Memory mapping

The memory map is shown in Figure 14.

Figure 14. Memory map 0xE010 0000 - 0xFFFF FFFF Reserved Cortex-M4 internal peripherals 0xE000 0000 - 0xE00F FFFF 0xDFFF FFFF Reserved 0x5004 0000 0x5003 FFFF AHB2 0x5000 0000 :4002 6800 - 0x4FFF FFFF 0x4002 67FF Reserved 0xFFFF FFFF 512-Mbyte block 7 Cortex-M4' internal AHB1 0xE000 0000 0xDFFF FFFF peripherals 512-Mbyte block 6 Not used 0xC000 0000 0xBFFF FFFF 0x4002 0000 Reserved 0x4001 4C00 - 0x4001 FFFF 0x4001 4BFF Reserved 0x6000 0000 0x5FFF FFFF 512-Mbyte block 2 Peripheral APB2 0x4000 0000 0x3FFF FFFF 512-Mbyte SRAM 0x2002 0001 - 0x3FFF FFFF Reserved 0x4001 0000 0x4000 7400 - 0x4000 FFFF 0x4000 73FF 0x2000 0000 SRAM (128 KB aliased by bit-banding) 0x2000 0000 - 0x2002 0000 0x1FFF FFFF 512-Mbyte block 0 Code Reserved 0x1FFF C008 - 0x1FFF FFFF Option bytes 0x1FFF C000 - 0x1FFF C007 Reserved 0x1FFF 7A10 - 0x1FFF BFFF 0x0000 0000 0x1FFF 0000 - 0x1FFF 7A0F Reserved 0x0808 0000 - 0x1FFE FFFF 0x0800 0000 - 0x0807 FFFF Flash memory 0x0008 0000 - 0x07FF FFFF Reserved APB1 Aliased to Flash, system, memory or SRAM depending, on the BOOT pins 0x0000 0000 - 0x0007 FFFF 0x4000 0000 MSv34706V1

Table 10. STM32F411xC/xE register boundary addresses

Bus	Boundary address	Peripheral
	0xE010 0000 - 0xFFFF FFFF	Reserved
Cortex [®] -M4	0xE000 0000 - 0xE00F FFFF	Cortex-M4 internal peripherals
	0x5004 0000 - 0xDFFF FFFF	Reserved

Table 10. STM32F411xC/xE register boundary addresses (continued)

Bus	Boundary address	Peripheral
AHB2	0x5000 0000 - 0x5003 FFFF	USB OTG FS
	0x4002 6800 - 0x4FFF FFFF	Reserved
	0x4002 6400 - 0x4002 67FF	DMA2
	0x4002 6000 - 0x4002 63FF	DMA1
	0x4002 5000 - 0x4002 4FFF	Reserved
	0x4002 3C00 - 0x4002 3FFF	Flash interface register
	0x4002 3800 - 0x4002 3BFF	RCC
	0x4002 3400 - 0x4002 37FF	Reserved
AHB1	0x4002 3000 - 0x4002 33FF	CRC
АПБТ	0x4002 2000 - 0x4002 2FFF	Reserved
	0x4002 1C00 - 0x4002 1FFF	GPIOH
	0x4002 1400 - 0x4002 1BFF	Reserved
	0x4002 1000 - 0x4002 13FF	GPIOE
	0x4002 0C00 - 0x4002 0FFF	GPIOD
	0x4002 0800 - 0x4002 0BFF	GPIOC
	0x4002 0400 - 0x4002 07FF	GPIOB
	0x4002 0000 - 0x4002 03FF	GPIOA

Table 10. STM32F411xC/xE register boundary addresses (continued)

Bus	Boundary address	Peripheral
	0x4001 5400- 0x4001 FFFF	Reserved
	0x4001 5000 - 0x4001 53FFF	SPI5/I2S5
	0x4001 4800 - 0x4001 4BFF	TIM11
	0x4001 4400 - 0x4001 47FF	TIM10
	0x4001 4000 - 0x4001 43FF	TIM9
	0x4001 3C00 - 0x4001 3FFF	EXTI
	0x4001 3800 - 0x4001 3BFF	SYSCFG
	0x4001 3400 - 0x4001 37FF	SPI4/I2S4
APB2	0x4001 3000 - 0x4001 33FF	SPI1/I2S1
APB2	0x4001 2C00 - 0x4001 2FFF	SDIO
	0x4001 2400 - 0x4001 2BFF	Reserved
	0x4001 2000 - 0x4001 23FF	ADC1
	0x4001 1800 - 0x4001 1FFF	Reserved
	0x4001 1400 - 0x4001 17FF	USART6
	0x4001 1000 - 0x4001 13FF	USART1
	0x4001 0400 - 0x4001 0FFF	Reserved
	0x4001 0000 - 0x4001 03FF	TIM1
	0x4000 7400 - 0x4000 FFFF	Reserved

Table 10. STM32F411xC/xE register boundary addresses (continued)

Bus	Boundary address	Peripheral
	0x4000 7000 - 0x4000 73FF	PWR
	0x4000 6000 - 0x4000 6FFF	Reserved
	0x4000 5C00 - 0x4000 5FFF	I2C3
	0x4000 5800 - 0x4000 5BFF	I2C2
	0x4000 5400 - 0x4000 57FF	I2C1
	0x4000 4800 - 0x4000 53FF	Reserved
	0x4000 4400 - 0x4000 47FF	USART2
	0x4000 4000 - 0x4000 43FF	I2S3ext
	0x4000 3C00 - 0x4000 3FFF	SPI3 / I2S3
APB1	0x4000 3800 - 0x4000 3BFF	SPI2 / I2S2
	0x4000 3400 - 0x4000 37FF	I2S2ext
	0x4000 3000 - 0x4000 33FF	IWDG
	0x4000 2C00 - 0x4000 2FFF	WWDG
	0x4000 2800 - 0x4000 2BFF	RTC & BKP Registers
	0x4000 1000 - 0x4000 27FF	Reserved
	0x4000 0C00 - 0x4000 0FFF	TIM5
	0x4000 0800 - 0x4000 0BFF	TIM4
	0x4000 0400 - 0x4000 07FF	TIM3
	0x4000 0000 - 0x4000 03FF	TIM2

6 Electrical characteristics

6.1 Parameter conditions

Unless otherwise specified, all voltages are referenced to V_{SS}.

6.1.1 Minimum and maximum values

Unless otherwise specified the minimum and maximum values are guaranteed in the worst conditions of ambient temperature, supply voltage and frequencies by tests in production on 100% of the devices with an ambient temperature at $T_A = 25$ °C and $T_A = T_A$ max (given by the selected temperature range).

Data based on characterization results, design simulation and/or technology characteristics are indicated in the table footnotes and are not tested in production. Based on characterization, the minimum and maximum values refer to sample tests and represent the mean value plus or minus three times the standard deviation (mean $\pm 3 \sigma$).

6.1.2 Typical values

Unless otherwise specified, typical data are based on T_A = 25 °C, V_{DD} = 3.3 V (for the 1.7 V \leq V_{DD} \leq 3.6 V voltage range). They are given only as design guidelines and are not tested.

Typical ADC accuracy values are determined by characterization of a batch of samples from a standard diffusion lot over the full temperature range, where 95% of the devices have an error less than or equal to the value indicated (mean $\pm 2 \sigma$).

6.1.3 Typical curves

Unless otherwise specified, all typical curves are given only as design guidelines and are not tested.

6.1.4 Loading capacitor

The loading conditions used for pin parameter measurement are shown in Figure 15.

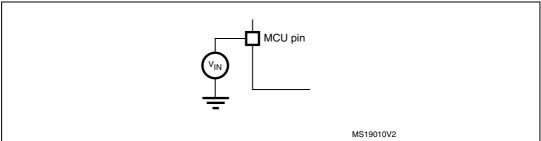
C = 50 pF

Figure 15. Pin loading conditions

6.1.5 Pin input voltage

The input voltage measurement on a pin of the device is described in *Figure 16*.

Figure 16. Input voltage measurement



6.1.6 Power supply scheme

VBAT Backup circuitry VBAT = (OSC32K,RTC, Power 1.65 to 3.6V Wakeup logic switch Backup registers) evel shifter IO **GPIOs** Logic VCAP 1 Kernel logic VCAP (CPU, digital & RAM) $2 \times 2.2 \,\mu\text{F}$ or $1 \times 4.7 \,\mu\text{F}$ VDD VDD Voltage 1/2/...4/5 regulator 6 × 100 nF VSS 1/2/...4/5 Flash memory BYPASS REG Reset PDR ON controller **VDD VDDA** VREF+ Analog: 100 nF VREF-ADC RCs, + 1 µF PLL VSSA MS31488V1

Figure 17. Power supply scheme

- 1. To connect PDR_ON pin, refer to Section 3.15: Power supply supervisor.
- 2. The 4.7 μ F ceramic capacitor must be connected to one of the V_{DD} pin.
- 3. VCAP_2 pad is only available on LQFP100 and UFBGA100 packages.
- 4. $V_{DDA}=V_{DD}$ and $V_{SSA}=V_{SS}$.

Caution:

Each power supply pair (for example V_{DD}/V_{SS} , V_{DDA}/V_{SSA}) must be decoupled with filtering ceramic capacitors as shown above. These capacitors must be placed as close as possible to, or below, the appropriate pins on the underside of the PCB to ensure good operation of the device. It is not recommended to remove filtering capacitors to reduce PCB size or cost. This might cause incorrect operation of the device.

6.1.7 Current consumption measurement

IDD_VBAT VBAT VDD VDD VDDA

Figure 18. Current consumption measurement scheme

6.2 Absolute maximum ratings

Stresses above the absolute maximum ratings listed in *Table 11: Voltage characteristics*, *Table 12: Current characteristics*, and *Table 13: Thermal characteristics* may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

Symbol	Ratings	Min	Max	Unit
V _{DD} -V _{SS}	External main supply voltage (including V_{DDA} , V_{DD} and V_{BAT}) ⁽¹⁾	-0.3	4.0	
	Input voltage on FT and TC pins ⁽²⁾	V _{SS} -0.3	V _{DD} +4.0	V
V _{IN}	Input voltage on any other pin	V _{SS} -0.3	4.0	
	Input voltage for BOOT0	V _{SS}	9.0	
∆V _{DDx}	Variations between different V _{DD} power pins	-	50	mV
V _{SSX} -V _{SS}	Variations between all the different ground pins	-	50	IIIV
V _{ESD(HBM)}	Electrostatic discharge voltage (human body model)	see Sectio Absolute n ratings (ele sensitivity)	naximum ectrical	

Table 11. Voltage characteristics

- 1. All main power (V_{DD} , V_{DDA}) and ground (V_{SS} , V_{SSA}) pins must always be connected to the external power supply, in the permitted range.
- 2. V_{IN} maximum value must always be respected. Refer to *Table 12* for the values of the maximum allowed injected current.

Table 12. Current characteristics

Symbol	Ratings	Max.	Unit
ΣI_{VDD}	Total current into sum of all V _{DD_x} power lines (source) ⁽¹⁾	160	
Σ I _{VSS}	Total current out of sum of all V _{SS_x} ground lines (sink) ⁽¹⁾	-160	
I _{VDD}	Maximum current into each V _{DD_x} power line (source) ⁽¹⁾	100	
I _{VSS}	Maximum current out of each V _{SS_x} ground line (sink) ⁽¹⁾	-100	
1	Output current sunk by any I/O and control pin	25	
I _{IO}	Output current sourced by any I/O and control pin	-25	mA
21	Total output current sunk by sum of all I/O and control pins (2)	120	
ΣI_{IO}	Total output current sourced by sum of all I/Os and control pins ⁽²⁾	-120	
(3)	Injected current on FT and TC pins (4)	5/.0	
I _{INJ(PIN)} ⁽³⁾	Injected current on NRST and B pins (4)	-5/+0	
ΣΙ _{ΙΝJ(PIN)}	Total injected current (sum of all I/O and control pins) ⁽⁵⁾	±25	

- All main power (V_{DD}, V_{DDA}) and ground (V_{SS}, V_{SSA}) pins must always be connected to the external power supply, in the permitted range.
- 2. This current consumption must be correctly distributed over all I/Os and control pins.
- 3. Negative injection disturbs the analog performance of the device. See note in Section 6.3.20: 12-bit ADC characteristics.
- 4. Positive injection is not possible on these I/Os and does not occur for input voltages lower than the specified maximum value.
- When several inputs are submitted to a current injection, the maximum ΣI_{INJ(PIN)} is the absolute sum of the positive and negative injected currents (instantaneous values).

Table 13. Thermal characteristics

Symbol	Ratings	Value	Unit
T _{STG}	Storage temperature range	-65 to +150	
T _J	Maximum junction temperature	130	
T _{LEAD}	Maximum lead temperature during soldering (WLCSP49, LQFP64/100, UFQFPN48, UFBGA100)	see note ⁽¹⁾	°C

Compliant with JEDEC Std J-STD-020D (for small body, Sn-Pb or Pb assembly), the ST ECOPACK[®]
7191395 specification, and the European directive on Restrictions on Hazardous Substances (ROHS
directive 2011/65/EU, July 2011).

6.3 Operating conditions

6.3.1 General operating conditions

Table 14. General operating conditions

Symbol	Parameter	Conditions	Min	Тур	Max	Unit	
		Power Scale3: Regulator ON, VOS[1:0] bits in PWR_CR register = 0x01	0	-	64		
f _{HCLK}	Internal AHB clock frequency	Power Scale2: Regulator ON, VOS[1:0] bits in PWR_CR register = 0x10	0	-	84	MHz	
		Power Scale1: Regulator ON, VOS[1:0] bits in PWR_CR register = 0x11	0	-	100		
f _{PCLK1}	Internal APB1 clock frequency		0	-	50	MHz	
f _{PCLK2}	Internal APB2 clock frequency		0	-	100	MHz	
V_{DD}	Standard operating voltage		1.7 ⁽¹⁾	-	3.6	٧	
V _{DDA} ⁽²⁾⁽³⁾	Analog operating voltage (ADC limited to 1.2 M samples)	Must be the same potential as $V_{DD}^{(4)}$	1.7 ⁽¹⁾	-	2.4	V	
V _{DDA} (E)(O)	Analog operating voltage (ADC limited to 2.4 M samples)	wider the same potential as VDD	2.4	-	3.6	V	
V_{BAT}	Backup operating voltage		1.65	-	3.6	V	
		VOS[1:0] bits in PWR_CR register = 0x01 Max frequency 64 MHz	1.08	1.14	1.20 ⁽⁵⁾		
V ₁₂	Regulator ON: 1.2 V internal voltage on VCAP_1/VCAP_2 pins	VOS[1:0] bits in PWR_CR register = 0x10 Max frequency 84 MHz	1.20 1.26 1.32 ⁽⁵⁾		٧		
		VOS[1:0] bits in PWR_CR register = 0x11 Max frequency 100 MHz	1.26	1.32	1.38		
	Regulator OFF: 1.2 V external	Max frequency 64 MHz	1.10	1.14	1.20		
V ₁₂	voltage must be supplied on	Max frequency 84 MHz	1.20	1.26	1.32	V	
	VCAP_1/VCAP_2 pins	Max frequency 100 MHz	1.26	1.32	1.38		
	Input voltage on RST, FT and	$2 \text{ V} \le \text{V}_{DD} \le 3.6 \text{ V}$	-0.3	-	5.5		
V_{IN}	TC pins ⁽⁶⁾	$V_{DD} \le 2 V$ -0.3 -		5.2	V		
	Input voltage on BOOT0 pin	-	0	-	9		
		UFQFPN48	-	-	625	mW	
	Power dissipation at	WLCSP49	-	-	392		
P_{D}	T _A = 85°C (range 6) or 105 °C (range 7) ⁽⁷⁾	LQFP64	-	-	425		
	105 C (range r).	LQFP100	-	-	465		
l		UFBGA100	-	-	323		

- 40

130

	Table 14. General operating conditions (continued)									
Symbol	Parameter	Conditions	Min	Тур	Max	Unit				
		UFQFPN48	-	-	156					
		WLCSP49	-	-	98					
P_{D}	Power dissipation at $T_A = 125 ^{\circ}\text{C} \text{ (range 3)}^{(7)}$	LQFP64	-	-	106	mW				
	Ac c (range c)	LQFP100	-	-	116					
		UFBGA100	-	-	81					
	Ambient temperature for	Maximum power dissipation	- 40	-	85					
	range 6	Low power dissipation ⁽⁸⁾	- 40	-	105					
TA	Ambient temperature for	Maximum power dissipation	- 40	-	105					
IA	range 7	Low power dissipation ⁽⁸⁾	- 40	-	125					
	Ambient temperature for	Maximum power dissipation	- 40	-	110	°C				
	range 3	Low power dissipation ⁽⁸⁾	- 40	-	130					
		Range 6	- 40	-	105					
TJ	Junction temperature range	Range 7	- 40	-	125					

Table 14. General operating conditions (continued)

- 1. V_{DD}/V_{DDA} minimum value of 1.7 V with the use of an external power supply supervisor (refer to *Section 3.15.2: Internal reset OFF*).
- 2. When the ADC is used, refer to Table 65: ADC characteristics.
- 3. If VREF+ pin is present, it must respect the following condition: VDDA-VREF+ < 1.2 $\rm V.$
- 4. It is recommended to power V_{DD} and V_{DDA} from the same source. A maximum difference of 300 mV between V_{DD} and V_{DDA} can be tolerated during power-up and power-down operation.
- 5. Guaranteed by test in production.
- 6. To sustain a voltage higher than VDD+0.3, the internal Pull-up and Pull-Down resistors must be disabled

Range 3

- 7. If T_A is lower, higher P_D values are allowed as long as T_J does not exceed T_{Jmax}.
- 8. In low power dissipation state, T_A can be extended to this range as long as T_J does not exceed T_{Jmax} .

Table 15. Features depending on the operating power supply range

Operating power supply range	ADC operation	Maximum Flash memory access frequency with no wait states (f _{Flashmax})	Maximum Flash memory access frequency with wait states (1)(2)	I/O operation	Clock output frequency on I/O pins ⁽³⁾	Possible Flash memory operations
V _{DD} =1.7 to 2.1 V ⁽⁴⁾	Conversion time up to 1.2 Msps	16 MHz ⁽⁵⁾	100 MHz with 6 wait states	- No I/O compensation	up to 30 MHz	8-bit erase and program operations only
V _{DD} = 2.1 to 2.4 V	Conversion time up to 1.2 Msps	18 MHz	100 MHz with 5 wait states	- No I/O compensation	up to 30 MHz	16-bit erase and program operations



Operating power supply range	ADC operation	Maximum Flash memory access frequency with no wait states (f _{Flashmax})	Maximum Flash memory access frequency with wait states (1)(2)	I/O operation	Clock output frequency on I/O pins ⁽³⁾	Possible Flash memory operations
V _{DD} = 2.4 to 2.7 V	Conversion time up to 2.4 Msps	24 MHz	100 MHz with 4 wait states	I/O compensation works	up to 50 MHz	16-bit erase and program operations
V _{DD} = 2.7 to 3.6 V ⁽⁶⁾	Conversion time up to 2.4 Msps	30 MHz	100 MHz with 3 wait states	I/O compensation works	 up to 100 MHz when V_{DD} = 3.0 to 3.6 V up to 50 MHz when V_{DD} = 2.7 to 3.0 V 	32-bit erase and program operations

Table 15. Features depending on the operating power supply range (continued)

- Applicable only when the code is executed from Flash memory. When the code is executed from RAM, no wait state is required.
- 2. Thanks to the ART accelerator and the 128-bit Flash memory, the number of wait states given here does not impact the execution speed from Flash memory since the ART accelerator allows to achieve a performance equivalent to 0 wait state program execution.
- 3. Refer to Table 55: I/O AC characteristics for frequencies vs. external load.
- 4. V_{DD}/V_{DDA} minimum value of 1.7 V, with the use of an external power supply supervisor (refer to Section 3.15.2: Internal reset OFF).
- 5. Prefetch is not available. Refer to AN3430 application note for details on how to adjust performance and power.
- 6. The voltage range for the USB full speed embedded PHY can drop down to 2.7 V. However the electrical characteristics of D- and D+ pins will be degraded between 2.7 and 3 V.

6.3.2 VCAP_1/VCAP_2 external capacitors

Stabilization for the main regulator is achieved by connecting the external capacitor C_{EXT} to the VCAP_1 and VCAP_2 pins. For packages supporting only 1 VCAP pin, the 2 CEXT capacitors are replaced by a single capacitor.

C_{FXT} is specified in *Table 16*.

Figure 19. External capacitor C_{EXT}

C

ESR

R

R

Leak

MS19044V2

1. Legend: ESR is the equivalent series resistance.

Table 16. VCAP_1/VCAP_2 operating conditions⁽¹⁾

Symbol	Parameter	Conditions
CEXT	Capacitance of external capacitor with a single VCAP pin available	4.7 μF
ESR	ESR of external capacitor with a single VCAP pin available	< 1 Ω

When bypassing the voltage regulator, the two 2.2 μF V_{CAP} capacitors are not required and should be replaced by two 100 nF decoupling capacitors.

6.3.3 Operating conditions at power-up/power-down (regulator ON)

Subject to general operating conditions for T_A.

Table 17. Operating conditions at power-up / power-down (regulator ON)

Symbol	Parameter	Min	Max	Unit	
	V _{DD} rise time rate	20	∞	µs/V	
^t ∨DD	V _{DD} fall time rate	20	8	μο/ ν	

6.3.4 Operating conditions at power-up / power-down (regulator OFF)

Subject to general operating conditions for T_A .

Table 18. Operating conditions at power-up / power-down (regulator OFF)⁽¹⁾

Symbol	Parameter	Conditions	Min	Max	Unit
t _{VDD}	V _{DD} rise time rate	Power-up	20	∞	
	V _{DD} fall time rate	Power-down	20	∞	μs/V
+ .	V _{CAP_1} and V _{CAP_2} rise time rate	Power-up	20	∞	μ5/ ν
t _{VCAP}	V_{CAP_1} and V_{CAP_2} fall time rate	Power-down	20	8	

^{1.} To reset the internal logic at power-down, a reset must be applied on pin PA0 when V_{DD} reach below 1.08 V.

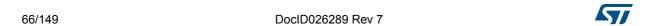
Note: This feature is only available for UFBGA100 package.

6.3.5 Embedded reset and power control block characteristics

The parameters given in *Table 19* are derived from tests performed under ambient temperature and V_{DD} supply voltage @ 3.3V.

Table 19. Embedded reset and power control block characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit			
		PLS[2:0]=000 (rising edge)	2.09	2.14	2.19				
		PLS[2:0]=000 (falling edge)	1.98	2.04	2.08				
		PLS[2:0]=001 (rising edge)	2.23	2.30	2.37				
		PLS[2:0]=001 (falling edge)	2.13	2.19	2.25				
		PLS[2:0]=010 (rising edge)	2.39	2.45	2.51	ı			
		PLS[2:0]=010 (falling edge)	2.29	2.35	2.39				
		PLS[2:0]=011 (rising edge)	2.54	2.60	2.65				
V	Programmable voltage	PLS[2:0]=011 (falling edge)	2.44	2.51	2.56	V			
V_{PVD}	detector level selection	PLS[2:0]=100 (rising edge)	2.70	2.76	2.82	V			
		PLS[2:0]=100 (falling edge)	2.59	2.66	2.71				
		PLS[2:0]=101 (rising edge)	2.86	2.93	2.99				
		PLS[2:0]=101 (falling edge)	2.77	2.82	2.89				
		PLS[2:0]=110 (rising edge)	2.96	3.03	3.10				
		PLS[2:0]=110 (falling edge)	2.85	2.93	2.99				
		PLS[2:0]=111 (rising edge)	3.07	3.14	3.21				
		PLS[2:0]=111 (falling edge)	2.95	3.03	3.09				
V _{PVDhyst} ⁽²⁾	PVD hysteresis	-	-	100	ı	mV			
V _{POR/PDR}	Power-on/power-down	Falling edge	1.60 ⁽¹⁾	1.68	1.76	5 _V			
*POR/PDR	reset threshold	Rising edge	1.64	1.72	1.80	v			
V _{PDRhyst} ⁽²⁾	PDR hysteresis	-	-	40	-	mV			
V	Brownout level 1	Falling edge	2.13	2.19	2.24				
V _{BOR1}	threshold	Rising edge	2.23	2.29	2.33				
V	Brownout level 2	Falling edge	2.44	2.50	2.56	.,			
V_{BOR2}	threshold	Rising edge	2.53	2.59	2.63	V			
V	Brownout level 3	Falling edge	2.75	2.83	2.88				
V_{BOR3}	threshold	Rising edge	2.85	2.92	2.97				
V _{BORhyst} ⁽²⁾	BOR hysteresis	-	-	100	-	mV			
T _{RSTTEMPO}	POR reset timing	-	0.5	1.5	3.0	ms			



Symbol	Parameter	Conditions	Min	Тур	Max	Unit
I _{RUSH} ⁽²⁾	In-Rush current on voltage regulator power- on (POR or wakeup from Standby)	-	-	160	200	mA
E _{RUSH} ⁽²⁾	In-Rush energy on voltage regulator power- on (POR or wakeup from Standby)	V _{DD} = 1.7 V, T _A = 125 °C, I _{RUSH} = 171 mA for 31 μs	-	-	5.4	μC

Table 19. Embedded reset and power control block characteristics (continued)

- The product behavior is guaranteed by design down to the minimum V_{POR/PDR} value.
- 2. Guaranteed by design.
- 3. The reset timing is measured from the power-on (POR reset or wakeup from V_{BAT}) to the instant when first instruction is fetched by the user application code.

6.3.6 Supply current characteristics

The current consumption is a function of several parameters and factors such as the operating voltage, ambient temperature, I/O pin loading, device software configuration, operating frequencies, I/O pin switching rate, program location in memory and executed binary code.

The current consumption is measured as described in *Figure 18: Current consumption measurement scheme*.

All the run-mode current consumption measurements given in this section are performed with a reduced code that gives a consumption equivalent to CoreMark code.

Typical and maximum current consumption

The MCU is placed under the following conditions:

- All I/O pins are in input mode with a static value at VDD or VSS (no load).
- All peripherals are disabled except if it is explicitly mentioned.
- The Flash memory access time is adjusted to both f_{HCLK} frequency and VDD ranges (refer to Table 15: Features depending on the operating power supply range).
- The voltage scaling is adjusted to f_{HCLK} frequency as follows:
 - Scale 3 for f_{HCLK} ≤ 64 MHz
 - Scale 2 for 64 MHz < f_{HCLK} ≤ 84 MHz
 - Scale 1 for 84 MHz < f_{HCLK} ≤ 100 MHz
- The system clock is HCLK, f_{PCI K1} = f_{HCI K}/2, and f_{PCI K2} = f_{HCI K}.
- External clock is 4 MHz and PLL is ON except if it is explicitly mentioned.
- The maximum values are obtained for V_{DD} = 3.6 V and a maximum ambient temperature (T_A), and the typical values for T_A= 25 °C and V_{DD} = 3.3 V unless otherwise specified.

Table 20. Typical and maximum current consumption, code with data processing (ART accelerator disabled) running from SRAM - V_{DD} = 1.7 V

				Тур		Ma	ax ⁽¹⁾		
Symbol	Parameter	Conditions	f _{HCLK} (MHz)	T _A = 25 °C	T _A = 25 °C	T _A = 85 °C	T _A = 105 °C	T _A = 125 °C	Unit
			100	21.4	23.0	23.6	24.0	25.0	
		External clock,	84	17.2	18.9 ⁽⁵⁾	19.1	19.2	20.2	
		PLL ON ⁽²⁾ , all peripherals	64	11.9	12.9	13.2	13.7	14.6	
		enabled ⁽³⁾⁽⁴⁾	50	9.4	10.1	10.4	11.0	11.9	
			20	4.3	4.8	5.0	5.6	6.5	
		HSI, PLL off, all	16	3.0	3.3	3.6	4.3	5.2	
l	Supply current in Run	peripherals enabled ⁽⁴⁾	1	0.5	0.7	1.0	1.7	2.6	mA
I _{DD}	mode		100	12.7	14.0	14.4	14.8	15.8	ША
		External clock,	84	10.2	11.6 ⁽⁵⁾	11.8	12.0	13.0	
		PLL on ⁽²⁾)all peripherals	64	7.1	7.9	8.2	8.7	9.7	
		disabled ⁽³⁾	50	5.6	6.3	6.5	7.1	8.0	
			20	2.5	3.0	3.3	3.9	4.8	
		HSI, PLL off, all	16	1.9	2.1	2.4	3.0	3.9	
	peripherals disabled ⁽⁴⁾	peripherals disabled ⁽⁴⁾	1	0.4	0.5	0.9	1.6	2.5	

^{1.} Guaranteed by characterization results.

^{2.} Refer to Table 41 and RM0383 for the possible PLL VCO setting

^{3.} When analog peripheral blocks such as ADC, HSE, LSE, HSI, or LSI are ON, an additional power consumption has to be considered.

^{4.} When the ADC is ON (ADON bit set in the ADC_CR2 register), add an additional power consumption of 1.6 mA for the analog part.

^{5.} Guaranteed by test in production.

Table 21. Typical and maximum current consumption, code with data processing (ART accelerator disabled) running from SRAM - V_{DD} = 3.6 V

						Ma	ax ⁽¹⁾		
Symbol	Parameter	Conditions	f _{HCLK} (MHz)	Тур	T _A = 25 °C	T _A = 85 °C	T _A = 105 °C	T _A = 125 °C	Unit
			100	21.7	23.3	23.9	24.3	25.3	
		External clock,	84	17.5	19.2 ⁽⁵⁾	19.4	19.5	20.5	
		PLL ON ⁽²⁾ , all peripherals	64	12.2	13.2	13.5	14.0	14.9	
		enabled ⁽³⁾⁽⁴⁾	50	9.6	10.4	10.7	11.2	12.1	
	Supply current in		20	4.5	5.0	5.3	5.9	6.8	
		HSI, PLL OFF, all peripherals enabled ⁽³⁾	16	3.0	3.3	3.6	4.3	5.2	
			1	0.5	0.7	1.0	1.7	2.6	mA
I _{DD}	Run mode		100	13.0	14.6 ⁽⁵⁾	14.6	14.9	16.0	ША
			84	10.5	11.9 ⁽⁵⁾	12.1	12.2	13.2	
			64	7.4	8.4 ⁽⁵⁾	8.8	8.9	9.9	
		disabled ⁽³⁾	50	5.9	6.6	6.8	7.3	8.2	
			20	2.8	3.3	3.5	4.2	5.1	
		HSI, PLL OFF, all	16	1.9	2.1	2.4	3.1	4.0	
		peripherals disabled ⁽³⁾	1	0.4	0.5	0.9	1.6	2.5	

^{1.} Guaranteed by characterization results.

^{2.} Refer to Table 41 and RM0383 for the possible PLL VCO setting

^{3.} When analog peripheral blocks such as ADC, HSE, LSE, HSI, or LSI are ON, an additional power consumption has to be considered.

^{4.} When the ADC is ON (ADON bit set in the ADC_CR2 register), add an additional power consumption of 1.6 mA for the analog part.

^{5.} Guaranteed by test in production.

Table 22. Typical and maximum current consumption in run mode, code with data processing (ART accelerator enabled except prefetch) running from Flash memory- V_{DD} = 1.7 V

Symbol	Parameter	Conditions	f _{HCLK} (MHz)	Тур	Max ⁽¹⁾				
					T _A = 25 °C	T _A = 85 °C	T _A = 105 °C	T _A = 125 °C	Unit
I _{DD}	Supply current in Run mode	External clock, PLL ON ⁽²⁾ , all peripherals enabled ⁽³⁾⁽⁴⁾	100	20.4	21.8	22.1	22.8	23.8	- mA
			84	16.5	17.6	17.8	18.6	19.6	
			64	11.4	12.3	12.5	13.1	14.1	
			50	9.0	9.7	10.0	10.6	11.6	
			20	4.6	5.0	5.3	6.0	7.0	
		HSI, PLL OFF ⁽²⁾ , all peripherals enabled ⁽³⁾	16	2.9	3.2	3.6	4.3	5.3	
			1	0.7	0.8	1.3	1.9	2.9	
		External clock, PLL ON ⁽²⁾ all peripherals disabled ⁽³⁾	100	11.2	12.2	12.4	13.2	14.2	
			84	9.1	9.9	10.1	10.9	11.9	
			64	6.4	7.0	7.3	7.9	8.9	
			50	5.1	5.6	5.9	6.6	7.6	
			20	2.6	3.0	3.3	4.0	5.0	
		HSI, PLL OFF ⁽²⁾ , all peripherals disabled ⁽³⁾	16	1.8	2.0	2.4	3.0	4.0	
			1	0.6	0.7	1.2	1.9	2.9	

^{1.} Guaranteed by characterization results.

^{2.} Refer to Table 41 and RM0383 for the possible PLL VCO setting

^{3.} Add an additional power consumption of 1.6 mA per ADC for the analog part. In applications, this consumption occurs only while the ADC is ON (ADON bit is set in the ADC_CR2 register).

^{4.} When the ADC is ON (ADON bit set in the ADC_CR2), add an additional power consumption of 1.6mA per ADC for the analog part.

Table 23. Typical and maximum current consumption in run mode, code with data processing (ART accelerator enabled except prefetch) running from Flash memory - V_{DD} = 3.6 V

Symbol	Parameter	Conditions	f _{HCLK} (MHz)	Тур	Max ⁽¹⁾				
					T _A = 25 °C	T _A = 85 °C	T _A = 105 °C	T _A = 125 °C	Unit
I _{DD}	Supply current in Run mode	External clock, PLL ON ⁽²⁾ , all peripherals enabled ⁽³⁾⁽⁴⁾	100	20.7	22.2	22.5	23.2	24.4	- mA
			84	16.8	18.0	18.3	19.0	20.1	
			64	11.8	12.7	12.9	13.6	14.6	
			50	9.3	10.2	10.4	11.1	12.0	
			20	4.8	5.5	5.8	6.5	7.4	
		HSI, PLL OFF ⁽²⁾ , all peripherals enabled ⁽³⁾	16	3.0	3.3	3.8	4.5	5.4	
			1	0.7	1.0	1.4	2.1	3.0	
		External clock, PLL ON ⁽²⁾ all peripherals disabled ⁽³⁾	100	11.6	12.6	12.9	13.6	14.8	
			84	9.7	10.2 ⁽⁵⁾	11.1	11.3	12.5	
			64	6.7	7.4	7.7	8.3	9.4	
			50	5.4	6.0	6.3	7.0	8.0	
			20	2.9	3.4	3.7	4.4	5.4	
		HSI, PLL OFF ⁽²⁾ , all peripherals disabled ⁽³⁾	16	1.9	2.2	2.6	3.3	4.3	
			1	0.7	0.9	1.3	2.1	3.1	

^{1.} Guaranteed by characterization results.

^{2.} Refer to Table 41 and RM0383 for the possible PLL VCO setting

^{3.} Add an additional power consumption of 1.6 mA per ADC for the analog part. In applications, this consumption occurs only while the ADC is ON (ADON bit is set in the ADC_CR2 register).

^{4.} When the ADC is ON (ADON bit set in the ADC_CR2), add an additional power consumption of 1.6mA per ADC for the analog part.

^{5.} Guaranteed by test in production.

Table 24. Typical and maximum current consumption in run mode, code with data processing (ART accelerator disabled) running from Flash memory - V_{DD} = 3.6 V

		Conditions	£			Max	x ⁽¹⁾		<u> </u>
Symbol	Parameter		f _{HCLK} (MHz)	Тур	T _A = 25 °C	T _A = 85 °C	T _A = 105 °C	T _A = 125 °C	Unit
		External clock, PLL	100	29.5	31.5	32.3	33.3	34.7	
			84	25.5	27.1	27.9	28.9	30.2	
		ON ⁽²⁾ , all peripherals	64	18.6	19.8	20.4	21.2	22.4	
	enabled ⁽³⁾⁽⁴⁾	50	15.2	16.4	16.9	17.7	18.7		
	Supply current		20	7.6	8.4	8.8	9.5	10.5	mA
		HSI, PLL OFF ⁽²⁾ , all peripherals enabled ⁽³⁾	16	4.8	5.2	5.7	6.5	7.5	
1			1	0.9	1.3	1.6	2.4	3.4	
I _{DD}	in Run mode		100	20.4	21.8	22.7	23.8	25.1	ША
			84	18.4	19.2 ⁽⁵⁾	20.9	21.1	22.4	
		External clock, PLL ON ⁽²⁾ all peripherals disabled ⁽³⁾	64	13.5	14.5	15.2	15.9	17.2	
			50	11.3	12.2	12.8	13.6	14.7	
			20	5.6	6.4	6.7	7.4	8.5	
		HSI, PLL OFF ⁽²⁾ , all	16	3.6	4.1	4.5	5.2	6.3	
		peripherals disabled ⁽³⁾	1	0.9	1.2	1.6	2.3	3.4	

^{1.} Guaranteed by characterization results.

^{2.} Refer to Table 41 and RM0383 for the possible PLL VCO setting

^{3.} Add an additional power consumption of 1.6 mA per ADC for the analog part. In applications, this consumption occurs only while the ADC is ON (ADON bit is set in the ADC_CR2 register).

^{4.} When the ADC is ON (ADON bit set in the ADC_CR2), add an additional power consumption of 1.6mA per ADC for the analog part.

^{5.} Guaranteed by test in production.

Table 25. Typical and maximum current consumption in run mode, code with data processing (ART accelerator enabled with prefetch) running from Flash memory - V_{DD} = 3.6 V

			f			Max	x ⁽¹⁾		
Symbol	Parameter	Conditions	f _{HCLK} (MHz)	Тур	T _A = 25 °C	T _A = 85 °C	T _A = 105 °C	T _A = 125 °C	Unit
		External clock, PLL	100	31.7	33.6	34.5	35.5	37.0	
			84	26.9	28.6	29.4	30.3	31.6	
		ON ⁽²⁾ , all peripherals	64	19.6	20.9	21.5	22.3	23.5	
		enabled ⁽³⁾⁽⁴⁾	50	15.6	16.7	17.2	18.0	19.1	
	Supply current		20	7.6	8.4	8.8	9.5	10.6	
		HSI, PLL OFF ⁽²⁾ , all peripherals enabled ⁽³⁾	16	5.1	5.6	6.1	6.8	7.9	
			1	1.0	1.3	1.7	2.3	3.4	mA
I _{DD}	in Run mode		100	22.5	24.2	24.9	26.0	27.3	ША
			84	19.5	21.1 ⁽⁵⁾	21.8	22.8	24.1	
		External clock, PLL ON ⁽²⁾ all peripherals disabled ⁽³⁾	64	14.5	15.7	16.3	17.1	18.3	
		an pempinerale aleasies	50	11.7	12.7	13.2	14.0	15.1	
			20	5.6	6.4	6.8	7.4	8.5	
		HSI, PLL OFF ⁽²⁾ , all	16	4.0	4.5	4.9	5.6	6.7	
	peri	peripherals disabled ⁽³⁾	1	0.9	1.2	1.6	2.2	3.3	

^{1.} Guaranteed by characterization results.

^{2.} Refer to Table 41 and RM0383 for the possible PLL VCO setting

^{3.} Add an additional power consumption of 1.6 mA per ADC for the analog part. In applications, this consumption occurs only while the ADC is ON (ADON bit is set in the ADC_CR2 register).

^{4.} When the ADC is ON (ADON bit set in the ADC_CR2), add an additional power consumption of 1.6mA per ADC for the analog part.

^{5.} Guaranteed by test in production.

Table 26. Typical and maximum current consumption in Sleep mode - V_{DD} = 3.6 V

		Conditions	f			Max	κ ⁽¹⁾		
Symbol	Parameter		f _{HCLK} (MHz)	Тур	T _A = 25 °C	T _A = 85 °C	T _A = 105 °C	T _A = 125 °C	Unit
		External clock, PLL	100	12.2	13.2	13.4	14.1	15.3	
			84	9.8	10.6	10.9	11.6	12.8	
		ON ⁽²⁾ , all peripherals	64	6.9	7.4	7.7	8.3	9.5	
	enabled ⁽³⁾⁽⁴⁾	50	5.4	5.9	6.2	6.8	8.0		
	Supply current		20	2.8	3.2	3.5	4.1	5.3	
		HSI, PLL OFF ⁽²⁾ , all peripherals enabled ⁽³⁾	16	1.3	1.7	2.2	2.8	4.0	
1			1	0.4	0.5	0.9	1.6	2.8	mA
I _{DD}	in Sleep mode	(2)	100	3.0	3.6	3.9	4.5	5.7	ША
			84	2.5	3.0	3.2	3.9	5.1	
		External clock, PLL ON ⁽²⁾ all peripherals disabled ⁽³⁾	64	1.9	2.2	2.5	3.0	4.2	
			50	1.6	1.9	2.1	2.7	3.9	
			20	1.1	1.4	1.7	2.3	3.5	
		HSI, PLL OFF ⁽²⁾ , all	16	0.4	0.5	0.9	1.6	2.8	
		peripherals disabled ⁽³⁾	1	0.3	0.4	0.8	1.5	2.7	

- 1. Guaranteed by characterization results.
- 2. Refer to Table 41 and RM0383 for the possible PLL VCO setting.
- 3. Add an additional power consumption of 1.6 mA per ADC for the analog part. In applications, this consumption occurs only while the ADC is ON (ADON bit is set in the ADC_CR2 register).
- 4. When the ADC is ON (ADON bit set in the ADC_CR2), add an additional power consumption of 1.6mA per ADC for the analog part.

Table 27. Typical and maximum current consumptions in Stop mode - V_{DD} = 1.7 V

			Typ ⁽¹⁾ Max ⁽¹⁾						
Symbol	Conditions	Parameter	T _A = 25 °C	T _A = 25 °C	T _A = 85 °C	T _A = 105 °C	T _A = 125 °C	Unit	
	•	Main regulator usage	112	142 ⁽²⁾	400	710	1200 ⁽²⁾		
	oscillators OFF, no independent watchdog	Low power regulator usage	42.6	67 ⁽²⁾	300	580	1044 ⁽²⁾		
I _{DD STOP}	Flash in Deep power	Main regulator usage	75	99 ⁽²⁾	310	580	993 ⁽²⁾	μΑ	
	down mode, all	Low power regulator usage	13.6	37 ⁽²⁾	265	550	1007 ⁽²⁾		
oscillators OFF, no independent watchdog		Low power low voltage regulator usage	9	28 ⁽²⁾	230	500	910 ⁽²⁾		

- 1. Guaranteed by characterization results.
- 2. Guaranteed by test in production.

Table 28. Typical and maximum current consumption in Stop mode - V_{DD} =3.6 V

			Тур	Max ⁽¹⁾				
Symbol	Conditions	Parameter	T _A = 25 °C	T _A = 25 °C	T _A = 85 °C	T _A = 105 °C	T _A = 125 °C	Unit
	Flash in Stop mode, all	Main regulator usage	113.7	145 ⁽²⁾	410	720	1217 ⁽²⁾	
	oscillators OFF, no independent watchdog	Low power regulator usage	43.1	68 ⁽²⁾	310	600	1073 ⁽²⁾	
I _{DD STOP}	Flash in Deep power	Main regulator usage	76.2	105 ⁽²⁾	320	600	1019 ⁽²⁾	μΑ
	down mode, all	Low power regulator usage	14	38 ⁽²⁾	275	560	1025 ⁽²⁾	
	oscillators OFF, no independent watchdog	Low power low voltage regulator usage	10	30 ⁽²⁾	235	510	928 ⁽²⁾	

- 1. Guaranteed by characterization results.
- 2. Guaranteed by test in production.

Table 29. Typical and maximum current consumption in Standby mode - V_{DD} = 1.7 V

		Conditions		Max ⁽²⁾				
Symbol	Parameter			T _A = 25 °C	T _A = 85 °C	T _A = 105 °C	1 ₄ =	Unit
	Cappiy Carrent	Low-speed oscillator (LSE) and RTC ON	2.4	4	12	25	50	μA
IDD_STBY	in Standby mode	RTC and LSE OFF	1.8	3 ⁽³⁾	11	24	49 ⁽³⁾	μΛ

- 1. When the PDR is OFF (internal reset is OFF), the typical current consumption is reduced by 1.2 μ A.
- 2. Guaranteed by characterization results.
- 3. Guaranteed by test in production.

Table 30. Typical and maximum current consumption in Standby mode - V_{DD} = 3.6 V

			Typ ⁽¹⁾					
Symbol	Parameter	Conditions		T _A = 25 °C	T _A = 85 °C	T _A = 105 °C	T _A = 125 °C	Unit
l .		Low-speed oscillator (LSE) and RTC ON	2.8	5	14	29	59	
IDD_STBY	in Standby mode	RTC and LSE OFF	2.1	4 ⁽³⁾	13.5	28	58 ⁽³⁾	μA

- 1. When the PDR is OFF (internal reset is OFF), the typical current consumption is reduced by 1.2 μ A.
- 2. Guaranteed by characterization results.
- 3. Guaranteed by test in production.

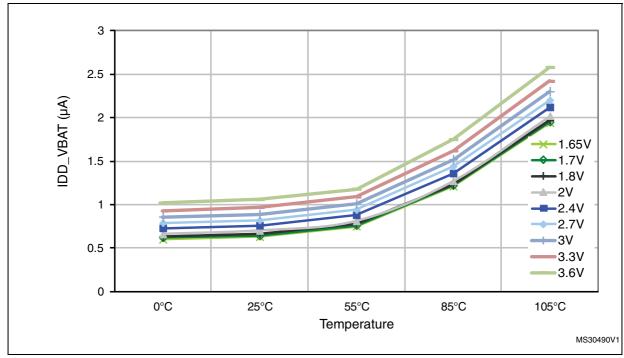
Table	31. Typical and maximum curre	nt consumptions in	V _{BAT} mode

			Тур			Max ⁽²⁾			
Symbol	Parameter	Conditions ⁽¹⁾	T _A = 25 °C			T _A = 85 °C	T _A = 105 °C	T _A = 125 °C	Unit
			V _{BAT} = 1.7 V	V _{BAT} = 2.4 V	V _{BAT} = 3.3 V	VE	_{BAT} = 3.6	V	
	Backup	Low-speed oscillator (LSE in low- drive mode) and RTC ON	0.7	0.8	1.0	3	5	6.8	
I _{DD_VBAT} domain supply current		Low-speed oscillator (LSE in high- drive mode) and RTC ON	1.5	1.6	1.9	3.8	5.8	8.6	μA
		RTC and LSE OFF	0.1	0.1	0.1	2	4	5.8	

^{1.} Crystal used: Abracon ABS07-120-32.768 kHz-T with a $\rm C_L$ of 6 pF for typical values.

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Figure 20. Typical V_{BAT} current consumption (LSE in low-drive mode and RTC ON)



^{2.} Guaranteed by characterization results.

I/O system current consumption

The current consumption of the I/O system has two components: static and dynamic.

I/O static current consumption

All the I/Os used as inputs with pull-up generate current consumption when the pin is externally held low. The value of this current consumption can be simply computed by using the pull-up/pull-down resistors values given in *Table 53: I/O static characteristics*.

For the output pins, any external pull-down or external load must also be considered to estimate the current consumption.

Additional I/O current consumption is due to I/Os configured as inputs if an intermediate voltage level is externally applied. This current consumption is caused by the input Schmitt trigger circuits used to discriminate the input value. Unless this specific configuration is required by the application, this supply current consumption can be avoided by configuring these I/Os in analog mode. This is notably the case of ADC input pins which should be configured as analog inputs.

Caution:

Any floating input pin can also settle to an intermediate voltage level or switch inadvertently, as a result of external electromagnetic noise. To avoid current consumption related to floating pins, they must either be configured in analog mode, or forced internally to a definite digital value. This can be done either by using pull-up/down resistors or by configuring the pins in output mode.

I/O dynamic current consumption

In addition to the internal peripheral current consumption (see *Table 33: Peripheral current consumption*), the I/Os used by an application also contribute to the current consumption. When an I/O pin switches, it uses the current from the MCU supply voltage to supply the I/O pin circuitry and to charge/discharge the capacitive load (internal or external) connected to the pin:

$$I_{SW} = V_{DD} \times f_{SW} \times C$$

where

 I_{SW} is the current sunk by a switching I/O to charge/discharge the capacitive load V_{DD} is the MCU supply voltage

f_{SW} is the I/O switching frequency

C is the total capacitance seen by the I/O pin: $C = C_{INT} + C_{EXT}$

The test pin is configured in push-pull output mode and is toggled by software at a fixed frequency.

Table 32. Switching output I/O current consumption

Symbol	Parameter	Conditions ⁽¹⁾	I/O toggling frequency (f _{SW})	Тур	Unit
			2 MHz	0.05	
			8 MHz	0.15	
			25 MHz	0.45	
		$V_{DD} = 3.3 V$ $C = C_{INT}$	50 MHz	0.85	
		O - OINT	60 MHz	1.00	
			84 MHz	1.40	
			90 MHz	1.67	
			2 MHz	0.10	
		V _{DD} = 3.3 V	8 MHz	0.35	
			25 MHz	1.05	
		C _{EXT} = 0 pF	50 MHz	2.20	
		$C = C_{INT} + C_{EXT} + C_{S}$	60 MHz	2.40	
			84 MHz	3.55	
	I/O switching		90 MHz	4.23	
IDDIO			2 MHz	0.20	- mA
IDDIO	current		8 MHz	0.65	
		V _{DD} = 3.3 V	25 MHz	1.85	
		C _{EXT} =10 pF	50 MHz	2.45	
		$C = C_{INT} + C_{EXT} + C_{S}$	60 MHz	4.70	
			84 MHz	8.80	
			90 MHz	10.47	
			2 MHz	0.25	
		V _{DD} = 3.3 V	8 MHz	1.00	
		C _{EXT} = 22 pF	25 MHz	3.45	
		$C = C_{INT} + C_{EXT} + C_{S}$	50 MHz	7.15	
			60 MHz	11.55	
			2 MHz	0.32	
		V _{DD} = 3.3 V	8 MHz	1.27	
		$C_{EXT} = 33 \text{ pF}$ $C = C_{INT} + C_{EXT} + C_{S}$	25 MHz	3.88	
		INI -EXI 95	50 MHz	12.34	

^{1.} CS is the PCB board capacitance including the pad pin. CS = 7 pF (estimated value).



On-chip peripheral current consumption

The MCU is placed under the following conditions:

- At startup, all I/O pins are in analog input configuration.
- All peripherals are disabled unless otherwise mentioned.
- The ART accelerator is ON.
- Voltage Scale 2 mode selected, internal digital voltage V12 = 1.26 V.
- HCLK is the system clock at 84 MHz. f_{PCLK1} = f_{HCLK}/2, and f_{PCLK2} = f_{HCLK}.
 The given value is calculated by measuring the difference of current consumption
 - with all peripherals clocked off
 - with only one peripheral clocked on
- Ambient operating temperature is 25 °C and V_{DD}=3.3 V.

Table 33. Peripheral current consumption

Perip	pheral	I _{DD} (Typ)	Unit
	GPIOA	1.55	
	GPIOB	1.55	
	GPIOC	1.55	
	GPIOD	1.55	
	GPIOE	1.55	
	GPIOH	1.55	
	CRC	0.36	
AHB1 (up to 100 MHz)	DMA1 ⁽¹⁾	14.96	μA/MHz
(up to 100 Wil 12)	DMA1 ⁽²⁾	1.54N+2.66	
	DMA2 ⁽¹⁾	14.96	
	DMA2 ⁽²⁾	1.54N+2.66	
	TIM2	11.19	
	TIM3	8.57	
	TIM4	8.33	
	TIM5	11.19	
	PWR	0.71	
APB1	USART2	3.33	A /NALI=
(up to 50 MHz)	I2C1/2/3	3.10	μA/MHz
	SPI2 ⁽³⁾	2.62	
	SPI3 ⁽³⁾	2.86	
	12S2	1.90	
	12S3	1.67	
	WWDG	0.71	

Table 33. Peripheral current consumption (continued)

Perij	oheral	I _{DD} (Typ)	Unit
	TIM1	5.71	
	TIM9	2.86	
	TIM10	1.79	
	TIM11	2.02	
	OTG_FS	23.93	
APB2	ADC1 ⁽⁴⁾	2.98	/ . /
(up to 100 MHz)	SPI1	1.19	μA/MHz
	USART1	3.10	
	USART6	2.86	
	SDIO	5.95	
	SPI4	1.31	
	SYSCFG	0.71	

- 1. Valid if all the DMA streams are activated (please refer to the reference manual RM0383).
- 2. For N DMA streams activated (up to 8 activated streams, refer to the reference manual RM0383).
- 3. I2SMOD bit set in SPI_I2SCFGR register, and then the I2SE bit set to enable I2S peripheral.
- 4. When the ADC is ON (ADON bit set in the ADC_CR2 register), add an additional power consumption of 1.6 mA for the analog part.

6.3.7 Wakeup time from low-power modes

The wakeup times given in *Table 34* are measured starting from the wakeup event trigger up to the first instruction executed by the CPU:

- For Stop or Sleep modes: the wakeup event is WFE.
- WKUP (PA0) pin is used to wakeup from Standby, Stop and Sleep modes.

Figure 21. Low-power mode wakeup Wakeup from Stop mode, Option bytes are not reloaded main regulator CPU restart HSI restart Flash stop exit Regulator ramp-up Wakeup from Stop mode, main regulator, Option bytes are not reloaded flash in Deep power down mode CPU restart HSI restart Flash Deep Pd recovery Regulator ramp-up Wakeup from Stop, Option bytes are not reloaded regulator in low power mode CPU restart Regulator HSI restart Flash stop exit ramp-up Wakeup from Stop, Option bytes are not reloaded regulator in low power mode, flash in Deep power down mode CPU restart Regulator Flash Deep Pd recovery ramp-up Wakeup from Standby mode Regulator OFF CPU restart HSI restart Flash Deep Pd recovery Option bytes loading Regulator restart Wakeup from Sleep and Regulator Option bytes are not reloaded Flash in Deep power down ON Flash Deep Pd recovery CPU restart MS35542V1

All timings are derived from tests performed under ambient temperature and V_{DD} =3.3 V.

Symbol	Parameter	Min ⁽¹⁾	Typ ⁽¹⁾	Max ⁽¹⁾	Unit	
t _{WUSLEEP} (2)	Wakeup from Sleep mode		4	6	CPU clock cycle	
	Wakeup from Stop mode, usage of main regulator	-	13.5	14.5		
t _{WUSTOP} ⁽²⁾	Wakeup from Stop mode, usage of main regulator, Flash memory in Deep power down mode	-	105	111	116	
'WUSTOP' '	Wakeup from Stop mode, regulator in low power mode	-	21	33	μs	
	Wakeup from Stop mode, regulator in low power mode, Flash memory in Deep power down mode	ı	113	130		
t _{WUSTDBY} (2)(3)	Wakeup from Standby mode		314	407	μs	
t _{WUFLASH}	Wakeup of Flash from Flash_Stop mode	-	-	8	116	
	Wakeup of Flash from Flash Deep power down mode		-	100	μs	

Table 34. Low-power mode wakeup timings⁽¹⁾

6.3.8 **External clock source characteristics**

High-speed external user clock generated from an external source

In bypass mode the HSE oscillator is switched off and the input pin is a standard I/O. The external clock signal has to respect the Table 53. However, the recommended clock input waveform is shown in Figure 22.

The characteristics given in Table 35 result from tests performed using an high-speed external clock source, and under ambient temperature and supply voltage conditions summarized in Table 14.

	Table 35. High-speed external	user clock chara	cteristics	s
Symbol	Parameter	Conditions	Min	T
fuce out	External user clock source frequency ⁽¹⁾		1	

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f _{HSE_ext}	External user clock source frequency ⁽¹⁾		1	-	50	MHz
V _{HSEH}	OSC_IN input pin high level voltage		0.7V _{DD}	-	V_{DD}	V
V_{HSEL}	OSC_IN input pin low level voltage		V _{SS}	-	$0.3V_{\mathrm{DD}}$	V
t _{w(HSEH)}	OSC_IN high or low time ⁽¹⁾	-	5	ı	ı	ns
$t_{r(HSE)}$ $t_{f(HSE)}$	OSC_IN rise or fall time ⁽¹⁾		-	-	10	115
C _{in(HSE)}	OSC_IN input capacitance ⁽¹⁾	-	-	5	-	pF
DuCy _(HSE)	Duty cycle	-	45	-	55	%
Ι <u></u>	OSC_IN Input leakage current	$V_{SS} \leq V_{IN} \leq V_{DD}$	-	-	±1	μΑ

^{1.} Guaranteed by design.



^{1.} Guaranteed by characterization results.

^{2.} The wakeup times are measured from the wakeup event to the point in which the application code reads the first instruction.

^{3.} $t_{WUSTDBY}$ maximum value is given at -40 °C.

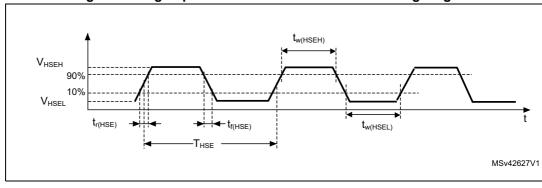


Figure 22. High-speed external clock source AC timing diagram

Low-speed external user clock generated from an external source

In bypass mode the LSE oscillator is switched off and the input pin is a standard I/O. The external clock signal has to respect the Table 53. However, the recommended clock input waveform is shown in Figure 23.

The characteristics given in *Table 36* result from tests performed using an low-speed external clock source, and under ambient temperature and supply voltage conditions summarized in Table 14.

Symbol Parameter Conditions Min Typ User External clock source 32.768 f_{LSE_ext} frequency⁽¹⁾ OSC32 IN input pin high level

Max 1000

 $V_{SS} \le V_{IN} \le V_{DD}$

 $0.7V_{DD}$

 V_{SS}

450

30

Table 36. Low-speed external user clock characteristics

 V_{LSEH}

 V_{LSEL}

tw(LSEH)

t_{w(LSEL)}

 $t_{r(LSE)}$

t_{f(LSE)}

C_{in(LSE)}

DuCy_(LSE)

voltage

Duty cycle

OSC32_IN input pin low level voltage

OSC32_IN high or low time(1)

OSC32 IN rise or fall time⁽¹⁾

OSC32_IN input capacitance⁽¹⁾

OSC32_IN Input leakage current

Unit

kHz

ns

pF

%

μΑ

 V_{DD}

 $0.3V_{DD}$

50

70

±1

5

^{1.} Guaranteed by design.

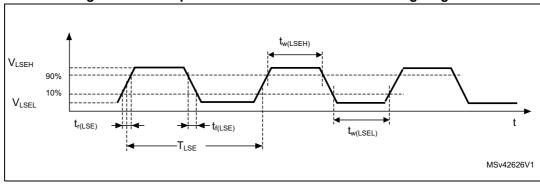


Figure 23. Low-speed external clock source AC timing diagram

High-speed external clock generated from a crystal/ceramic resonator

The high-speed external (HSE) clock can be supplied with a 4 to 26 MHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on characterization results obtained with typical external components specified in *Table 37*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f _{OSC_IN}	Oscillator frequency		4	-	26	MHz
R _F	Feedback resistor		-	200	-	kΩ
I _{DD}	HSE current consumption	V_{DD} =3.3 V, ESR= 30 Ω C_L =5 pF @25 MHz	-	450	-	
		V_{DD} =3.3 V, ESR= 30 Ω C_L =10 pF @25 MHz	-	530	-	μΑ
G _{m_crit_max}	Maximum critical crystal g _m	al g _m Startup		-	1	mA/V
t _{SU(HSE)} ⁽²⁾	Startup time	V _{DD} is stabilized	-	2	-	ms

Table 37. HSE 4-26 MHz oscillator characteristics⁽¹⁾

For C_{L1} and C_{L2} , it is recommended to use high-quality external ceramic capacitors in the 5 pF to 25 pF range (Typ.), designed for high-frequency applications, and selected to match the requirements of the crystal or resonator (see *Figure 24*). C_{L1} and C_{L2} are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of C_{L1} and C_{L2} . PCB and MCU pin capacitance must be included (10 pF can be used as a rough estimate of the combined pin and board capacitance) when sizing C_{L1} and C_{L2} .

Note: For information on selecting the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com.



^{1.} Guaranteed by design.

t_{SU(HSE)} is the startup time measured from the moment it is enabled (by software) to a stabilized 8 MHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer

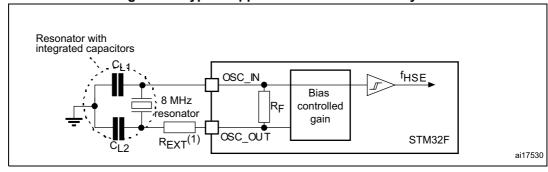


Figure 24. Typical application with an 8 MHz crystal

1. R_{EXT} value depends on the crystal characteristics.

Low-speed external clock generated from a crystal/ceramic resonator

The low-speed external (LSE) clock can be supplied with a 32.768 kHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on characterization results obtained with typical external components specified in *Table 38*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

The LSE high-power mode allows to cover a wider range of possible crystals but with a cost of higher power consumption.

(-L3E								
Symbol	Parameter	Conditions	Min	Тур	Max	Unit		
R_{F}	Feedback resistor	-	-	18.4	-	МΩ		
I _{DD}	LSE current consumption	Low-power mode (default)	-	-	1	μA		
	Consumption	High-drive mode	-	-	3			
G _m _crit_max	Maximum critical crystal	Startup, low-power mode	-	-	0.56	μΑ/V		
G _{m_} crit_max	g _m	Startup, high-drive mode	-	-	1.50	μν		
t _{SU(LSE)} ⁽²⁾	startup time	V _{DD} is stabilized	-	2	-	S		

Table 38. LSE oscillator characteristics (f_{LSE} = 32.768 kHz) ⁽¹⁾

Note:

For information on selecting the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com. For information about the LSE high-power mode, refer to the reference manual RM0383.

^{1.} Guaranteed by design.

t_{SU(LSE)} is the startup time measured from the moment it is enabled (by software) to a stabilized 32.768 kHz oscillation is reached. This value is guaranteed by characterization. It is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer.

Resonator with integrated capacitors

CL1

OSC32

RF

Bias controlled gain

STM32F

ai17531

Figure 25. Typical application with a 32.768 kHz crystal

6.3.9 Internal clock source characteristics

The parameters given in *Table 39* and *Table 40* are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 14*.

High-speed internal (HSI) RC oscillator

Table 39. HSI oscillator characteristics (1)

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
f _{HSI}	Frequency			-	16	-	MHz
		User-trimm register ⁽²⁾	Jser-trimmed with the RCC_CR register ⁽²⁾		-	1	%
	Factor		$T_A = -40 \text{ to } 125 ^{\circ}\text{C}^{(3)}$	- 8	-	5.5	
ACC _{HSI}		Factory-	$T_A = -40 \text{ to } 105 ^{\circ}\text{C}^{(3)}$	- 8	-	4.5	%
		calibrated	$T_A = -10 \text{ to } 85 ^{\circ}\text{C}^{(3)}$	- 4	-	4	%
			T _A = 25 °C ⁽⁴⁾	- 1	-	1	%
t _{su(HSI)} ⁽²⁾	HSI oscillator startup time			-	2.2	4	μs
I _{DD(HSI)} ⁽²⁾	HSI oscillator power consumption			-	60	80	μΑ

- 1. V_{DD} = 3.3 V, T_A = -40 to 125 °C unless otherwise specified.
- 2. Guaranteed by design.
- 3. Guaranteed by characterization results.
- 4. Factory calibrated non-soldered parts.

57/

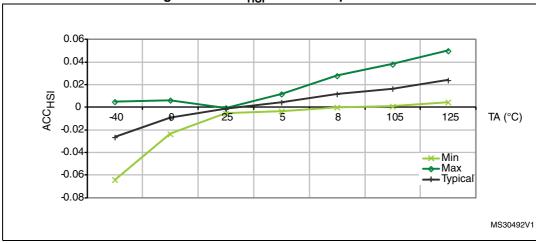


Figure 26. ACC_{HSI} versus temperature

1. Guaranteed by characterization results.

Low-speed internal (LSI) RC oscillator

Table 40. LSI oscillator characteristics ⁽¹⁾

Symbol	Parameter	Min	Тур	Max	Unit
f _{LSI} ⁽²⁾	Frequency	17	32	47	kHz
t _{su(LSI)} (3)	LSI oscillator startup time	-	15	40	μs
I _{DD(LSI)} ⁽³⁾	LSI oscillator power consumption	-	0.4	0.6	μA

- 1. V_{DD} = 3 V, T_A = -40 to 125 °C unless otherwise specified.
- 2. Guaranteed by characterization results.
- 3. Guaranteed by design.

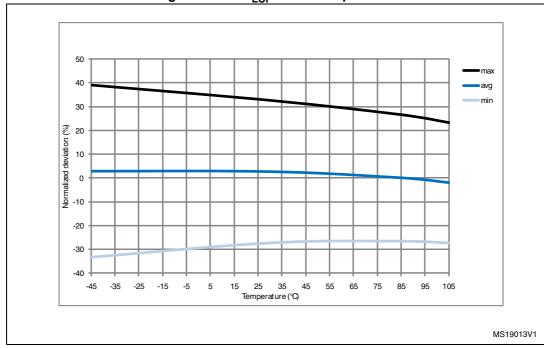


Figure 27. ACC_{LSI} versus temperature

6.3.10 PLL characteristics

The parameters given in *Table 41* and *Table 42* are derived from tests performed under temperature and V_{DD} supply voltage conditions summarized in *Table 14*.

Symbol	Parameter	Condition	ns	Min	Тур	Max	Unit
f _{PLL_IN}	PLL input clock ⁽¹⁾			0.95 ⁽²⁾	1	2.10	MHz
f _{PLL_OUT}	PLL multiplier output clock			24	-	100	MHz
f _{PLL48_OUT}	48 MHz PLL multiplier output clock			-	48	75	MHz
f _{VCO_OUT}	PLL VCO output				-	432	MHz
t. ook	PLL lock time	VCO freq = 100 MHz		75	-	200	116
t _{LOCK}		VCO freq = 432 MHz		100	-	300	μs
			RMS	-	25	-	
Jitter ⁽³⁾	Cycle-to-cycle jitter	System clock	peak to peak	-	±150	-	
		100 MHz	RMS	-	15	-	ps
	Period Jitter		peak to peak	-	±200	-	1

Table 41. Main PLL characteristics

Table 41. Main PLL characteristics (continued)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
I _{DD(PLL)} ⁽⁴⁾	PLL power consumption on VDD	VCO freq = 100 MHz VCO freq = 432 MHz	0.15 0.45	-	0.40 0.75	- mA
I _{DDA(PLL)} ⁽⁴⁾	PLL power consumption on VDDA	VCO freq = 100 MHz VCO freq = 432 MHz	0.30 0.55	-	0.40 0.85	IIIA

Take care of using the appropriate division factor M to obtain the specified PLL input clock values. The M factor is shared between PLL and PLLI2S.

Table 42. PLLI2S (audio PLL) characteristics

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
f _{PLLI2S_IN}	PLLI2S input clock ⁽¹⁾	-		0.95 ⁽²⁾	1	2.10	
f _{PLLI2S_OUT}	PLLI2S multiplier output clock	-		-	ï	216	MHz
f _{VCO_OUT}	PLLI2S VCO output	-		100	-	432	
+	PLLI2S lock time	VCO freq = 100 MHz	<u> </u>	75	-	200	
t _{LOCK}	PLLI23 lock tillle	VCO freq = 432 MHz	<u> </u>	100	-	300	μs
		Cycle to cycle at	RMS	-	90	-	
Jitter ⁽³⁾	Master I2S clock jitter	48 kHz period,	peak to peak	-	±280	-	
		Average frequency of 12.288 MHz N = 432, R = 5 on 1000 samples		-	90	-	ps
	WS I2S clock jitter	Cycle to cycle at 48 I on 1000 samples	Cycle to cycle at 48 KHz on 1000 samples		400	-	
I _{DD(PLLI2S)} ⁽⁴⁾	PLLI2S power consumption on V _{DD}	VCO freq = 100 MHz VCO freq = 432 MHz		0.15 0.45	-	0.40 0.75	- mA
I _{DDA(PLLI2S)} (4)	PLLI2S power consumption on V _{DDA}	VCO freq = 100 MHz VCO freq = 432 MHz		0.30 0.55	1	0.40 0.85	IIIA

^{1.} Take care of using the appropriate division factor M to have the specified PLL input clock values.

^{2.} Guaranteed by design.

^{3.} The use of two PLLs in parallel could degraded the Jitter up to +30%.

^{4.} Guaranteed by characterization results.

^{2.} Guaranteed by design.

^{3.} Value given with main PLL running.

^{4.} Guaranteed by characterization results.

6.3.11 PLL spread spectrum clock generation (SSCG) characteristics

The spread spectrum clock generation (SSCG) feature allows to reduce electromagnetic interferences (see *Table 49: EMI characteristics for LQFP100*). It is available only on the main PLL.

Table 43. SSCG parameter constraints

Symbol	Parameter	Min	Тур	Max ⁽¹⁾	Unit
f _{Mod}	Modulation frequency	·	·	10	kHz
md	Peak modulation depth	0.25	-	2	%
MODEPER * INCSTEP	(Modulation period) * (Increment Step)	i	-	2 ¹⁵ -1	-

^{1.} Guaranteed by design.

Equation 1

The frequency modulation period (MODEPER) is given by the equation below:

$$\mathsf{MODEPER} = \mathsf{round}[\mathsf{f}_{\mathsf{PLL_IN}} / \ (4 \times \mathsf{f}_{\mathsf{Mod}})]$$

 $f_{PLL\ IN}$ and f_{Mod} must be expressed in Hz.

As an example:

If f_{PLL_IN} = 1 MHz, and f_{MOD} = 1 kHz, the modulation depth (MODEPER) is given by equation 1:

MODEPER = round[
$$10^6 / (4 \times 10^3)$$
] = 250

Equation 2

Equation 2 allows to calculate the increment step (INCSTEP):

INCSTEP = round[
$$((2^{15} - 1) \times md \times PLLN) / (100 \times 5 \times MODEPER)$$
]

f_{VCO OUT} must be expressed in MHz.

With a modulation depth (md) = ±2 % (4 % peak to peak), and PLLN = 240 (in MHz):

INCSTEP = round[
$$((2^{15} - 1) \times 2 \times 240) / (100 \times 5 \times 250)$$
] = 126md(quantitazed)%

An amplitude quantization error may be generated because the linear modulation profile is obtained by taking the quantized values (rounded to the nearest integer) of MODPER and INCSTEP. As a result, the achieved modulation depth is quantized. The percentage quantized modulation depth is given by the following formula:

$$md_{quantized}\% = (MODEPER \times INCSTEP \times 100 \times 5) / ((2^{15} - 1) \times PLLN)$$

As a result:

$$md_{quantized}\% = (250 \times 126 \times 100 \times 5) / ((2^{15} - 1) \times 240) = 2.002\%$$
(peak)

577

Figure 28 and *Figure 29* show the main PLL output clock waveforms in center spread and down spread modes, where:

F0 is f_{PLL_OUT} nominal.

 T_{mode} is the modulation period.

md is the modulation depth.

Figure 28. PLL output clock waveforms in center spread mode

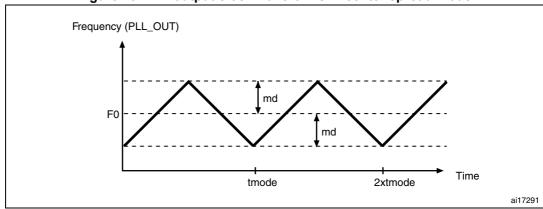
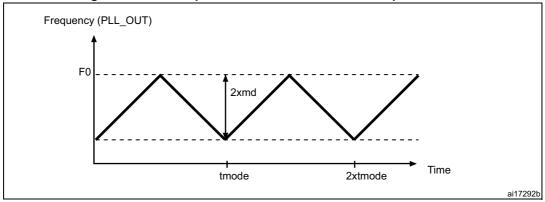


Figure 29. PLL output clock waveforms in down spread mode



6.3.12 Memory characteristics

Flash memory

The characteristics are given at T_A = -40 to 125 °C unless otherwise specified.

The devices are shipped to customers with the Flash memory erased.

Table 44. Flash memory characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
	Supply current	Write / Erase 8-bit mode, V _{DD} = 1.7 V	-	5	-	
I_{DD}		Write / Erase 16-bit mode, V _{DD} = 2.1 V	-	8	-	mA
		Write / Erase 32-bit mode, V _{DD} = 3.3 V	-	12	-	

Table 45. Flash memory programming

Symbol	Parameter	Conditions	Min ⁽¹⁾	Тур	Max ⁽¹⁾	Unit
t _{prog}	Word programming time	Program/erase parallelism (PSIZE) = x 8/16/32	-	16	100 ⁽²⁾	μs
		Program/erase parallelism (PSIZE) = x 8	-	400	800	
t _{ERASE16KB}	Sector (16 KB) erase time	Program/erase parallelism (PSIZE) = x 16	-	300	600	ms
		Program/erase parallelism (PSIZE) = x 32	-	250	500	
		Program/erase parallelism (PSIZE) = x 8	-	1200	2400	
t _{ERASE64KB}	Sector (64 KB) erase time	Program/erase parallelism (PSIZE) = x 16	-	700	1400	ms
		Program/erase parallelism (PSIZE) = x 32	-	550	1100	
	Sector (128 KB) erase time	Program/erase parallelism (PSIZE) = x 8	-	2	4	
t _{ERASE128KB}		Program/erase parallelism (PSIZE) = x 16	-	1.3	2.6	s
		Program/erase parallelism (PSIZE) = x 32	-	1	2	
		Program/erase parallelism (PSIZE) = x 8	-	8	16	
t _{ME}	Mass erase time	Program/erase parallelism (PSIZE) = x 16	-	5.5	11	s
		Program/erase parallelism (PSIZE) = x 32	-	4	8	
		32-bit program operation	2.7	-	3.6	V
V_{prog}	Programming voltage	16-bit program operation	2.1	-	3.6	٧
		8-bit program operation	1.7	-	3.6	V

^{1.} Guaranteed by characterization results.

Table 46. Flash memory programming with V_{PP} voltage

Symbol	Parameter	Conditions	Min ⁽¹⁾	Тур	Max ⁽¹⁾	Unit
t _{prog}	Double word programming		-	16	100 ⁽²⁾	μs
t _{ERASE16KB}	Sector (16 KB) erase time	T _A = 0 to +40 °C	-	230	-	
t _{ERASE64KB}	Sector (64 KB) erase time	V _{DD} = 3.3 V	-	490	-	ms
t _{ERASE128KB}	Sector (128 KB) erase time	$V_{PP} = 8.5 \text{ V}$	-	875	-	
t _{ME}	Mass erase time		-	3.50	-	s



^{2.} The maximum programming time is measured after 100K erase operations.

Symbol	Parameter	Conditions	Min ⁽¹⁾	Тур	Max ⁽¹⁾	Unit
V_{prog}	Programming voltage		2.7	-	3.6	V
V _{PP}	V _{PP} voltage range		7	-	9	V
I _{PP}	Minimum current sunk on the V _{PP} pin		10	-	-	mA
t _{VPP} (3)	Cumulative time during which V _{PP} is applied		-	-	1	hour

Table 46. Flash memory programming with V_{PP} voltage (continued)

- 1. Guaranteed by design.
- 2. The maximum programming time is measured after 100K erase operations.
- 3. V_{PP} should only be connected during programming/erasing.

Table 47. Flash memory endurance and data retention

Symbol	Darameter	Parameter Conditions -		Unit
Symbol	Parameter	Conditions	Min ⁽¹⁾	Offic
N _{END}	Endurance	T_A = - 40 to + 85 °C (temp. range 6) T_A = - 40 to + 105 °C (temp. range 7) T_A = - 40 to + 125 °C (temp. range 3)	10	kcycles
		1 kcycle ⁽²⁾ at T _A = 85 °C	30	
+		1 kcycle ⁽²⁾ at T _A = 105 °C	10	Years
t _{RET}		1 kcycle ⁽²⁾ at T _A = 125 °C	3	Teals
		10 kcycle ⁽²⁾ at T _A = 55 °C	20	

- 1. Guaranteed by characterization results.
- 2. Cycling performed over the whole temperature range.

6.3.13 EMC characteristics

Susceptibility tests are performed on a sample basis during device characterization.

Functional EMS (electromagnetic susceptibility)

While a simple application is executed on the device (toggling 2 LEDs through I/O ports). the device is stressed by two electromagnetic events until a failure occurs. The failure is indicated by the LEDs:

- Electrostatic discharge (ESD) (positive and negative) is applied to all device pins until a functional disturbance occurs. This test is compliant with the IEC 61000-4-2 standard.
- **FTB**: A burst of fast transient voltage (positive and negative) is applied to V_{DD} and V_{SS} through a 100 pF capacitor, until a functional disturbance occurs. This test is compliant with the IEC 61000-4-4 standard.

A device reset allows normal operations to be resumed.

The test results are given in *Table 49*. They are based on the EMS levels and classes defined in application note AN1709.



Symbol	Parameter	Conditions	Level/ Class
V _{FESD}	Voltage limits to be applied on any I/O pin to induce a functional disturbance	$V_{DD} = 3.3 \text{ V, LQFP100, WLCSP49,} \ T_A = +25 ^{\circ}\text{C, f}_{HCLK} = 100 \text{ MHz,} \ \text{conforms to IEC 61000-4-2}$	2B
V _{EFTB}	Fast transient voltage burst limits to be applied through 100 pF on V _{DD} and V _{SS} pins to induce a functional disturbance	$V_{DD} = 3.3 \text{ V, LQFP100, WLCSP49,}$ $T_{A} = +25 \text{ °C, } f_{HCLK} = 100 \text{ MHz,}$ conforms to IEC 61000-4-4	4A

Table 48. EMS characteristics for LQFP100 package

When the application is exposed to a noisy environment, it is recommended to avoid pin exposition to disturbances. The pins showing a middle range robustness are: PA0, PA1, PA2, on LQFP100 packages and PDR ON on WLCSP49.

As a consequence, it is recommended to add a serial resistor (1 k Ω maximum) located as close as possible to the MCU to the pins exposed to noise (connected to tracks longer than 50 mm on PCB).

Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It should be noted that good EMC performance is highly dependent on the user application and the software in particular.

Therefore it is recommended that the user applies EMC software optimization and prequalification tests in relation with the EMC level requested for his application.

Software recommendations

The software flowchart must include the management of runaway conditions such as:

- Corrupted program counter
- Unexpected reset
- Critical Data corruption (control registers...)

Prequalification trials

Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the NRST pin or the Oscillator pins for 1 second.

To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring (see application note AN1015).

Electromagnetic Interference (EMI)

The electromagnetic field emitted by the device are monitored while a simple application, executing EEMBC code, is running. This emission test is compliant with SAE IEC61967-2 standard which specifies the test board and the pin loading.

Table 49. EMI characteristics for LQFP100

Symbol	Parameter	Conditions	Monitored frequency band	Max vs. [f _{HSE} /f _{CPU}] 8/84 MHz	Unit
		Peak level $V_{DD} = 3.6 \text{ V}, T_A = 25 ^{\circ}\text{C}, \text{ conforming to }$ IEC61967-2	0.1 to 30 MHz	19	
6	Dook lovel		30 to 130 MHz	17	dΒμV
S _{EMI}	reak level		130 MHz to 1 GHz	12	
			SAE EMI Level	3.5	-

6.3.14 Absolute maximum ratings (electrical sensitivity)

Based on three different tests (ESD, LU) using specific measurement methods, the device is stressed in order to determine its performance in terms of electrical sensitivity.

Electrostatic discharge (ESD)

Electrostatic discharges (a positive then a negative pulse separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device (3 parts × (n+1) supply pins). This test conforms to the JESD22-A114/C101 standard.

Table 50. ESD absolute maximum ratings

Symbol	Ratings	Conditions		Class	Maximum value ⁽¹⁾	Unit
V _{ESD(HBM)}	Electrostatic discharge voltage (human body model)	T _A = +25 °C conforming to JESD22-A114		2	2000	
		T _A = +25 °C conforming to ANSI/ESD STM5.3.1	UFBGA100, UFQFPN48	4	500	٧
V _{ESD(CDM)}			WLCSP49	3	400	
, ,			LQPF64, LQFP100	3	250	

^{1.} Guaranteed by characterization results.

Static latchup

Two complementary static tests are required on six parts to assess the latchup performance:

- A supply overvoltage is applied to each power supply pin
- A current injection is applied to each input, output and configurable I/O pin

These tests are compliant with EIA/JESD 78A IC latchup standard.

Table 51. Electrical sensitivities

Symbol	Parameter	Conditions	Class
LU	Static latch-up class	T _A = + 125 °C conforming to JESD78A	II level A

6.3.15 I/O current injection characteristics

As a general rule, current injection to the I/O pins, due to external voltage below V_{SS} or above V_{DD} (for standard, 3 V-capable I/O pins) should be avoided during normal product operation. However, in order to give an indication of the robustness of the microcontroller in cases when abnormal injection accidentally happens, susceptibility tests are performed on a sample basis during device characterization.

Functional susceptibility to I/O current injection

While a simple application is executed on the device, the device is stressed by injecting current into the I/O pins programmed in floating input mode. While current is injected into the I/O pin, one at a time, the device is checked for functional failures.

The failure is indicated by an out of range parameter: ADC error above a certain limit (>5 LSB TUE), out of conventional limits of induced leakage current on adjacent pins (out of $-5 \,\mu\text{A/+0}\,\mu\text{A}$ range), or other functional failure (for example reset, oscillator frequency deviation).

Negative induced leakage current is caused by negative injection and positive induced leakage current by positive injection.

The test results are given in Table 52.

Table 52. I/O current injection susceptibility⁽¹⁾

		Functional s	usceptibility	
Symbol	Description	Negative injection	Positive injection	Unit
	Injected current on BOOT0 pin	-0	NA	
	Injected current on NRST pin	-0	NA	
I _{INJ}	Injected current on PB3, PB4, PB5, PB6, PB7, PB8, PB9, PC13, PC14, PC15, PH1, PDR_ON, PC0, PC1,PC2, PC3, PD1, PD5, PD6, PD7, PE0, PE2, PE3, PE4, PE5, PE6	-0	NA	mA
	Injected current on any other FT pin	– 5	NA	
	Injected current on any other pins	- 5	+5	

^{1.} NA = not applicable.



Note:

It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative currents.

6.3.16 I/O port characteristics

General input/output characteristics

Unless otherwise specified, the parameters given in *Table 53* are derived from tests performed under the conditions summarized in *Table 14*. All I/Os are CMOS and TTL compliant.

Table 53. I/O static characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
	FT, TC and NRST I/O input low level voltage	1.7 V≤V _{DD} ≤3.6 V	-	-	0.3V _{DD} ⁽¹⁾	
V _{IL}	BOOT0 I/O input low level voltage	1.75 V≤V _{DD} ≤3.6 V, -40 °C≤T _A ≤125 °C	-	-	0.1V _{DD} +0.1	V
		1.7 V≤V _{DD} ≤3.6 V, 0 °C≤T _A ≤125 °C	-	-	(2)	
	FT, TC and NRST I/O input high level voltage ⁽⁵⁾	1.7 V≤V _{DD} ≤3.6 V	0.7V _{DD} ⁽¹	-	-	
V _{IH}	BOOT0 I/O input high level voltage	1.75 V≤V _{DD} ≤3.6 V, -40 °C≤T _A ≤125 °C	0.17V _{DD} +0.7 ⁽²⁾	_	_	V
		1.7 V≤V _{DD} ≤3.6 V, 0 °C≤T _A ≤125 °C	+0.7(2)			
	FT, TC and NRST I/O input hysteresis	1.7 V≤V _{DD} ≤3.6 V	-	10% V _{DD} ⁽³⁾	-	V
V _{HYS}	BOOT0 I/O input hysteresis	1.75 V≤V _{DD} ≤3.6 V, -40 °C≤T _A ≤125 °C		100	_	mV
	Boo to we input thystoresis	1.7 V≤V _{DD} ≤3.6 V, 0 °C≤T _A ≤125 °C		100		111.0
	I/O input leakage current (4)	$V_{SS} \leq V_{IN} \leq V_{DD}$	-	-	±1	
I _{lkg}	I/O FT/TC input leakage current (5)	V _{IN} = 5 V	-	-	3	μΑ

Symbol	Param	Parameter		Min	Тур	Max	Unit
R _{PU}	Weak pull-up equivalent resistor ⁽⁶⁾	All pins except for PA10 (OTG_FS_ID)	$V_{IN} = V_{SS}$	30	40	50	
	resistor	PA10 (OTG_FS_ID)	-	7	10	14	kΩ
R _{PD}	Weak pull-down equivalent resistor ⁽⁷⁾	All pins except for PA10 (OTG_FS_ID)	$V_{IN} = V_{DD}$	30	40	50	KS2
		PA10 (OTG_FS_ID)	-	7	10	14	
C _{IO} (8)	I/O pin capacitano	e	-	-	5	-	pF

Table 53. I/O static characteristics (continued)

- 1. Guaranteed by test in production.
- 2. Guaranteed by design.
- 3. With a minimum of 200 mV.
- Leakage could be higher than the maximum value, if negative current is injected on adjacent pins, Refer to Table 52: I/O current injection susceptibility
- To sustain a voltage higher than VDD +0.3 V, the internal pull-up/pull-down resistors must be disabled. Leakage could be higher than the maximum value, if negative current is injected on adjacent pins. Refer to Table 52: I/O current injection susceptibility
- 6. Pull-up resistors are designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance is minimum (~10% order).
- Pull-down resistors are designed with a true resistance in series with a switchable NMOS. This NMOS contribution to the series resistance is minimum (~10% order).
- 8. Hysteresis voltage between Schmitt trigger switching levels. Guaranteed by characterization results.

All I/Os are CMOS and TTL compliant (no software configuration required). Their characteristics cover more than the strict CMOS-technology or TTL parameters. The coverage of these requirements for FT and TC I/Os is shown in *Figure 30*.

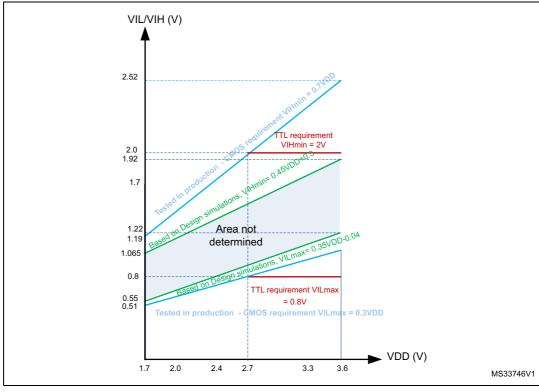


Figure 30. FT/TC I/O input characteristics

Output driving current

The GPIOs (general purpose input/outputs) can sink or source up to ± 20 mA (with a relaxed V_{OL}/V_{OH}) except PC13, PC14 and PC15 which can sink or source up to ± 3 mA. When using the PC13 to PC15 GPIOs in output mode, the speed should not exceed 2 MHz with a maximum load of 30 pF.

In the user application, the number of I/O pins which can drive current must be limited to respect the absolute maximum rating specified in *Section 6.2*. In particular:

- The sum of the currents sourced by all the I/Os on V_{DD}, plus the maximum Run consumption of the MCU sourced on V_{DD}, cannot exceed the absolute maximum rating ΣI_{VDD} (see *Table 12*).
- The sum of the currents sunk by all the I/Os on V_{SS} plus the maximum Run consumption of the MCU sunk on V_{SS} cannot exceed the absolute maximum rating ΣI_{VSS} (see *Table 12*).

Output voltage levels

Unless otherwise specified, the parameters given in *Table 54* are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 14*. All I/Os are CMOS and TTL compliant.

Table 54. Output voltage characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
V _{OL} ⁽¹⁾	Output low level voltage for an I/O pin	CMOS port ⁽²⁾	-	0.4	
V _{OH} ⁽³⁾	Output high level voltage for an I/O pin	I_{IO} = +8 mA 2.7 V \leq V _{DD} \leq 3.6 V	V _{DD} -0.4	-	V
V _{OL} ⁽¹⁾	Output low level voltage for an I/O pin	TTL port ⁽²⁾	-	0.4	
V _{OH} ⁽³⁾	Output high level voltage for an I/O pin	I_{IO} =+8 mA 2.7 V \leq V _{DD} \leq 3.6 V	2.4	-	V
V _{OL} ⁽¹⁾	Output low level voltage for an I/O pin	I _{IO} = +20 mA	-	1.3 ⁽⁴⁾	V
V _{OH} ⁽³⁾	Output high level voltage for an I/O pin	2.7 V ≤V _{DD} ≤3.6 V	V _{DD} -1.3 ⁽⁴⁾	-	V
V _{OL} ⁽¹⁾	Output low level voltage for an I/O pin	I _{IO} = +6 mA	-	0.4 ⁽⁴⁾	V
V _{OH} ⁽³⁾	Output high level voltage for an I/O pin	1.8 V ≤V _{DD} ≤3.6 V	V _{DD} -0.4 ⁽⁴⁾	-	V
V _{OL} ⁽¹⁾	Output low level voltage for an I/O pin	I _{IO} = +4 mA	-	0.4 ⁽⁵⁾	V
V _{OH} ⁽³⁾	Output high level voltage for an I/O pin	1.7 V ≤V _{DD} ≤3.6 V	V _{DD} -0.4 ⁽⁵⁾	-	

^{1.} The $I_{|O}$ current sunk by the device must always respect the absolute maximum rating specified in *Table 12*. and the sum of $I_{|O}$ (I/O ports and control pins) must not exceed I_{VSS} .

^{2.} TTL and CMOS outputs are compatible with JEDEC standards JESD36 and JESD52.

The I_{IO} current sourced by the device must always respect the absolute maximum rating specified in Table 12 and the sum of I_{IO} (I/O ports and control pins) must not exceed I_{VDD}.

^{4.} Guaranteed by characterization results.

^{5.} Guaranteed by design.

Input/output AC characteristics

The definition and values of input/output AC characteristics are given in *Figure 31* and *Table 55*, respectively.

Unless otherwise specified, the parameters given in *Table 55* are derived from tests performed under the ambient temperature and V_{DD} supply voltage conditions summarized in *Table 14*.

Table 55. I/O AC characteristics⁽¹⁾⁽²⁾

OSPEEDRy [1:0] bit value ⁽¹⁾	Symbol	Parameter	Conditions	Min	Тур	Max	Unit
		Maximum frequency ⁽³⁾	$C_L = 50 \text{ pF}, V_{DD} \ge 2.70 \text{ V}$	-	-	4	
	f 400 4		$C_L = 50 \text{ pF}, V_{DD} \ge 1.7 \text{ V}$	-	-	2	MHz
00	'max(IO)out		$C_L = 10 \text{ pF}, V_{DD} \ge 2.70 \text{ V}$	-	-	8	IVII IZ
00			C _L = 10 pF, V _{DD} ≥ 1.7 V	-	-	4	
	t _{f(IO)out} / t _{r(IO)out}	Output high to low level fall time and output low to high level rise time	C _L = 50 pF, V _{DD} = 1.7 V to 3.6 V	-	-	100	ns
		Maximum frequency ⁽³⁾	C _L = 50 pF, V _{DD} ≥ 2.70 V	-	-	25	
	f _{max(IO)out}		C _L = 50 pF, V _{DD} ≥ 1.7 V	-	-	12.5	MHz
			C _L = 10 pF, V _{DD} ≥ 2.70 V	-	-	50	
01			C _L = 10 pF, V _{DD} ≥ 1.7 V	-	-	20	
01	t _{f(IO)out} / t _{r(IO)out}	Output high to low level fall time and output low to high level rise time	C _L = 50 pF, V _{DD} ≥2.7 V	-	-	10	ns
			$C_L = 50 \text{ pF}, V_{DD} \ge 1.7 \text{ V}$	-	-	20	
			C _L = 10 pF, V _{DD} ≥ 2.70 V	-	-	6	
			C _L = 10 pF, V _{DD} ≥ 1.7 V	-	-	10	
			C _L = 40 pF, V _{DD} ≥ 2.70 V	-	-	50 ⁽⁴⁾	
)out Maximum frequency ⁽³⁾	C _L = 40 pF, V _{DD} ≥ 1.7 V	-	-	25	MHz
10	Imax(IO)out		C _L = 10 pF, V _{DD} ≥ 2.70 V	-	-	100 ⁽⁴⁾	
			C _L = 10 pF, V _{DD} ≥ 1.7 V	-	-	50 ⁽⁴⁾	
		$\begin{array}{c} t_{f(\text{IO})\text{out}} \\ t_{r(\text{IO})\text{out}} \end{array} \\ \begin{array}{c} \text{Output high to low level fall} \\ \text{time and output low to high} \\ \text{level rise time} \end{array}$	C _L = 40 pF, V _{DD} ≥ 2.70 V	-	-	6	
	t _{f(IO)out} /		C _L = 40 pF, V _{DD} ≥ 1.7 V	-	-	10	ns
			C _L = 10 pF, V _{DD} ≥ 2.70 V	-	-	4	
			C _L = 10 pF, V _{DD} ≥ 1.7 V	-	-	6	



OSPEEDRy [1:0] bit value ⁽¹⁾	Symbol	Parameter	Conditions	Min	Тур	Max	Unit
			$C_L = 30 \text{ pF}, V_{DD} \ge 2.70 \text{ V}$	-	-	100 ⁽⁴⁾	
	F _{max(IO)out}	Maximum frequency ⁽³⁾	$C_L = 30 \text{ pF}, V_{DD} \ge 1.7 \text{ V}$	-	-	50 ⁽⁴⁾	MHz
11	time and output low to high level rise time	Output high to low level fall	$C_L = 30 \text{ pF}, V_{DD} \ge 2.70 \text{ V}$	-	-	4	
			C _L = 30 pF, V _{DD} ≥ 1.7 V	-	-	6	ns
		C _L = 10 pF, V _{DD} ≥ 2.70 V	-	-	2.5	115	
			C _L = 10 pF, V _{DD} ≥ 1.7 V	-	-	4	
-	t _{EXTIpw}	Pulse width of external signals detected by the EXTI controller		10	-	-	ns

Table 55. I/O AC characteristics⁽¹⁾⁽²⁾ (continued)

- 1. Guaranteed by characterization results.
- The I/O speed is configured using the OSPEEDRy[1:0] bits. Refer to the STM32F4xx reference manual for a description of the GPIOx_SPEEDR GPIO port output speed register.
- 3. The maximum frequency is defined in Figure 31.
- 4. For maximum frequencies above 50 MHz and V_{DD} > 2.4 V, the compensation cell should be used.

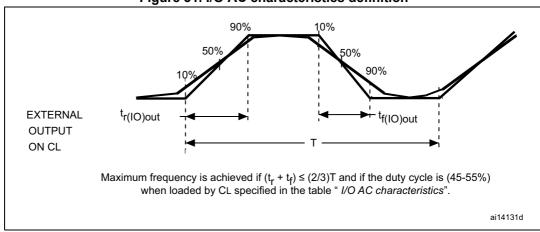


Figure 31. I/O AC characteristics definition

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6.3.17 NRST pin characteristics

The NRST pin input driver uses CMOS technology. It is connected to a permanent pull-up resistor, R_{PLI} (see *Table 53*).

Unless otherwise specified, the parameters given in *Table 56* are derived from tests performed under the ambient temperature and V_{DD} supply voltage conditions summarized in *Table 14*. Refer to *Table 53: I/O static characteristics* for the values of VIH and VIL for NRST pin.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
R _{PU}	Weak pull-up equivalent resistor ⁽¹⁾	$V_{IN} = V_{SS}$	30	40	50	kΩ
V _{F(NRST)} ⁽²⁾	NRST Input filtered pulse		-	-	100	ns
V _{NF(NRST)} ⁽²⁾	NRST Input not filtered pulse	V _{DD} > 2.7 V	300	ı	-	ns
T _{NRST_OUT}	Generated reset pulse duration	Internal Reset source	20	-	-	μs

Table 56. NRST pin characteristics

2. Guaranteed by design.

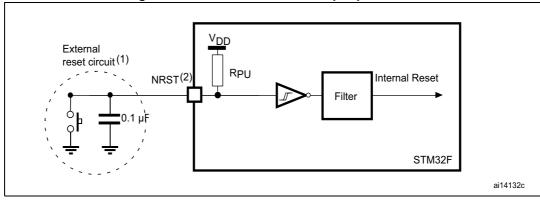


Figure 32. Recommended NRST pin protection

- 1. The reset network protects the device against parasitic resets.
- The user must ensure that the level on the NRST pin can go below the V_{IL(NRST)} max level specified in Table 56. Otherwise the reset is not taken into account by the device.

^{1.} The pull-up is designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance must be minimum (~10% order).

6.3.18 TIM timer characteristics

The parameters given in *Table 57* are guaranteed by design.

Refer to Section 6.3.16: I/O port characteristics for details on the input/output alternate function characteristics (output compare, input capture, external clock, PWM output).

Symbol	Parameter	Conditions ⁽³⁾	Min	Max	Unit
	Timer resolution time	AHB/APBx prescaler=1	1	-	t _{TIMxCLK}
t _{res(TIM)}		or 2 or 4, f _{TIMxCLK} = 100 MHz	11.9	-	ns
1es(TIM)		AHB/APBx prescaler>4,	1	-	t _{TIMxCLK}
		f _{TIMxCLK} = 100 MHz	11.9	-	ns
f _{EXT}	Timer external clock frequency on CH1 to CH4		0	f _{TIMxCLK} /2	MHz
'EXI		f _{TIMxCLK} = 100 MHz	0	50	MHz
Res _{TIM}	Timer resolution		-	16/32	bit
t _{COUNTER}	16-bit counter clock period when internal clock is selected	f _{TIMxCLK} = 100 MHz	0.0119	780	μs
t _{MAX_COUNT}	Maximum possible count with 32-bit counter		-	65536 × 65536	t _{TIMxCLK}
	with 32-bit counter	f _{TIMxCLK} = 100 MHz	-	51.1	S

Table 57. TIMx characteristics⁽¹⁾⁽²⁾

6.3.19 Communications interfaces

I²C interface characteristics

The I 2 C interface meets the requirements of the standard I 2 C communication protocol with the following restrictions: the I/O pins SDA and SCL are mapped to are not "true" opendrain. When configured as open-drain, the PMOS connected between the I/O pin and V_{DD} is disabled, but is still present.

The I²C characteristics are described in *Table 58*. Refer also to *Section 6.3.16: I/O port characteristics* for more details on the input/output alternate function characteristics (SDA and SCL).

The I²C bus interface supports standard mode (up to 100 kHz) and fast mode (up to 400 kHz). The I²C bus frequency can be increased up to 1 MHz. For more details about the complete solution, please contact your local ST sales representative.



^{1.} TIMx is used as a general term to refer to the TIM1 to TIM11 timers.

^{2.} Guaranteed by design.

The maximum timer frequency on APB1 is 50 MHz and on APB2 is up to 100 MHz, by setting the TIMPRE bit in the RCC_DCKCFGR register, if APBx prescaler is 1 or 2 or 4, then TIMxCLK = HCKL, otherwise TIMxCLK >= 4x PCLKx.

Table 58. I²C characteristics

Symbol	Parameter	Standard mode I ² C ⁽¹⁾⁽²⁾		Fast mode	Unit		
		Min	Max	Min	Max		
t _{w(SCLL)}	SCL clock low time	4.7	-	1.3	-	116	
t _{w(SCLH)}	SCL clock high time	4.0	-	0.6	-	μs	
t _{su(SDA)}	SDA setup time	250	-	100	-		
t _{h(SDA)}	SDA data hold time	0	3450 ⁽³⁾	0	900 ⁽⁴⁾		
t _{r(SDA)}	SDA and SCL rise time	-	1000	-	300	ns	
$\begin{matrix} t_{f(SDA)} \\ t_{f(SCL)} \end{matrix}$	SDA and SCL fall time	1	300	1	300		
t _{h(STA)}	Start condition hold time	4.0	-	0.6	-		
t _{su(STA)}	Repeated Start condition setup time	4.7	-	0.6	-	μs	
t _{su(STO)}	Stop condition setup time	4.0	-	0.6	-	μs	
t _{w(STO:STA)}	Stop to Start condition time (bus free)	4.7	-	1.3	-	μs	
t _{SP}	Pulse width of the spikes that are suppressed by the analog filter for standard fast mode	0	50 ⁽⁵⁾	0	50 ⁽⁵⁾	ns	
C _b	Capacitive load for each bus line	-	400	-	400	pF	

- 1. Guaranteed by design.
- 2. f_{PCLK1} must be at least 2 MHz to achieve standard mode I²C frequencies. It must be at least 4 MHz to achieve fast mode I²C frequencies, and a multiple of 10 MHz to reach the 400 kHz maximum I²C fast mode clock
- 3. The device must internally provide a hold time of at least 300 ns for the SDA signal in order to bridge the undefined region of the falling edge of SCL.
- The maximum data hold time has only to be met if the interface does not stretch the low period of SCL signal.
- 5. The minimum width of the spikes filtered by the analog filter is above $t_{\mbox{\footnotesize{SP}}}$ (max)

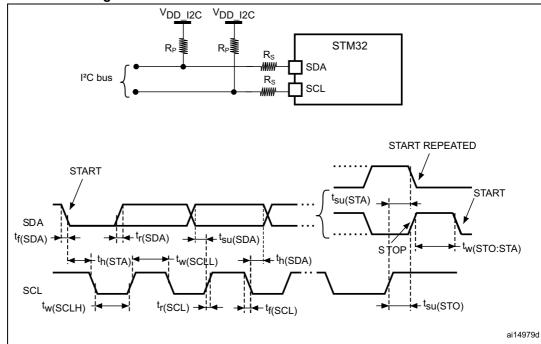


Figure 33. I²C bus AC waveforms and measurement circuit

- 1. R_S = series protection resistor.
- 2. R_P = external pull-up resistor.
- 3. $V_{DD\ I2C}$ is the I2C bus power supply.

Table 59. SCL frequency (f_{PCLK1}= 50 MHz, $V_{DD} = V_{DD_I2C} = 3.3 \text{ V})^{(1)(2)}$

f (I/LI-)	I2C_CCR value			
f _{SCL} (kHz)	R_P = 4.7 k Ω			
400	0x8019			
300	0x8021			
200	0x8032			
100	0x0096			
50	0x012C			
20	0x02EE			

- 1. R_P = External pull-up resistance, f_{SCL} = I^2C speed
- For speeds around 200 kHz, the tolerance on the achieved speed is of ±5%. For other speed ranges, the
 tolerance on the achieved speed is ±2%. These variations depend on the accuracy of the external
 components used to design the application.

SPI interface characteristics

Unless otherwise specified, the parameters given in *Table 60* for the SPI interface are derived from tests performed under the ambient temperature, f_{PCLKX} frequency and V_{DD} supply voltage conditions summarized in *Table 14*, with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 10
- Capacitive load C = 30 pF
- Measurement points are done at CMOS levels: 0.5V_{DD}

Refer to Section 6.3.16: I/O port characteristics for more details on the input/output alternate function characteristics (NSS, SCK, MOSI, MISO for SPI).

Table 60. SPI dynamic characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
		Master full duplex/receiver mode, 2.7 V < V _{DD} < 3.6 V SPI1/4/5	-	-	42	
	SPI clock frequency	Master full duplex/receiver mode, 3.0 V < V _{DD} < 3.6 V SPI1/4/5	-	-	50	
		Master transmitter mode 1.7 V < V _{DD} < 3.6 V SPI1/4/5	-	-	50	
f _{SCK} 1/t _{c(SCK)}		Master mode 1.7 V < V _{DD} < 3.6 V SPI1/2/3/4/5	-	-	25	MHz
		Slave transmitter/full duplex mode 2.7 V < V _{DD} < 3.6 V SPI1/4/5	-	-	38 ⁽²⁾	
		Slave receiver mode, 1.8 V < V _{DD} < 3.6 V SPI1/4/5	-	-	50	
		Slave mode, 1.8 V < V _{DD} < 3.6 V SPI1/2/3/4/5	-	-	25	
Duty(SCK)	Duty cycle of SPI clock frequency	Slave mode	30	50	70	%
t _{w(SCKH)} t _{w(SCKL)}	SCK high and low time	Master mode, SPI presc = 2	T _{PCLK} -1.5	T _{PCLK}	T _{PCLK} +1.5	ns
t _{su(NSS)}	NSS setup time	Slave mode, SPI presc = 2	3T _{PCLK}	-	-	ns
t _{h(NSS)}	NSS hold time	Slave mode, SPI presc = 2	2T _{PCLK}	-	-	ns
t _{su(MI)}	Data input setup time	Master mode	4	-	-	ns
t _{su(SI)}	Data input setup time	Slave mode	2.5	-	-	ns
t _{h(MI)}	Data input hold time	Master mode	7.5	-	-	ns
t _{h(SI)}	Data input floid tiffle	Slave mode	3.5	-	-	ns



Symbol	Parameter	Conditions	Min	Тур	Max	Unit
t _{a(SO})	Data output access time	Slave mode	7	-	21	ns
t _{dis(SO)}	Data output disable time	Slave mode	5	-	12	ns
+	Data output valid time	Slave mode (after enable edge), 2.7 V < V _{DD} < 3.6 V	-	11	13	ns
t _{v(SO)}	Data output valid time	Slave mode (after enable edge), 1.7 V < V _{DD} < 3.6 V	-	11	18.5	ns
t _{h(SO)}	Data output hold time	Slave mode (after enable edge), 1.7 V < V _{DD} < 3.6 V	8	-	-	ns
t _{v(MO)}	Data output valid time	Master mode (after enable edge)	-	4	6	ns
t _{h(MO)}	Data output hold time	Master mode (after enable edge)	0	-	-	ns

Table 60. SPI dynamic characteristics⁽¹⁾ (continued)

Maximum frequency in Slave transmitter mode is determined by the sum of t_{v(SO)} and t_{su(MI)} which has to fit into SCK low or high phase preceding the SCK sampling edge. This value can be achieved when the SPI communicates with a master having t_{su(MI)} = 0 while Duty(SCK) = 50%

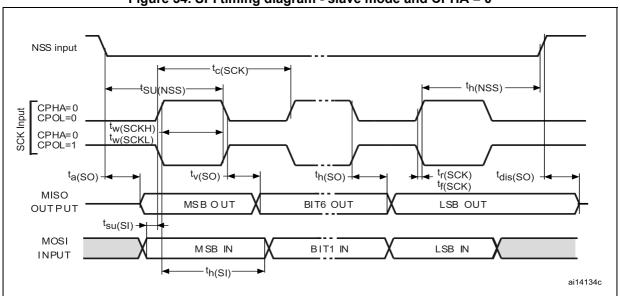


Figure 34. SPI timing diagram - slave mode and CPHA = 0

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^{1.} Guaranteed by characterization results.

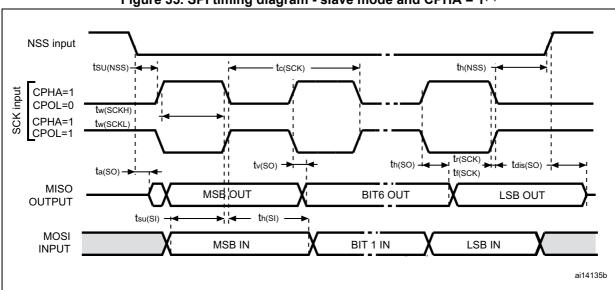
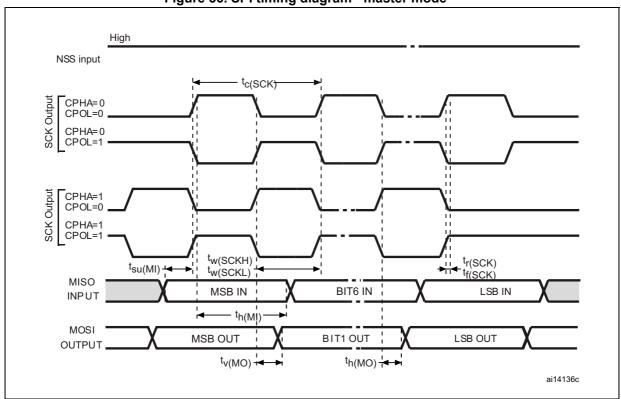


Figure 35. SPI timing diagram - slave mode and CPHA = $1^{(1)}$





I²S interface characteristics

Unless otherwise specified, the parameters given in *Table 61* for the I²S interface are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and V_{DD} supply voltage conditions summarized in *Table 14*, with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 10
- Capacitive load C = 30 pF
- Measurement points are done at CMOS levels: 0.5V_{DD}

Refer to Section 6.3.16: I/O port characteristics for more details on the input/output alternate function characteristics (CK, SD, WS).

Symbol	Parameter	Conditions	Min	Max	Unit
f _{MCK}	I2S Main clock output	-	256x8K	256xFs ⁽²⁾	MHz
£	IOC aloak from an and	Master data: 32 bits	-	64xFs	N 41 1-
f _{CK}	I2S clock frequency	Slave data: 32 bits	-	64xFs	MHz
D _{CK}	I2S clock frequency duty cycle	Slave receiver	30	70	%
t _{v(WS)}	WS valid time	Master mode	0	7	
t _{h(WS)}	WS hold time	Master mode	1.5	-	
t _{su(WS)}	WS setup time	Slave mode	1.5	-	
t _{h(WS)}	WS hold time	Slave mode	3	-	
t _{su(SD_MR)}	Data input setup time	Master receiver	1	-	
t _{su(SD_SR)}	Data input setup time	Slave receiver	2.5	-	ns
t _{h(SD_MR)}	Data input hold time	Master receiver	7	-	
t _{h(SD_SR)}	Data input hold time	Slave receiver	2.5	-	
t _{v(SD_ST)}	Data output valid time	Slave transmitter (after enable edge)	-	20	
t _{v(SD_MT)}	Data output valid time	Master transmitter (after enable edge)	-	6	
t _{h(SD_ST)}	Data and all discount	Slave transmitter (after enable edge)	8	-	
t _{h(SD MT)}	Data output hold time	Master transmitter (after enable edge)	2	-	

Table 61. I²S dynamic characteristics⁽¹⁾

Note: Refer to the I2S section of RM0383 reference manual for more details on the sampling frequency (F_S) .

> f_{MCK} , f_{CK} , and D_{CK} values reflect only the digital peripheral behavior. The values of these parameters might be slightly impacted by the source clock precision. D_{CK} depends mainly on the value of ODD bit. The digital contribution leads to a minimum value of (I2SDIV/(2*I2SDIV+ODD) and a maximum value of (I2SDIV+ODD)/(2*I2SDIV+ODD). FS maximum value is supported for each mode/condition.

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^{1.} Guaranteed by characterization results.

^{2.} The maximum value of 256xFs is 50 MHz (APB1 maximum frequency).

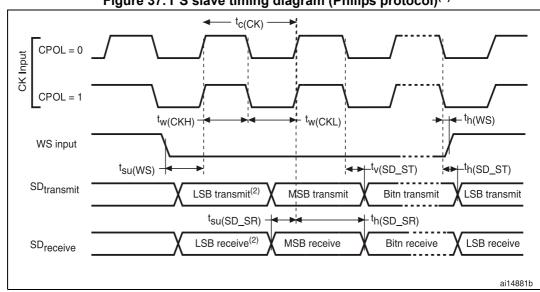
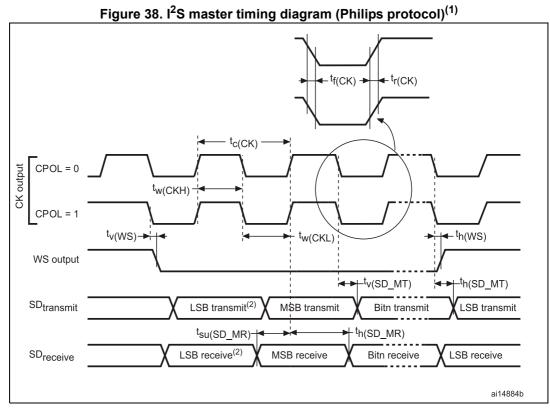


Figure 37. I²S slave timing diagram (Philips protocol)⁽¹⁾

 LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first byte.



 LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first byte.

USB OTG full speed (FS) characteristics

This interface is present in USB OTG FS controller.

Table 62. USB OTG FS startup time

Symbol	Parameter	Max	Unit
t _{STARTUP} ⁽¹⁾	USB OTG FS transceiver startup time	1	μs

^{1.} Guaranteed by design.

Table 63. USB OTG FS DC electrical characteristics

Symbol		Parameter	Conditions	Min. ⁽¹⁾	Тур.	Max. ⁽¹⁾	Unit
	V_{DD}	USB OTG FS operating voltage		3.0 ⁽²⁾	-	3.6	V
Input	V _{DI} ⁽³⁾	Differential input sensitivity	I(USB_FS_DP/DM)	0.2	-	-	
levels	V _{CM} ⁽³⁾	Differential common mode range	Includes V _{DI} range	0.8	-	2.5	V
	V _{SE} ⁽³⁾	Single ended receiver threshold		1.3	-	2.0	
Output	V _{OL}	Static output level low	R_L of 1.5 k Ω to 3.6 $V^{(4)}$	-	-	0.3	V
levels	V _{OH}	Static output level high	R_L of 15 k Ω to $V_{SS}^{(4)}$	2.8	-	3.6	v
R _F	PD	PA11, PA12 (USB_FS_DM/DP)	V _{IN} = V _{DD}	17	21	24	
		PA9 (OTG_FS_VBUS)		0.65	1.1	2.0	kΩ
R _F	·U	PA11, PA12 (USB_FS_DM/DP)	V _{IN} = V _{SS}	1.5	1.8	2.1	N22
		PA9 (OTG_FS_VBUS)	V _{IN} = V _{SS}	0.25	0.37	0.55	

^{1.} All the voltages are measured from the local ground potential.

Note:

When VBUS sensing feature is enabled, PA9 should be left at their default state (floating input), not as alternate function. A typical 200 µA current consumption of the embedded sensing block (current to voltage conversion to determine the different sessions) can be observed on PA9 when the feature is enabled.

^{2.} The USB OTG FS functionality is ensured down to 2.7 V but not the full USB full speed electrical characteristics which are degraded in the 2.7-to-3.0 V V_{DD} voltage range.

^{3.} Guaranteed by design.

^{4.} R_L is the load connected on the USB OTG FS drivers.

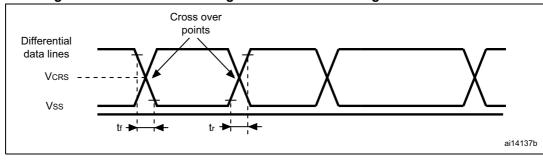


Figure 39. USB OTG FS timings: definition of data signal rise and fall time

Table 64. USB OTG FS electrical characteristics⁽¹⁾

	Driver characteristics						
Symbol	Parameter	Conditions	Min	Max	Unit		
t _r	Rise time ⁽²⁾	C _L = 50 pF	4	20	ns		
t _f	Fall time ⁽²⁾	C _L = 50 pF	4	20	ns		
t _{rfm}	Rise/ fall time matching	t _r /t _f	90	110	%		
V _{CRS}	Output signal crossover voltage		1.3	2.0	V		

^{1.} Guaranteed by design.

6.3.20 12-bit ADC characteristics

Unless otherwise specified, the parameters given in *Table 65* are derived from tests performed under the ambient temperature, f_{PCLK2} frequency and V_{DDA} supply voltage conditions summarized in *Table 14*.

Table 65. ADC characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V_{DDA}	Power supply	V V <12V	1.7 ⁽¹⁾	-	3.6	V
V _{REF+}	Positive reference voltage	$V_{DDA} - V_{REF+} < 1.2 V$	1.7 ⁽¹⁾	-	V_{DDA}	V
f. no	ADC clock frequency	$V_{DDA} = 1.7^{(1)} \text{ to } 2.4 \text{ V}$	0.6	15	18	MHz
f _{ADC}	ADC clock frequency	V _{DDA} = 2.4 to 3.6 V	0.6	30	36	MHz
f _{TRIG} ⁽²⁾	External trigger frequency	f _{ADC} = 30 MHz, 12-bit resolution	-	-	1764	kHz
			-	1	17	1/f _{ADC}
V _{AIN}	Conversion voltage range ⁽³⁾		0 (V _{SSA} or V _{REF} - tied to ground)	-	V _{REF+}	V
R _{AIN} ⁽²⁾	External input impedance	See <i>Equation 1</i> for details	-	-	50	kΩ
R _{ADC} ⁽²⁾⁽⁴⁾	Sampling switch resistance		-	-	6	kΩ
C _{ADC} ⁽²⁾	Internal sample and hold capacitor		-	4	7	pF



Measured from 10% to 90% of the data signal. For more detailed informations, please refer to USB Specification - Chapter 7 (version 2.0).

Table 65. ADC characteristics (continued)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
t _{lat} ⁽²⁾	Injection trigger conversion	f _{ADC} = 30 MHz	-	-	0.100	μs
lat` ´	latency		-	-	3 ⁽⁵⁾	1/f _{ADC}
t _{latr} (2)	Regular trigger conversion	f _{ADC} = 30 MHz	-	-	0.067	μs
'latr`	latency		-	-	2 ⁽⁵⁾	1/f _{ADC}
t _S (2)	Sampling time	f _{ADC} = 30 MHz	0.100	-	16	μs
'S	Camping time		3	-	480	1/f _{ADC}
t _{STAB} ⁽²⁾	Power-up time		-	2	3	μs
		f _{ADC} = 30 MHz 12-bit resolution	0.50	-	16.40	μs
	Total conversion time (including sampling time)	f _{ADC} = 30 MHz 10-bit resolution	0.43	-	16.34	μs
t _{CONV} ⁽²⁾		f _{ADC} = 30 MHz 8-bit resolution	0.37	-	16.27	μs
		f _{ADC} = 30 MHz 6-bit resolution	0.30	-	16.20	μs
		9 to 492 (t _S for sampling approximation)	+n-bit resolution f	or succes	ssive	1/f _{ADC}
		12-bit resolution Single ADC	-	-	2	Msps
f _S ⁽²⁾	Sampling rate (f _{ADC} = 30 MHz, and t _S = 3 ADC cycles)	12-bit resolution Interleave Dual ADC mode	-	-	3.75	Msps
	t _S = 3 ADC cycles)	12-bit resolution Interleave Triple ADC mode	-	-	6	Msps
I _{VREF+} (2)	ADC V _{REF} DC current consumption in conversion mode		-	300	500	μA
I _{VDDA} ⁽²⁾	ADC V _{DDA} DC current consumption in conversion mode		-	1.6	1.8	mA

V_{DDA} minimum value of 1.7 V is possible with the use of an external power supply supervisor (refer to Section 3.15.2: Internal reset OFF).

^{2.} Guaranteed by characterization results.

^{3.} V_{REF+} is internally connected to V_{DDA} and V_{REF-} is internally connected to V_{SSA} .

^{4.} R_{ADC} maximum value is given for V_{DD} =1.7 V, and minimum value for V_{DD} =3.3 V.

^{5.} For external triggers, a delay of 1/f_{PCLK2} must be added to the latency specified in *Table* 65.

Equation 1: R_{AIN} max formula

$$R_{AIN} = \frac{(k-0.5)}{f_{ADC} \times C_{ADC} \times ln(2^{N+2})} - R_{ADC}$$

The formula above (Equation 1) is used to determine the maximum external impedance allowed for an error below 1/4 of LSB. N = 12 (from 12-bit resolution) and k is the number of sampling periods defined in the ADC_SMPR1 register.

Table 66. ADC accuracy at $f_{ADC} = 18 \text{ MHz}^{(1)}$

Symbol	Parameter	Test conditions	Тур	Max ⁽²⁾	Unit
ET	Total unadjusted error	f 40 MH-	±3	±4	
EO	Offset error	f _{ADC} =18 MHz V _{DDA} = 1.7 to 3.6 V	±2	±3	
EG	Gain error	$V_{REF} = 1.7 \text{ to } 3.6 \text{ V}$	±1	±3	LSB
ED	Differential linearity error	$V_{DDA} - V_{REF} < 1.2 V$	±1	±2	
EL	Integral linearity error		±2	±3	

- 1. Better performance could be achieved in restricted V_{DD} , frequency and temperature ranges.
- 2. Guaranteed by characterization results.

Table 67. ADC accuracy at f_{ADC} = 30 MHz⁽¹⁾

Symbol	Parameter	Test conditions	Тур	Max ⁽²⁾	Unit
ET	Total unadjusted error		±2	±5	
EO	Offset error	f _{ADC} = 30 MHz, R _{AIN} < 10 kΩ	±1.5	±2.5	
EG	Gain error	$V_{DDA} = 2.4 \text{ to } 3.6 \text{ V},$	±1.5	±4	LSB
ED	Differential linearity error	V _{REF} = 1.7 to 3.6 V, V _{DDA} –V _{REF} < 1.2 V	±1	±2	
EL	Integral linearity error	DDA ILLI	±1.5	±3	

- 1. Better performance could be achieved in restricted V_{DD}, frequency and temperature ranges.
- 2. Guaranteed by characterization results.

Table 68. ADC accuracy at f_{ADC} = 36 MHz⁽¹⁾

Symbol	Parameter	Test conditions	Тур	Max ⁽²⁾	Unit
ET	Total unadjusted error		±4	±7	
EO	Offset error	f _{ADC} =36 MHz, V _{DDA} = 2.4 to 3.6 V,	±2	±3	
EG	Gain error	$V_{DDA} = 2.4 \text{ to } 3.6 \text{ V}$ $V_{RFF} = 1.7 \text{ to } 3.6 \text{ V}$	±3	±6	LSB
ED	Differential linearity error	V _{DDA} –V _{REF} < 1.2 V	±2	±3	
EL	Integral linearity error		±3	±6	

- 1. Better performance could be achieved in restricted $V_{\mbox{\scriptsize DD}}$, frequency and temperature ranges.
- 2. Guaranteed by characterization results.

Table 69. ADC dynamic accuracy at f_{ADC} = 18 MHz - limited test conditions⁽¹⁾

Symbol	Parameter	Test conditions	Min	Тур	Max	Unit
ENOB	Effective number of bits	f _{ADC} =18 MHz	10.3	10.4	-	bits
SINAD	Signal-to-noise and distortion ratio	$V_{DDA} = V_{REF+} = 1.7 \text{ V}$	64	64.2	-	
SNR	Signal-to-noise ratio	Input Frequency = 20 KHz	64	65	-	dB
THD	Total harmonic distortion	Temperature = 25 °C	-	-72	-67	

^{1.} Guaranteed by characterization results.

Table 70. ADC dynamic accuracy at f_{ADC} = 36 MHz - limited test conditions⁽¹⁾

Symbol	Parameter	Test conditions	Min	Тур	Max	Unit
ENOB	Effective number of bits	f _{ADC} = 36 MHz	10.6	10.8	-	bits
SINAD	Signal-to noise and distortion ratio	$V_{DDA} = V_{REF+} = 3.3 \text{ V}$	66	67	-	
SNR	Signal-to noise ratio	Input Frequency = 20 KHz	64	68	-	dB
THD	Total harmonic distortion	Temperature = 25 °C	-	-72	-70	

^{1.} Guaranteed by characterization results.

Note:

ADC accuracy vs. negative injection current: injecting a negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative currents.

Any positive injection current within the limits specified for $I_{INJ(PIN)}$ and $\Sigma I_{INJ(PIN)}$ in Section 6.3.16 does not affect the ADC accuracy.



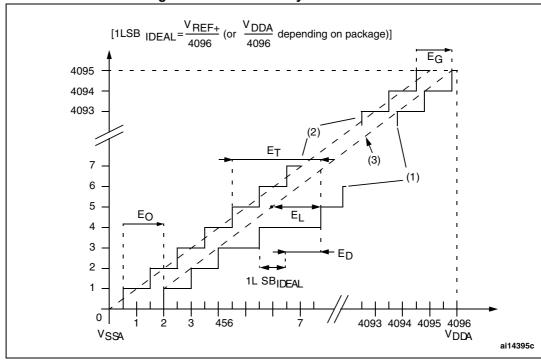


Figure 40. ADC accuracy characteristics

- 1. See also Table 67.
- 2. Example of an actual transfer curve.
- 3. Ideal transfer curve.
- 4. End point correlation line.
- 5. E_T = Total Unadjusted Error: maximum deviation between the actual and the ideal transfer curves. EO = Offset Error: deviation between the first actual transition and the first ideal one. EG = Gain Error: deviation between the last ideal transition and the last actual one. ED = Differential Linearity Error: maximum deviation between actual steps and the ideal one. EL = Integral Linearity Error: maximum deviation between any actual transition and the end point correlation line.

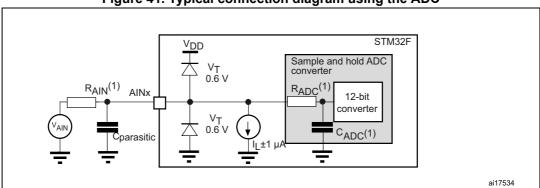


Figure 41. Typical connection diagram using the ADC

- 1. Refer to Table 65 for the values of $\rm R_{AIN},\, R_{ADC}$ and $\rm C_{ADC}.$
- C_{parasitic} represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (roughly 5 pF). A high C_{parasitic} value downgrades conversion accuracy. To remedy this, f_{ADC} should be reduced.



General PCB design guidelines

Power supply decoupling should be performed as shown in Figure 42 or Figure 43, depending on whether V_{REF+} is connected to V_{DDA} or not. The 10 nF capacitors should be ceramic (good quality). They should be placed them as close as possible to the chip.

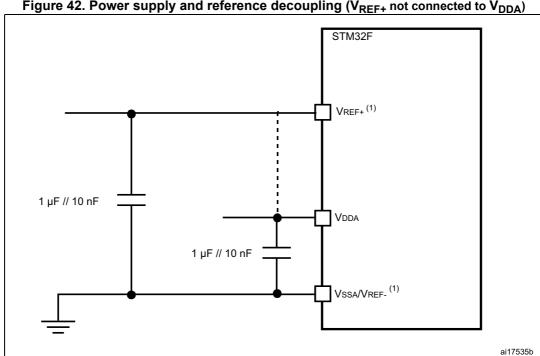


Figure 42. Power supply and reference decoupling (V_{REF+} not connected to V_{DDA})



 V_{REF+} and V_{REF-} inputs are both available on UFBGA100. V_{REF+} is also available on LQFP100. When V_{REF+} and V_{REF-} are not available, they are internally connected to V_{DDA} and V_{SSA} .

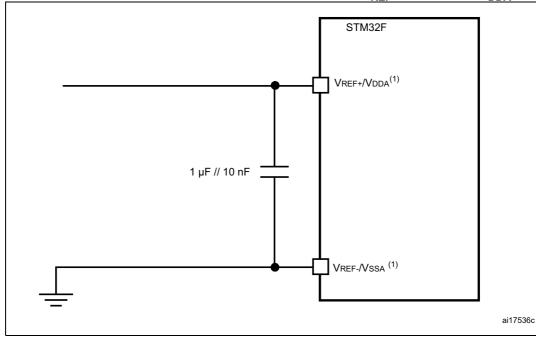


Figure 43. Power supply and reference decoupling (V_{REF+} connected to V_{DDA})

6.3.21 Temperature sensor characteristics

Table 71. Temperature sensor characteristics

Symbol	Parameter		Тур	Max	Unit		
T _L ⁽¹⁾	V _{SENSE} linearity with temperature	-	±1	<u>+2</u>	°C		
Avg_Slope ⁽¹⁾	Average slope	-	2.5	1	mV/°C		
V ₂₅ ⁽¹⁾	Voltage at 25 °C	-	0.76	-	٧		
t _{START} (2)	Startup time	-	6	10	μs		
T _{S_temp} ⁽²⁾	ADC sampling time when reading the temperature (1 °C accuracy)	10	-	-	μs		

^{1.} Guaranteed by characterization results.

Table 72. Temperature sensor calibration values

Symbol	Parameter	Memory address
TS_CAL1	TS ADC raw data acquired at temperature of 30 °C, V _{DDA} = 3.3 V	0x1FFF 7A2C - 0x1FFF 7A2D
TS_CAL2	TS ADC raw data acquired at temperature of 110 °C, V _{DDA} = 3.3 V	0x1FFF 7A2E - 0x1FFF 7A2F

^{1.} V_{REF+} and V_{REF-} inputs are both available on UFBGA100. V_{REF+} is also available on LQFP100. When V_{REF+} and V_{REF-} are not available, they are internally connected to V_{DDA} and V_{SSA} .

^{2.} Guaranteed by design.

6.3.22 V_{BAT} monitoring characteristics

Table 73. V_{BAT} monitoring characteristics

Symbol	Parameter		Тур	Max	Unit
R	Resistor bridge for V _{BAT}		50	-	ΚΩ
Q	Ratio on V _{BAT} measurement		4	-	
Er ⁽¹⁾	Error on Q	-1	-	+1	%
T _{S_vbat} ⁽²⁾⁽²⁾	ADC sampling time when reading the V _{BAT} 1 mV accuracy		-	-	μs

^{1.} Guaranteed by design.

6.3.23 Embedded reference voltage

The parameters given in *Table 74* are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 14*.

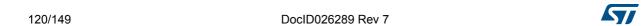
Table 74. Embedded internal reference voltage

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V _{REFINT}	Internal reference voltage	- 40 °C < T _A < + 125 °C	1.18	1.21	1.24	V
T _{S_vrefint} (1)	ADC sampling time when reading the internal reference voltage	-	10	-	-	μs
V _{RERINT_s} ⁽²⁾	Internal reference voltage spread over the temperature range	V _{DD} = 3V ± 10mV	-	3	5	mV
T _{Coeff} ⁽²⁾	Temperature coefficient	-	-	30	50	ppm/°C
t _{START} (2)	Startup time	-	-	6	10	μs

^{1.} Shortest sampling time can be determined in the application by multiple iterations.

Table 75. Internal reference voltage calibration values

Symbol	Parameter	Memory address
V _{REFIN_CAL}	Raw data acquired at temperature of 30 °C V _{DDA} = 3.3 V	0x1FFF 7A2A - 0x1FFF 7A2B



^{2.} Shortest sampling time can be determined in the application by multiple iterations.

^{2.} Guaranteed by design.

6.3.24 SD/SDIO MMC/eMMC card host interface (SDIO) characteristics

Unless otherwise specified, the parameters given in *Table 76* for the SDIO/MMC/eMMC interface are derived from tests performed under the ambient temperature, f_{PCLK2} frequency and V_{DD} supply voltage conditions summarized in *Table 14*, with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 10
- Capacitive load C = 30 pF (for eMMC C = 20 pF)
- Measurement points are done at CMOS levels: 0.5V_{DD}

Refer to Section 6.3.16: I/O port characteristics for more details on the input/output characteristics.

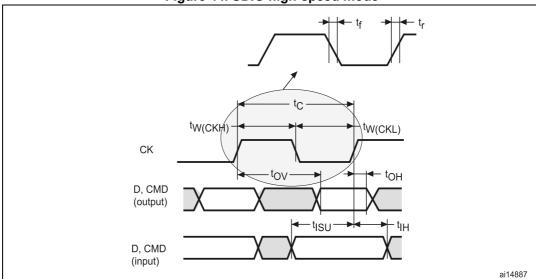


Figure 44. SDIO high-speed mode



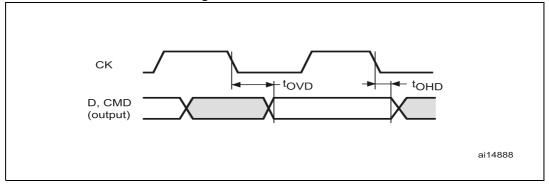


Table 76. Dynamic characteristics: SD / MMC characteristics⁽¹⁾⁽²⁾

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f _{PP}	Clock frequency in data transfer mode	-	0	-	50	MHz
-	SDIO_CK/fPCLK2 frequency ratio	-	-	-	8/3	-
t _{W(CKL)}	Clock low time	fpp = 50 MHz	10.5	11	-	no
t _{W(CKH)}	Clock high time	fpp = 50 MHz	8.5	9	-	ns
CMD, D ir	nputs (referenced to CK) in MMC and	d SD HS mode				
t _{ISU}	Input setup time HS	fpp = 50 MHz	2.5	-	-	
t _{IH}	Input hold time HS	fpp = 50 MHz -40°C <t<sub>A< 125°C</t<sub>	5	-	-	ns
		fpp = 50 MHz -40°C <t<sub>A<+85°C</t<sub>	2.5	-	-	
CMD, D o	utputs (referenced to CK) in MMC a	nd SD HS mode				
t _{OV}	Output valid time HS	fpp = 50 MHz	-	3.5	4	T
t _{OH}	Output hold time HS	fpp = 50 MHz	2	-	-	- ns
CMD, D ir	nputs (referenced to CK) in SD defau	ılt mode				
t _{ISUD}	Input setup time SD	fpp = 25 MHz	3	-	-	
t _{IHD}	Input hold time SD	fpp = 25 MHz	4	-	-	– ns
CMD, D o	utputs (referenced to CK) in SD def	ault mode				
t _{OVD}	Output valid default time SD	fpp =25 MHz	-	5	5.5	T
t _{OHD}	Output hold default time SD	fpp =25 MHz	4.5	-	-	- ns

^{1.} Guaranteed by characterization results.

^{2.} $V_{DD} = 2.7 \text{ to } 3.6 \text{ V}.$

6.3.25 RTC characteristics

Table 77. Dynamic characteristics: eMMC characteristics V_{DD} = 1.7 V to 1.9 $V^{(1)(2)}$

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f _{PP}	Clock frequency in data transfer mode	-	0	-	50	MHz
-	SDIO_CK/fPCLK2 frequency ratio	-	-	-	8/3	-
t _{W(CKL)}	Clock low time	fpp = 50 MHz	10	10.5	-	no
t _{W(CKH)}	Clock high time	fpp = 50 MHz	9	9.5	-	ns
CMD, D ir	nputs (referenced to CK) in eMMC mo	ode		•	•	•
t _{ISU}	Input setup time HS	fpp = 50 MHz	0	-	-	ns
t _{IH}	Input hold time HS	fpp = 50 MHz	6	-	-	
CMD, D o	utputs (referenced to CK) in eMMC n	node				•
t _{OV}	Output valid time HS	fpp = 50 MHz	-	3.5	5	no
t _{OH}	Output hold time HS	fpp = 50 MHz	2	-	-	ns

^{1.} Guaranteed by characterization results.

Table 78. RTC characteristics

Symbol	Parameter	arameter Conditions		Max
-	f _{PCLK1} /RTCCLK frequency ratio	Any read/write operation from/to an RTC register	4	-

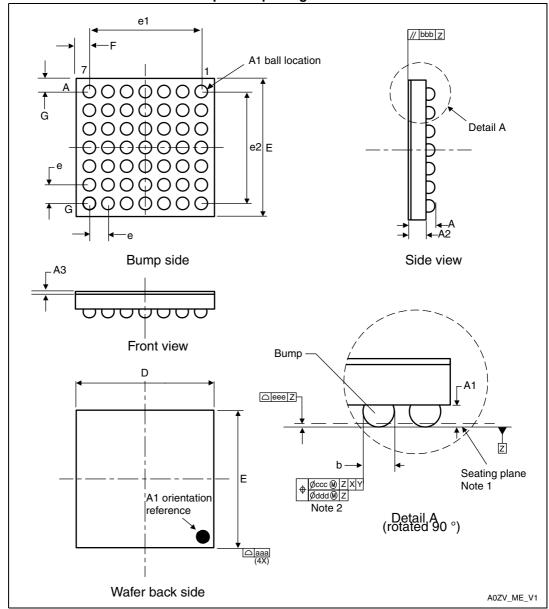
^{2.} C_{load} = 20 pF

7 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK[®] packages, depending on their level of environmental compliance. ECOPACK[®] specifications, grade definitions and product status are available at: www.st.com. ECOPACK[®] is an ST trademark.

7.1 WLCSP49 package information

Figure 46. WLCSP49 - 49-ball, 2.999 x 3.185 mm, 0.4 mm pitch wafer level chip scale package outline



1. Drawing is not to scale.

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Table 79. WLCSP49 - 49-ball, 2.999 x 3.185 mm, 0.4 mm pitch wafer level chip scale package mechanical data

Symphol		millimeters	g		inches ⁽¹⁾	
Symbol	Min	Тур	Max	Min	Тур	Max
Α	0.525	0.555	0.585	0.0207	0.0219	0.0230
A1	-	0.175	-	-	0.0069	-
A2	-	0.380	-	-	0.0150	-
A3 ⁽²⁾	-	0.025	-	-	0.0010	-
b ⁽³⁾	0.220	0.250	0.280	0.0087	0.0098	0.0110
D	2.964	2.999	3.034	0.1167	0.1181	0.1194
Е	3.150	3.185	3.220	0.1240	0.1254	0.1268
е	-	0.400	-	-	0.0157	-
e1	-	2.400	-	-	0.0945	-
e2	-	2.400	-	-	0.0945	-
F	-	0.2995	-	-	0.0118	-
G	-	0.3925	-	-	0.0155	-
aaa	-	0.100	-	-	0.0039	-
bbb	-	0.100	-	-	0.0039	-
ccc	-	0.100	-	-	0.0039	-
ddd	-	0.050	-	-	0.0020	-
eee	-	0.050	-	-	0.0020	-

- 1. Values in inches are converted from mm and rounded to 4 decimal digits.
- 2. Back side coating
- 3. Dimension is measured at the maximum bump diameter parallel to primary datum Z.

Figure 47. WLCSP49 - 49-ball, 2.999 x 3.185 mm, 0.4 mm pitch wafer level chip scale recommended footprint

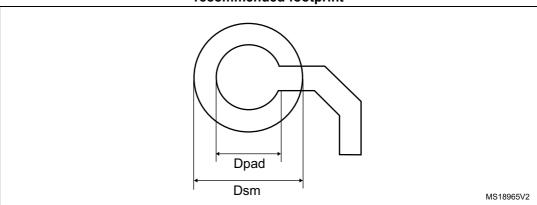


Table 80. WLCSP49 recommended PCB design rules (0.4 mm pitch)

Dimension	Recommended values
Pitch	0.4 mm
Dpad	260 µm max. (circular) 220 µm recommended
Dsm	300 μm min. (for 260 μm diameter pad)
PCB pad design	Non-solder mask defined via underbump allowed

Device marking for WLCSP49

The following figure gives an example of topside marking orientation versus ball A1 identifier location.

Other optional marking or inset/upset marks, which depend on supply chain operations, are not indicated below.

Ball 1 indentifier Product identification⁽¹⁾ E411CEB Revision code R MSv36161V1

Figure 48. WLCSP49 marking (package top view)

Parts marked as ES or E or accompanied by an Engineering Sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

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7.2 UFQFPN48 package information

Pin 1 identifier laser marking area D Seating Detail Y D Exposed pad D2 area booooolooooob C 0.500x45° pin1 corner R 0.125 typ. Detail Z E2 A0B9_ME_V3

Figure 49. UFQFPN48 - 48-lead, 7 x 7 mm, 0.5 mm pitch, ultra thin fine pitch quad flat package outline

- 1. Drawing is not to scale.
- 2. All leads/pads should also be soldered to the PCB to improve the lead/pad solder joint life.
- There is an exposed die pad on the underside of the UFQFPN package. It is recommended to connect and solder this back-side pad to PCB ground.

Table 81. UFQFPN48 - 48-lead, 7 x 7 mm, 0.5 mm pitch, ultra thin fine pitch quad flat package mechanical data

Symbol		millimeters				
Symbol	Min.	Тур.	Max.	Min.	Тур.	Max.
Α	0.500	0.550	0.600	0.0197	0.0217	0.0236
A1	0.000	0.020	0.050	0.0000	0.0008	0.0020
D	6.900	7.000	7.100	0.2717	0.2756	0.2795
Е	6.900	7.000	7.100	0.2717	0.2756	0.2795
D2	5.500	5.600	5.700	0.2165	0.2205	0.2244

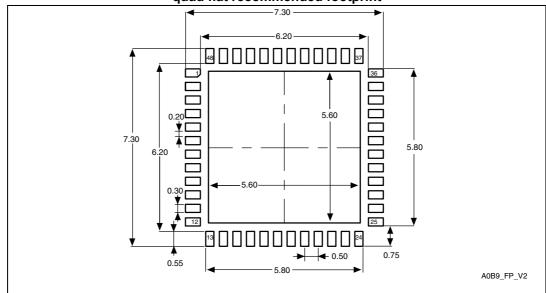


Table 81. UFQFPN48 - 48-lead, 7 x 7 mm, 0.5 mm pitch, ultra thin fine pitch quad flat package mechanical data (continued)

Sumb al		millimeters				
Symbol	Min.	Тур.	Max.	Min. Typ.		Max.
E2	5.500	5.600	5.700	0.2165	0.2205	0.2244
L	0.300	0.400	0.500	0.0118	0.0157	0.0197
Т	-	0.152	-	-	0.0060	-
b	0.200	0.250	0.300	0.0079	0.0098	0.0118
е	-	0.500	-	-	0.0197	-
ddd	-	-	0.080	-	-	0.0031

^{1.} Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 50. UFQFPN48 - 48-lead, 7 x 7 mm, 0.5 mm pitch, ultra thin fine pitch quad flat recommended footprint



1. Dimensions are in millimeters.

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Device marking for UFQFPN48

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

Other optional marking or inset/upset marks, which depend on supply chain operations, are not indicated below.

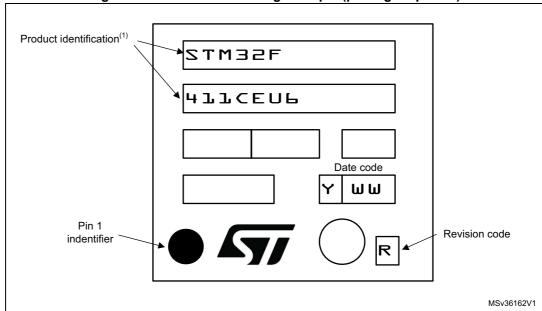


Figure 51. UFQFPN48 marking example (package top view)

^{1.} Parts marked as ES or E or accompanied by an Engineering Sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

7.3 LQFP64 package information

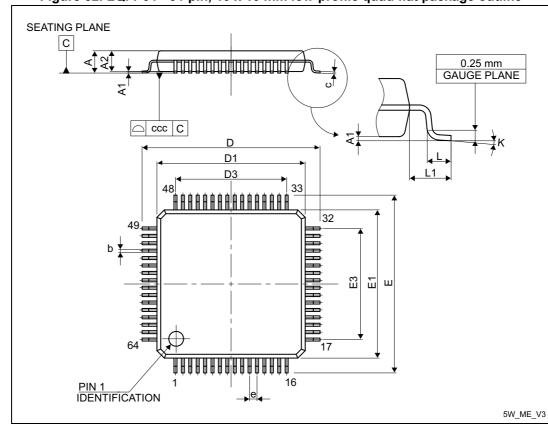


Figure 52. LQFP64 - 64-pin, 10 x 10 mm low-profile quad flat package outline

1. Drawing is not to scale.

Table 82. LQFP64 - 64-pin, 10 x 10 mm low-profile quad flat package mechanical data

	millimeters			inches ⁽¹⁾		
Symbol	Min	Тур	Max	Min	Тур	Max
Α	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
С	0.090	-	0.200	0.0035	-	0.0079
D	-	12.000	-	-	0.4724	-
D1	-	10.000	-	-	0.3937	-
D3	-	7.500	-	-	0.2953	-
E	-	12.000	-	-	0.4724	-
E1	-	10.000	-	-	0.3937	-

577

CCC

millimeters inches⁽¹⁾ **Symbol** Min Тур Max Min Тур Max E3 7.500 0.2953 е 0.500 0.0197 Κ 0° 3.5° 7° 0° 3.5° 7° L 0.450 0.600 0.750 0.0177 0.0236 0.0295 L1 1.000 0.0394 0.080 0.0031

Table 82. LQFP64 - 64-pin, 10 x 10 mm low-profile quad flat package mechanical data (continued)

^{1.} Values in inches are converted from mm and rounded to 4 decimal digits.

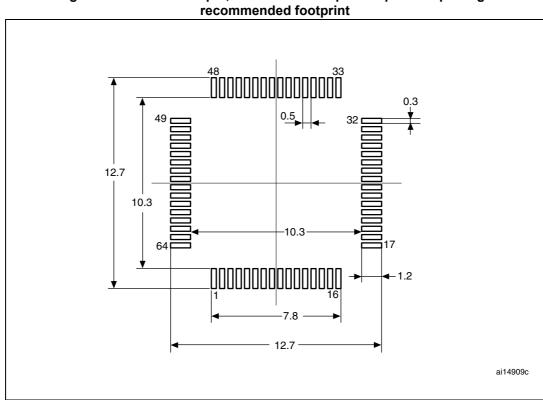


Figure 53. LQFP64 - 64-pin, 10 x 10 mm low-profile quad flat package

1. Dimensions are expressed in millimeters.

Device marking for LQFP64

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

Other optional marking or inset/upset marks, which depend on supply chain operations, are not indicated below.

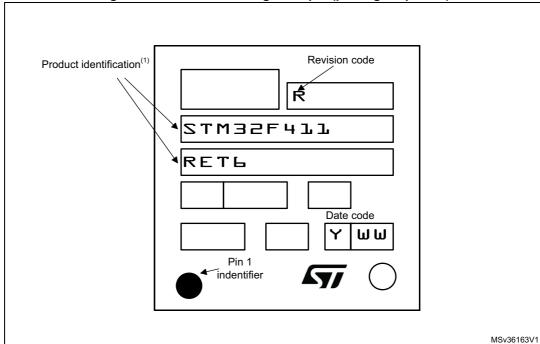


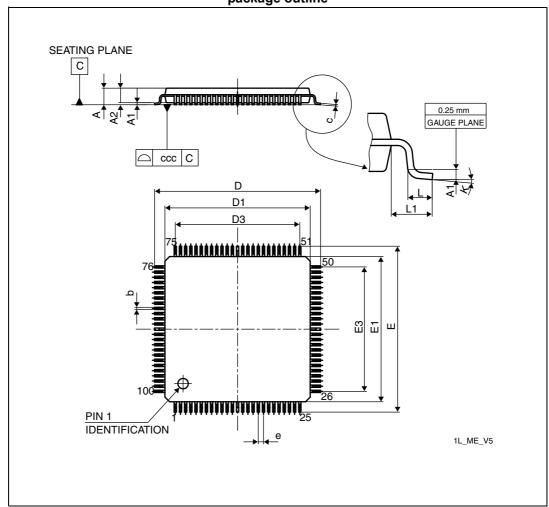
Figure 54. LQFP64 marking example (package top view)

1. Parts marked as ES or E or accompanied by an Engineering Sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.



7.4 LQFP100 package information

Figure 55. LQFP100 - 100-pin, 14 x 14 mm, 100-pin low-profile quad flat package outline



1. Drawing is not to scale.

Table 83. LQPF100 - 100-pin, 14 x 14 mm, 100-pin low-profile quad flat package mechanical data

Counch al		millimeters			inches ⁽¹⁾		
Symbol	Min.	Тур.	Max.	Min.	Тур.	Max.	
Α	-	-	1.60	-	-	0.063	
A1	0.050	-	0.150	0.002	-	0.0059	
A2	1.350	1.40	1.450	0.0531	0.0551	0.0571	
b	0.170	0.220	0.270	0.0067	0.0087	0.0106	
С	0.090	-	0.200	0.0035	-	0.0079	
D	15.800	16.000	16.200	0.622	0.6299	0.6378	
D1	13.800	14.000	14.200	0.5433	0.5512	0.5591	
D3	-	12.000	-	-	0.4724	-	
E	15.800	16.000	16.200	0.622	0.6299	0.6378	
E1	13.800	14.000	14.200	0.5433	0.5512	0.5591	
E3	-	12.000	-	-	0.4724	-	
е	-	0.500	-	-	0.0197	-	
L	0.450	0.600	0.750	0.0177	0.0236	0.0295	
L1	-	1.000	-	-	0.0394	-	
K	0.0°	3.5°	7.0°	0.0°	3.5°	7.0°	
CCC	0.080				0.0031		

^{1.} Values in inches are converted from mm and rounded to 4 decimal digits.

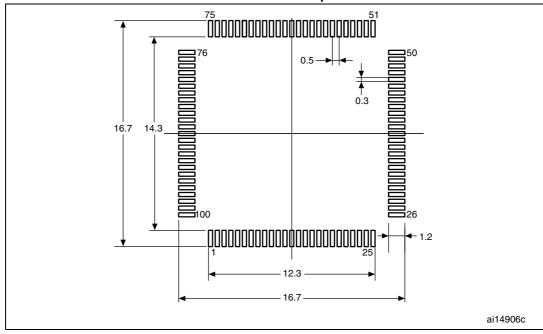


Figure 56. LQFP100 - 100-pin, 14 x 14 mm, 100-pin low-profile quad flat recommended footprint

1. Dimensions are in millimeters.

Device marking for LQFP100

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

Other optional marking or inset/upset marks, which depend on supply chain operations, are not indicated below.

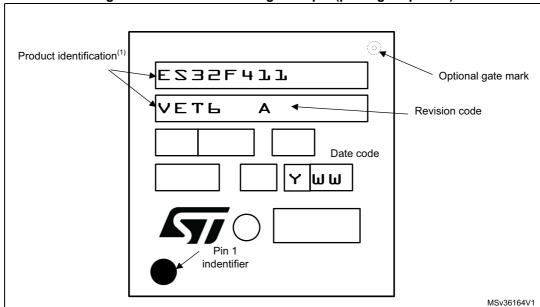


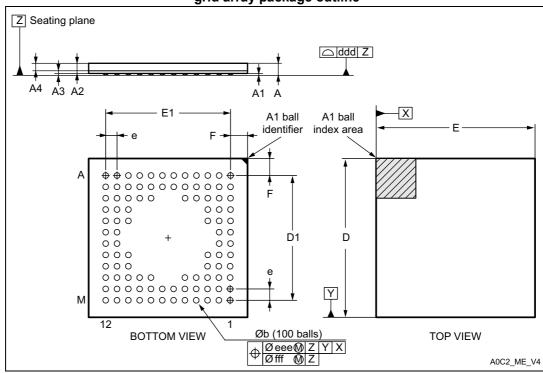
Figure 57. LQPF100 marking example (package top view)

1. Parts marked Parts marked as ES or E or accompanied by an Engineering Sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.



7.5 UFBGA100 package information

Figure 58. UFBGA100 - 100-ball, 7 x 7 mm, 0.50 mm pitch, ultra fine pitch ball grid array package outline



1. Drawing is not to scale.

Table 84. UFBGA100 - 100-ball, 7 x 7 mm, 0.50 mm pitch, ultra fine pitch ball grid array package mechanical data

Symbol		millimeters			inches ⁽¹⁾		
Symbol	Min.	Тур.	Max.	Min.	Тур.	Max.	
Α	0.460	0.530	0.600	0.0181	0.0209	0.0236	
A1	0.050	0.080	0.110	0.0020	0.0031	0.0043	
A2	0.400	0.450	0.500	0.0157	0.0177	0.0197	
A3	-	0.130	-	-	0.0051	-	
A4	0.270	0.320	0.370	0.0106	0.0126	0.0146	
b	0.200	0.250	0.300	0.0079	0.0098	0.0118	
D	6.950	7.000	7.050	0.2736	0.2756	0.2776	
D1	5.450	5.500	5.550	0.2146	0.2165	0.2185	
Е	6.950	7.000	7.050	0.2736	0.2756	0.2776	
E1	5.450	5.500	5.550	0.2146	0.2165	0.2185	
е	-	0.500	-	-	0.0197	-	
F	0.700	0.750	0.800	0.0276	0.0295	0.0315	

Table 84. UFBGA100 - 100-ball, 7 x 7 mm, 0.50 mm pitch, ultra fine pitch ball grid array package mechanical data (continued)

Symbol		millimeters		inches ⁽¹⁾		
Symbol	Min.	Тур.	Max.	Min.	Тур.	Max.
ddd	-	-	0.100	-	-	0.0039
eee	-	-	0.150	-	-	0.0059
fff	-	-	0.050	-	-	0.0020

^{1.} Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 59. UFBGA100 - 100-ball, 7 x 7 mm, 0.50 mm pitch, ultra fine pitch ball grid array recommended footprint

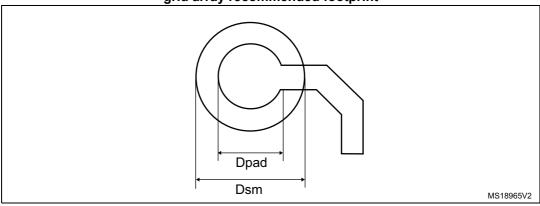


Table 85. UFBGA100 recommended PCB design rules (0.5 mm pitch BGA)

Dimension	Recommended values
Pitch	0.5
Dpad	0.27 mm
Dsm	0.35 mm typ. (depends on the soldermask registration tolerance)
Solder paste	0.27 mm aperture diameter.

- 1. Non-solder mask defined (NSMD) pads are recommended.
- 2. 4 to 6 mils solder paste screen printing process.

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Device marking for UFBGA100

The following figure gives an example of topside marking orientation versus ball A1 identifier location.

Other optional marking or inset/upset marks, which depend on supply chain operations, are not indicated below.

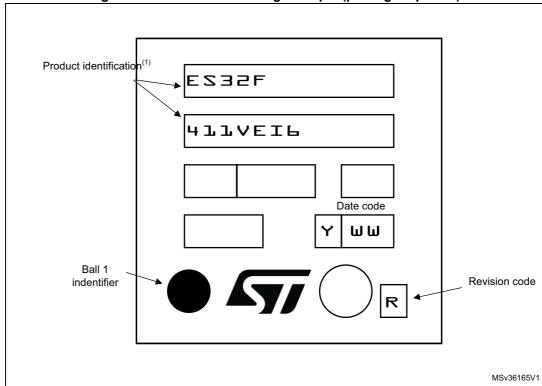


Figure 60. UFBGA100 marking example (package top view)

^{1.} Parts marked as ES or E or accompanied by an Engineering Sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

7.6 Thermal characteristics

The maximum chip junction temperature (T_Jmax) must never exceed the values given in *Table 14: General operating conditions on page 59*.

The maximum chip-junction temperature, T_J max., in degrees Celsius, may be calculated using the following equation:

 $T_J \max = T_A \max + (PD \max x \Theta_{JA})$

Where:

- T_A max is the maximum ambient temperature in °C,
- Θ_{JA} is the package junction-to-ambient thermal resistance, in $^{\circ}$ C/W,
- PD max is the sum of P_{INT} max and $P_{I/O}$ max (PD max = P_{INT} max + $P_{I/O}$ max),
- P_{INT} max is the product of I_{DD} and V_{DD}, expressed in Watts. This is the maximum chip internal power.

P_{I/O} max represents the maximum power dissipation on output pins where:

$$P_{I/O} \max = \sum (V_{OL} \times I_{OL}) + \sum ((V_{DD} - V_{OH}) \times I_{OH}),$$

taking into account the actual V_{OL} / I_{OL} and V_{OH} / I_{OH} of the I/Os at low and high level in the application.

Symbol	Parameter	Value	Unit
	Thermal resistance junction-ambient UFQFPN48	32	
	Thermal resistance junction-ambient WLCSP49	51	
$\Theta_{\sf JA}$	Thermal resistance junction-ambient LQFP64	47	°C/W
	Thermal resistance junction-ambient LQFP100	43	
	Thermal resistance junction-ambient UFBGA100	62	

Table 86. Package thermal characteristics

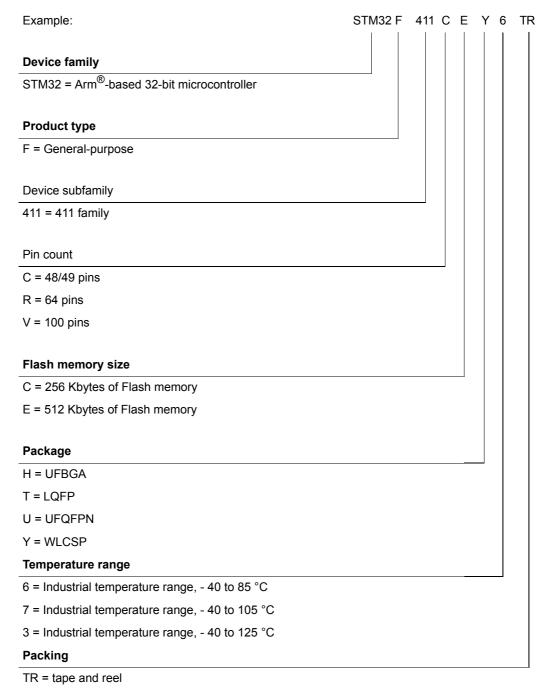
7.6.1 Reference document

JESD51-2 Integrated Circuits Thermal Test Method Environment Conditions - Natural Convection (Still Air). Available from www.jedec.org.

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Ordering information 8

Table 87. Ordering information scheme



No character = tray or tube

Recommendations when using the internal Appendix A reset OFF

When the internal reset is OFF, the following integrated features are no longer supported:

- The integrated power-on-reset (POR)/power-down reset (PDR) circuitry is disabled.
- The brownout reset (BRO) circuitry must be disabled. By default BOR is OFF.
- The embedded programmable voltage detector (PVD) is disabled.
- V_{BAT} functionality is no more available and VBAT pin should be connected to V_{DD}.

A.1 Operating conditions

Table 88. Limitations depending on the operating power supply range

Operating power supply range	ADC operation	Maximum Flash memory access frequency with no wait state (f _{Flashmax})	Maximum Flash memory access frequency with no wait states ⁽¹⁾ (2)	I/O operation	Possible Flash memory operations
$V_{DD} = 1.7 \text{ to}$ 2.1 $V^{(3)}$	Conversion time up to 1.2 Msps	20 MHz ⁽⁴⁾	100 MHz with 6 wait states	No I/O compensation	8-bit erase and program operations only

Applicable only when the code is executed from Flash memory. When the code is executed from RAM, no wait state is required.

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^{2.} Thanks to the ART accelerator and the 128-bit Flash memory, the number of wait states given here does not impact the execution speed from Flash memory since the ART accelerator allows to achieve a performance equivalent to 0 wait state program execution.

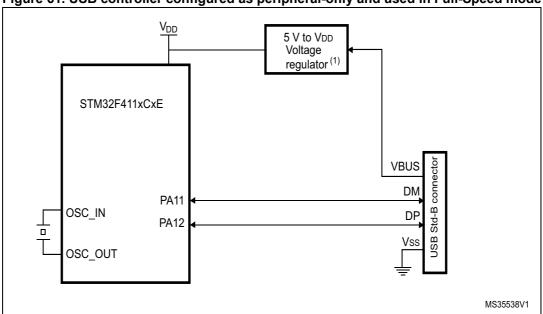
V_{DD}/V_{DDA} minimum value of 1.7 V, with the use of an external power supply supervisor (refer to Section 3.15.1: Internal reset ON).

^{4.} Prefetch is not available. Refer to AN3430 application note for details on how to adjust performance and power.

Appendix B Application block diagrams

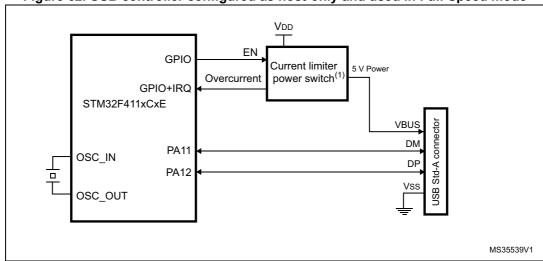
B.1 USB OTG Full Speed (FS) interface solutions

Figure 61. USB controller configured as peripheral-only and used in Full-Speed mode



1. The external voltage regulator is only needed when building a $V_{\mbox{\scriptsize BUS}}$ powered device.

Figure 62. USB controller configured as host-only and used in Full-Speed mode



 The current limiter is required only if the application has to support a V_{BUS} powered device. A basic power switch can be used if 5V are available on the application board.

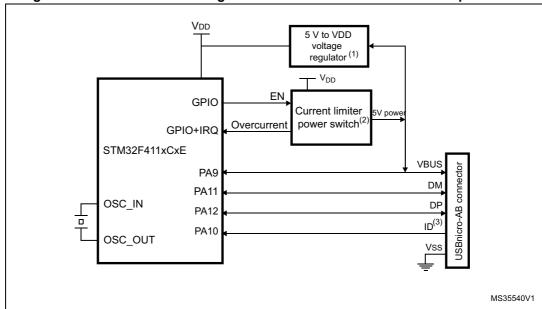


Figure 63. USB controller configured in dual mode and used in Full-Speed mode

- 1. The external voltage regulator is only needed when building a $V_{\mbox{\scriptsize BUS}}$ powered device.
- The current limiter is required only if the application has to support a V_{BUS} powered device. A basic power switch can be used if 5 V are available on the application board.
- 3. The ID pin is required in dual role only.

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B.2 Sensor Hub application example

Accelerometer - Gyroscope Magnetometer STM32F411xE 48- and 49-pin package PB6/PB10/PA8 I2C Pressure PB7/PB9/PB4 Ambient light PB13 15x GPIO I2S PB15 Proximity воото PA9 VDD **UART** - Micro PA 10 PDRON SWDIC NSS PA4 JTAG SWCL PA 14 SCK PA5 SPI SWO PB3 - HOST MISO PA6 NRST MOSI PA7 OSC 32k PC14 PA1/PA3 Temperature/Humidity PC15 Up to 10 ADC inputs possible for the 48 and 49 pins package MS35548V1

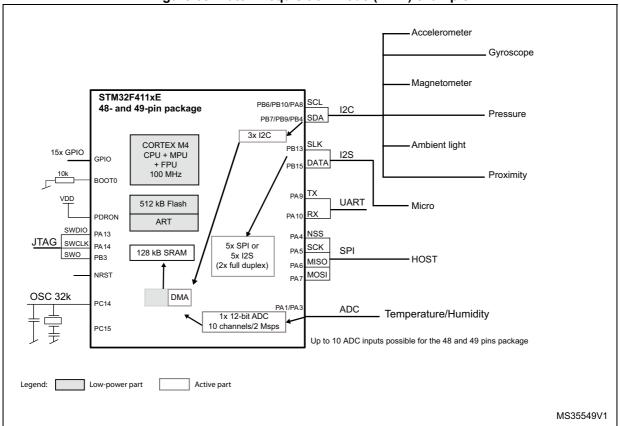
Figure 64. Sensor Hub application example

B.3 Batch Acquisition Mode (BAM) example

Data is transferred through the DMA from interfaces into the internal SRAM while the rest of the MCU is set in low power mode.

- Code execution from RAM before switching off the Flash.
- Flash is set in power down and flash interface (ART™ accelerator) clock is stopped.
- The clocks are enabled only for the required interfaces.
- MCU core is set in sleep mode (core clock stopped waiting for interrupt).
- Only the needed DMA channels are enabled and running.

Figure 65. Batch Acquisition Mode (BAM) example



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Revision history

Table 89. Document revision history

Date	Revision	Changes
19-Jun-2014	1	Initial release.
10-Sep-2014	2	Introduced the BAM feature in Features, Section 2: Description., and Section 3.3: Batch Acquisition mode (BAM). Updated Section 3.5: Embedded Flash memory, Section 3.14: Power supply schemes and Section 3.18: Low-power modes, Section 3.20.2: General-purpose timers (TIMx) and Section 3.30: Temperature sensor. Modified Table 8: STM32F411xC/xE pin definitions, Table 9: Alternate function mapping and APB2 in Table 10: STM32F411xC/xE register boundary addresses. Modified Table 34: Low-power mode wakeup timings ⁽¹⁾ , Table 20: Typical and maximum current consumption, code with data processing (ART accelerator disabled) running from SRAM - V_{DD} = 1.7 V, Table 21: Typical and maximum current consumption, code with data processing (ART accelerator disabled) running from SRAM - V_{DD} = 3.6 V, Table 25: Typical and maximum current consumption in run mode, code with data processing (ART accelerator enabled with prefetch) running from Flash memory - V_{DD} = 3.6 V, Table 26: Typical and maximum current consumption in Sleep mode - V_{DD} = 3.6 V and Table 58: I^2 C characteristics and Figure 33: I^2 C bus AC waveforms and
		measurement circuit. Added Figure 21: Low-power mode wakeup, Section Appendix A: Recommendations when using the internal reset OFF and Section Appendix B: Application block diagrams.
27-Nov-2014	3	Changed datasheet status to Production Data. Updated <i>Table 31: Typical and maximum current consumptions in V_{BAT} mode.</i> Section: On-chip peripheral current consumption: changed HCLK frequency and updated DMA1 and DMA2 current consumption in <i>Table 33: Peripheral current consumption.</i> Updated <i>Table 55: I/O AC characteristics.</i> Updated THD in <i>Table 69: ADC dynamic accuracy at f_{ADC} = 18 MHz - limited test conditions</i> and <i>Table 70: ADC dynamic accuracy at f_{ADC} = 36 MHz - limited test conditions.</i> Updated <i>Table 55: I/O AC characteristics.</i> Updated <i>Table 55: I/O AC characteristics.</i> Updated <i>Figure 46: WLCSP49 - 49-ball, 2.999 x 3.185 mm, 0.4 mm pitch wafer level chip scale package outline</i> and <i>Figure 48: WLCSP49 marking (package top view).</i> Added <i>Figure 47: WLCSP49 - 49-ball, 2.999 x 3.185 mm, 0.4 mm pitch wafer level chip scale recommended footprint</i> and <i>Table 80: WLCSP49 recommended PCB design rules (0.4 mm pitch).</i> Updated <i>Figure 51: UFQFPN48 marking example (package top view), Figure 54: LQFP64 marking example (package top view), Figure 57: LQPF100 marking example (package top view), and Figure 84: UFBGA100 - 100-ball, 7 x 7 mm, 0.50 mm pitch, ultra fine pitch ball grid array package mechanical data.</i>
04-Feb-2015	4	Added VPP alternate function for BOOT0 in <i>Table 8: STM32F411xC/xE pin definitions</i> . Added TC inputs in <i>Table 11: Voltage characteristics, Table 12: Current characteristics, Table 14: General operating conditions, Table 53: I/O static characteristics</i> and <i>Figure 30: FT/TC I/O input characteristics</i> . Updated V _{ESD(CDM)} in <i>Table 50: ESD absolute maximum ratings</i> . A3 minimum and maximum values removed in <i>Table 83: UFBGA100 - 100-ball, 7 x 7 mm, 0.50 mm pitch, ultra fine pitch ball grid array package mechanical data</i> .



Table 89. Document revision history

Date	Revision	Changes
21-Nov-2016	5	Updated: - Features - Figure 1: Compatible board design for LQFP100 package - Figure 2: Compatible board design for LQFP64 package - Figure 3: STM32F411xC/xE block diagram - Figure 22: High-speed external clock source AC timing diagram - Figure 23: Low-speed external clock source AC timing diagram - Figure 33: I²C bus AC waveforms and measurement circuit - Figure 58: UFBGA100 - 100-ball, 7 x 7 mm, 0.50 mm pitch, ultra fine pitch ball grid array package outline - Table 2: STM32F411xC/xE features and peripheral counts - Table 8: STM32F411xC/xE pin definitions - Table 13: Thermal characteristics - Table 14: General operating conditions - From Table 20: Typical and maximum current consumption, code with data processing (ART accelerator disabled) running from SRAM - V _{DD} = 1.7 V to Table 31: Typical and maximum current consumptions in V _{BAT} mode - Table 35: High-speed external user clock characteristics - Table 36: Low-speed external user clock characteristics - Table 39: HSI oscillator characteristics - Table 51: Electrical sensitivities - Table 51: Electrical sensitivities - Table 53: I/O static characteristics - Table 87: Ordering information scheme Added: - One-time programmable bytes - Table 86: Package thermal characteristics
05-Dec-2016	6	Updated: - Table 27: Typical and maximum current consumptions in Stop mode - V_{DD} = 1.7 V - Table 28: Typical and maximum current consumption in Stop mode - V_{DD} =3.6 V - Table 29: Typical and maximum current consumption in Standby mode - V_{DD} = 1.7 V - Table 30: Typical and maximum current consumption in Standby mode - V_{DD} = 3.6 V
14-Dec-2017	7	Updated: — Table 27: Typical and maximum current consumptions in Stop mode - V_{DD} = 1.7 V — Table 28: Typical and maximum current consumption in Stop mode - V_{DD} =3.6 V — Table 29: Typical and maximum current consumption in Standby mode - V_{DD} = 1.7 V — Table 30: Typical and maximum current consumption in Standby mode - V_{DD} = 3.6 V



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